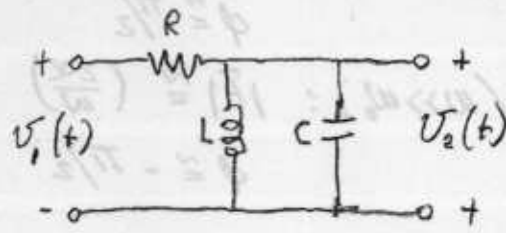


NOTES FOR 6.002 LECTURE #16, APRIL 8, 2008

SINUSOIDAL STEADY STATE IN RLC CIRCUITS: READ: CH 14.



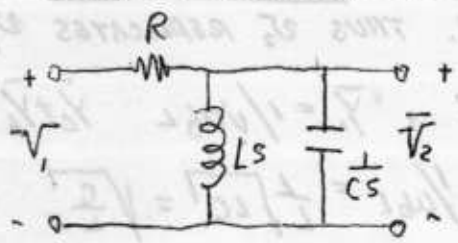
DRIVE THIS NETWORK WITH A SINUSOID

V1(t) = V1 cos ωt

V2(t) IS A SINUSOID IN THE STEADY STATE V2(t) = V2(ω) cos [ωt + φ(ω)]

DETERMINE V2(ω) AND φ(ω).

FOR ANALYSIS, DRIVE THE CIRCUIT WITH V e^{jωt}, EMPLOY ELEMENT IMPEDANCES AND FOCUS ON COMPLEX AMPLITUDES.



FIRST THINK ABOUT WHAT MUST HAPPEN FOR VERY LOW AND VERY HIGH FREQUENCIES.

ANALYSIS: THINK VOLTAGE DIVIDER: $\bar{A}(s) = \frac{\bar{V}_2}{\bar{V}_1} = \frac{Ls \parallel \frac{1}{Cs}}{Ls \parallel \frac{1}{Cs} + R}$

$$\bar{A}(s) = \frac{\frac{Ls(\frac{1}{Cs})}{Ls + 1/Cs}}{\frac{Ls(\frac{1}{Cs})}{Ls + 1/Cs} + R} = \frac{Ls(\frac{1}{Cs})}{Ls(\frac{1}{Cs}) + R Ls + \frac{R}{Cs}} = \frac{Ls}{R L C s^2 + Ls + R}$$

OR: $\bar{A}(s) = \frac{\frac{1}{RC} s}{s^2 + \frac{1}{RC} s + \frac{1}{LC}}$

DEFINE: $2\alpha = 1/RC$
 $\omega_0^2 = 1/LC$

THESE DEFINED PARAMETERS HAVE USEFUL INTERPRETATIONS LATER

$$\bar{A}(s) = \frac{2\alpha s}{s^2 + 2\alpha s + \omega_0^2}$$

OR FOR $s = j\omega$

$$\bar{A}(j\omega) = \frac{2\alpha(j\omega)}{-\omega^2 + 2\alpha(j\omega) + \omega_0^2} = \frac{2\alpha\omega}{\sqrt{(\omega_0^2 - \omega^2)^2 + (2\alpha\omega)^2}} e^{j(\frac{\pi}{2} - \theta)}$$

$\frac{|\bar{V}_2(\omega)|}{V_1}$ WHERE $\theta = \tan^{-1} \left(\frac{2\alpha\omega}{\omega_0^2 - \omega^2} \right)$

FOR VERY LOW FREQUENCIES ($\omega \ll \omega_0$): $|\bar{A}| \approx \left(\frac{2\alpha}{\omega_0^2}\right) \omega$

$\phi \approx \pi/2$

FOR VERY HIGH FREQUENCIES ($\omega \gg \omega_0$): $|\bar{A}| \approx \left(\frac{2\alpha}{\omega}\right)$

$\phi \approx -\pi/2$

INTUITIVE
CONFIRMED

INTERESTING BEHAVIOR NEAR ω_0 :

FOR $\omega = \omega_0$ $|\bar{A}| = 1$ $\phi = 0$ $\underline{v_2(t) = v_1(t)}$

ω_0 IS THE RESONANT FREQUENCY. AT THIS FREQUENCY, THE

PARALLEL LC COMBINATION HAS: ZERO ADMITTANCE AND THERE IS NO VOLTAGE DROP IN R. THUS v_2 REPLICATES v_1 .

