# Massachusetts Institute of Technology Department of Electrical Engineering and Computer Science <br> <br> 6.002 - Circuits and Electronics <br> <br> 6.002 - Circuits and Electronics <br> <br> Spring 2003 

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## Handout S03-013 - Homework \#2

## Issued: Wed. Feb 12

Due: Fri. Feb 21

Note: Do your work on problems 2.2 through 2.4 directly on these sheets and turn them in with your solutions to the remaining problems.

Problem 2.1: Imagine a device called a "widget" which has the electrical characteristics shown below:


There is no electrical connection between the input terminals and the output terminals.
When a signal of 0 Volts is applied to $a a^{\prime}$, the resistance between b and $\mathrm{b}^{\prime}$ is negligibly small, and may be approximated by a short circuit.

When a $\pm 5 \mathrm{~V}$ signal is applied to aa', the resistance between b and $\mathrm{b}^{\prime}$ is very large, and may be approximated by an open circuit.

In the following problems, you are free to use constant voltage sources and resistors.
(A) Can widgets be combined to create a NAND gate? Demonstrate.
(B) Can widgets be combined to create a AND gate? Demonstrate.
(C) Can widgets be combined to create a NOR gate? Demonstrate.
(D) Can widgets be combined to create a OR gate? Demonstrate.

Name: $\qquad$

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Problem 2.2: In each of the following circuits, determine the values of the indicated voltages and/or currents.. (Voltages are in volts and currents are in amperes.)
(a)


$$
\begin{aligned}
& v_{1}= \\
& v_{2}=\square \\
& v_{3}=\square \\
& v_{1}=\square \\
& v_{2}= \\
& v_{3}= \\
& i_{1}= \\
& i_{2}= \\
& \hline
\end{aligned}
$$

(b)


Name: $\qquad$

Section: $\qquad$

Problem 2.3: For each of the circuits below express the resistance $R$, or the equivalent conductance $G=\frac{1}{R}$, at the terminals in terms of the element resistances $R_{n}$ (or conductances $\left.G_{n}=\frac{1}{R_{n}}\right)$. Write your answers next to the circuits.
(a)

(b)

(c)

(d)


## Name:

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$\qquad$
(e) Can the same methods used in Parts (a) through (d) be used to find the resistance at the terminals of the circuit below? If so, express it; if not, explain why not.


Problem 2.4: In each of the following circuits determine the voltages and/or currents indicated. The units are volts (V), milliamperes (mA), and kilo-ohms ( $\mathrm{k} \Omega$ ).
(a)

(b)


Problem 2.5: The circuit show below has two independent sources. All elements are assigned numerical values.


The units are volts (V), milliamperes ( mA ), and millimhos (mmho) ( 1 milliampere $=10^{-3}$ amperes, 1 millimho $=10^{-3}$ mhos. The conductance of a resistor of $R$ ohms is $\frac{1}{R}$ mhos).
The questions which follow illustrate the utility of the superposition principle in linear circuits.

1) Assume that the 2 mA current source $I_{1}$ is active and the voltage source is inactive or dead. Redraw the circuit on your solution pages under these conditions.
2) Analyze this reduced circuit to obtain symbolic expressions for the voltages $e_{1}$ and $e_{2}$ in terms of the conductances. You should be able to do this without solving simultaneous equations by using parallel and series reductions to find $e_{1}$, and then working to the right to find $e_{2}$.
3) Substitute numbers for the symbolic element values in your results for part 2 and determine the values of $e_{1}$ and $e_{2}$ for $I_{1}$ acting alone. Specify units!
4) Repeat parts 1, 2, and 3 for the 1.5 V voltage source active and the other source dead. Use
series and parallel reductions.
5) Verify that the node equations for the complete circuit are:

$$
\begin{array}{rlll}
\left(G_{1}+G_{2}\right) e_{1} & - & G_{2} e_{2} & = \\
I_{1} \\
-G_{2} e_{1} & +\left(G_{2}+G_{3}+G_{4}\right) e_{2} & = & G_{3} V
\end{array}
$$

After turning the algebraic crank, the symbolic forms of the two node equations for the original two-source circuit shown above yield the following solutions:

$$
\begin{aligned}
e_{1} & =\frac{I_{1}\left(G_{2}+G_{3}+G_{4}\right)+V G_{2} G_{3}}{G_{1}\left(G_{2}+G_{3}+G_{4}\right)+G_{2}\left(G_{3}+G_{4}\right)} \\
e_{2} & =\frac{I_{1} G_{2}+V\left(G_{1}+G_{2}\right) G_{3}}{G_{1}\left(G_{2}+G_{3}+G_{4}\right)+G_{2}\left(G_{3}+G_{4}\right)}
\end{aligned}
$$

When the component values are inserted in these equations, the results are:

$$
\begin{aligned}
& e_{1}=1.55 \mathrm{~V} \\
& e_{2}=0.65 \mathrm{~V}
\end{aligned}
$$

Add together the two component parts of $e_{1}$ and $e_{2}$ calculated in parts 3$)$ and 4$)$. That is:

$$
e_{1}=\left.e_{1}\right|_{I_{1} \text { Acting }} ^{\text {Alone }}+\left.e_{1}\right|_{V \underset{\text { Alone }}{V}}
$$

and similiarly for $e_{2}$. If your analysis is correct, the values of the voltages obtained in this manner - that is, by employing the superposition principle - will be the same as those obtained by solving the node equations simultaneously.

The amount of grind required is less when superposition is employed. This will be true whenever the circuit topology is such that series/parallel reductions can be employed