AT RESONANCE $\bar{Y}_C = j\omega_0 C$, $\bar{Y}_L = 1/j\omega_0 L$ $\bar{Y}_C + \bar{Y}_L = j(\omega_0 C - 1/\omega_0 L)$

$\omega_0 C = C \sqrt{\frac{1}{LC}} = \sqrt{\frac{C}{L}}$ $1/\omega_0 L = \frac{1}{L} \sqrt{LC} = \sqrt{\frac{C}{L}}$ $\bar{Y}_C + \bar{Y}_L \equiv 0$

THERE IS CURRENT IN C AND L NONETHELESS. EVALUATE THE CURRENTS DOWN IN L, C:

$\bar{I}_C = \bar{V}_2(j\omega_0 C) = j \sqrt{\frac{C}{L}} \bar{V}_2$ THIS IS A CIRCULATING CURRENT IN L & C

$\bar{I}_L = \bar{V}_2(1/j\omega_0 L) = -j \sqrt{\frac{C}{L}} \bar{V}_2$

NOTE THAT \bar{V}_2 AND \bar{I}_L ARE ORTHOGONAL ($\frac{\pi}{2}$ OUT OF PHASE)

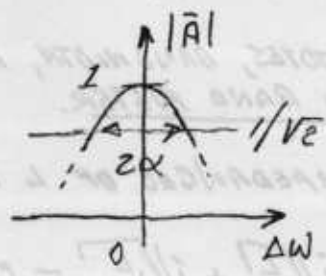
THUS WHEN $v_2(t)$ IS A MAXIMUM, $i_L(t)$ IS ZERO AND VICE-VERSA

AT RESONANCE STORED ENERGY IS FLOPPING BACK AND FORTH BETWEEN L AND C.

BEHAVIOR NEAR ω_0 LET $\omega = \omega_0 + \Delta\omega$ $\Delta\omega \ll \omega_0$

$|\bar{A}(j\omega)| = \frac{2\alpha(\omega_0 + \Delta\omega)}{\sqrt{[(\omega_0 + \omega_0 + \Delta\omega)(\omega_0 - \omega_0 - \Delta\omega)]^2 + [2\alpha(\omega_0 + \Delta\omega)]^2}}$
 $\approx \frac{2\alpha\omega_0}{\sqrt{(2\omega_0)^2(-\Delta\omega)^2 + 2\alpha\omega_0^2}} = \frac{2\alpha}{\sqrt{(-2\Delta\omega)^2 + (2\alpha)^2}}$

WHEN $\Delta\omega = 0$ $|\bar{A}| = 1$ WHEN $\Delta\omega = \pm\alpha$ $|\bar{A}| = 1/\sqrt{2}$
 NEAR RESONANCE, $|\bar{A}|$ LOOKS LIKE



$\omega_0 \pm \alpha$ ARE REFERRED TO AS
HALF-POWER FREQUENCIES
 BECAUSE $|V_2|^2 = \frac{1}{2} |V_2|_{MAX}^2$

2α IS THE BANDWIDTH OF THE
 RESONANT PEAK.

CONSTRUCT THE BODE PLOT:

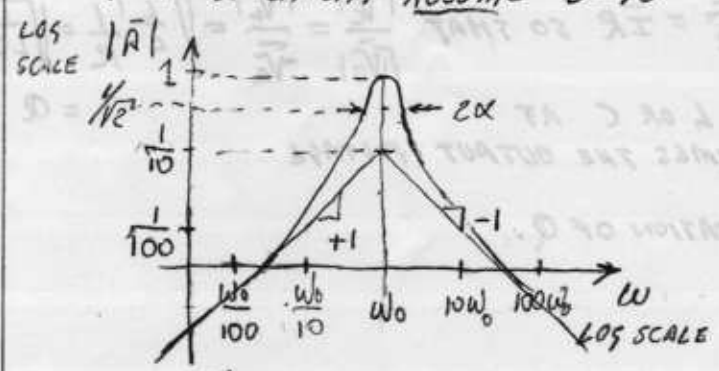
$Q = \frac{\omega_0}{2\alpha}$ ANOTHER INTERPRETATION
 OF Q

WE HAVE SYMPTOTES ON \log SCALE

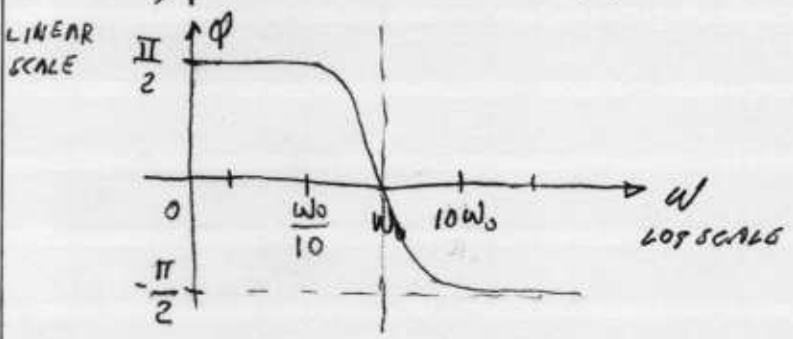
FOR L.F. $|\bar{A}| = \frac{2\alpha}{\omega_0^2} \omega$ (WHICH EQUALS $\frac{2\alpha}{\omega_0} = \frac{1}{Q}$ AT $\omega = \omega_0$), $\phi = \pi/2$

FOR H.F. $|\bar{A}| = \frac{2\alpha}{\omega}$ (WHICH EQUALS $\frac{2\alpha}{\omega_0} = 1/Q$ AT $\omega = \omega_0$) $\phi = -\pi/2$

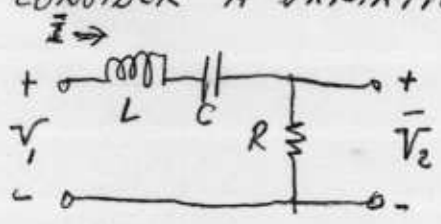
FOR SPECIFICITY ASSUME $Q = 10$



A NARROW-BAND FILTER



CONSIDER A VARIATION:



AGAIN A VOLTAGE DIVIDER:

$$\bar{A}(s) = \frac{R}{R + Ls + \frac{1}{Cs}} = \frac{RCs}{RCs + LCs^2 + 1}$$

THINK ABOUT LF, HF ASYMPTOTES

$$\bar{A}(s) = \frac{2\alpha s}{s^2 + 2\alpha s + \omega_0^2} \quad \left. \begin{array}{l} 2\alpha = \frac{R}{L} \\ \omega_0^2 = 1/LC \end{array} \right\} \text{NOTE DIFFERENT FORM OF } \alpha \text{ FOR SERIES RLC.}$$

INTERPRETATION INCLUDING ASYMPTOTES, BANDWIDTH, BODE PLOT IDENTICAL TO PARALLEL RLC. A NARROW BAND FILTER.

AT RESONANCE ($\omega = \omega_0$) THE IMPEDANCES OF L AND C ADD TO ZERO

$$Z_C + Z_L = \frac{1}{j\omega_0 C} + j\omega_0 L = -j\sqrt{\frac{L}{C}} + j\sqrt{\frac{L}{C}} \equiv 0 \quad \text{AND } \bar{V}_2 = V$$

SO THAT $v_2(t) = v_1(t)$

THERE ARE EQUAL AND OPPOSITE VOLTAGES ACROSS L AND C

$$\left. \begin{array}{l} \bar{V}_C = \frac{1}{j\omega_0 C}(\bar{I}) = -j\sqrt{\frac{L}{C}}\bar{I} \\ \bar{V}_L = j\omega_0 L(\bar{I}) = +j\sqrt{\frac{L}{C}}\bar{I} \end{array} \right\} \bar{V}_L + \bar{V}_C \equiv 0 \text{ AT } \omega_0$$

NOTE THAT $\bar{V}_R = \bar{V}_2 = \bar{I}R$ SO THAT $\frac{|\bar{V}_C|}{|\bar{V}_R|} = \frac{|\bar{V}_L|}{|\bar{V}_R|} = \sqrt{\frac{L}{C}} \frac{1}{R} = \sqrt{\frac{1}{LC}} \left(\frac{L}{R}\right) = Q$

THE VOLTAGE ACROSS L OR C AT RESONANCE IS Q TIMES THE OUTPUT VOLTAGE

ANOTHER INTERPRETATION OF Q.

