# 6.003: Signals and Systems

**Discrete Approximation of Continuous-Time Systems** 

# Mid-term Examination #1

Wednesday, October 5, 7:30-9:30pm, 26-310, 26-322, 26-328.

No recitations on the day of the exam.

Coverage: CT and DT Systems, Z and Laplace Transforms
Lectures 1–7

Recitations 1–7

Homeworks 1–4

Homework 4 will not collected or graded. Solutions will be posted. Closed book: 1 page of notes  $(8\frac{1}{2} \times 11 \text{ inches}; \text{ front and back})$ .

No calculators, computers, cell phones, music players, or other aids.

Designed as 1-hour exam; two hours to complete.

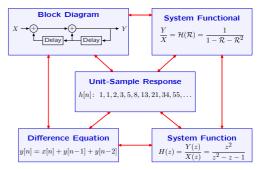
Review sessions during open office hours.

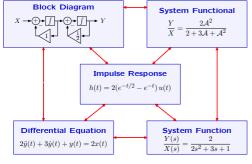
Conflict? Contact freeman@mit.edu before Friday, Sept. 30, 5pm.

Prior term midterm exams have been posted on the 6.003 website.

#### **Concept Map**

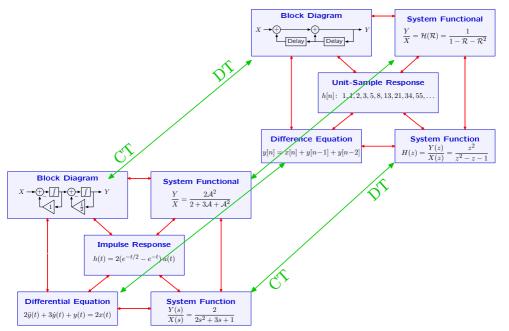
Today we will look at relations between CT and DT representations.



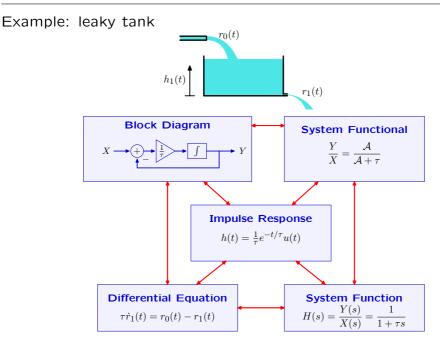


#### **Concept Map**

Today we will look at relations between CT and DT representations.

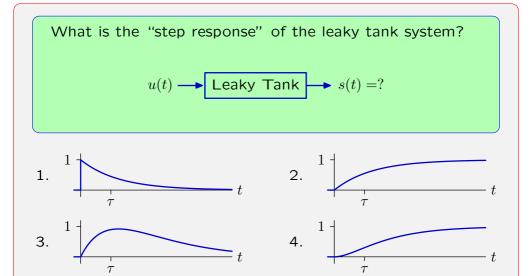


## **Discrete Approximation of CT Systems**



Today: compare step responses of leaky tank and DT approximation.

# Check Yourself (Practice for Exam)



5. none of the above

What is the "step response" of the leaky tank system?

de: 
$$\tau \dot{r}_1(t) = u(t) - r_1(t)$$

$$t < 0$$
:  $r_1(t) = 0$ 

$$t > 0$$
:  $r_1(t) = c_1 + c_2 e^{-t/\tau}$ 

$$\dot{r}_1(t) = -\frac{c_2}{\tau}e^{-t/\tau}$$

Substitute into de: 
$$\tau\left(-\frac{c_2}{\tau}\right)e^{-t/\tau} = 1 - c_1 - c_2e^{-t/\tau} \rightarrow c_1 = 1$$

Combine t < 0 and t > 0:

$$r_1(t) = u(t) + c_2 e^{-t/\tau} u(t)$$

$$\dot{r}_1(t) = \delta(t) + c_2 \delta(t) - \frac{c_2}{\tau} e^{-t/\tau} u(t)$$

Substitute into de:

$$\tau(1+c_2)\delta(t) - \tau \frac{c_2}{\tau} e^{-t/\tau} u(t) = u(t) - u(t) - c_2 e^{-t/\tau} u(t) \quad \to \quad c_2 = -1$$
$$r_1(t) = (1 - e^{-t/\tau}) u(t)$$

Alternatively, reason with systems!

$$\delta(t) \longrightarrow \boxed{\frac{A}{A+\tau}} \longrightarrow h(t) = \frac{1}{\tau}e^{-t/\tau}u(t)$$

$$u(t) \longrightarrow \boxed{\frac{A}{A+\tau}} \longrightarrow s(t) = ?$$

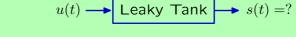
$$\delta(t) \longrightarrow \boxed{\frac{A}{A+\tau}} \longrightarrow s(t) = ?$$

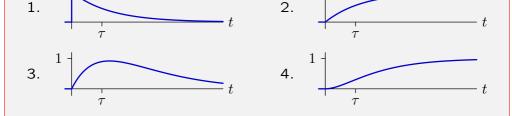
$$\delta(t) \longrightarrow \boxed{\frac{A}{A+\tau}} \longrightarrow h(t)$$

$$\delta(t) \longrightarrow \boxed{\frac{A}{A+\tau}} \longrightarrow s(t) = \int_{-\infty}^{t} h(t')dt'$$

$$s(t) = \int_{-\infty}^{t} \frac{1}{\tau} e^{-t'/\tau} u(t') dt' = \int_{0}^{t} \frac{1}{\tau} e^{-t'/\tau} dt' = (1 - e^{-t/\tau}) u(t)$$

What is the "step response" of the leaky tank system? 2





5. none of the above

# Forward Euler Approximation

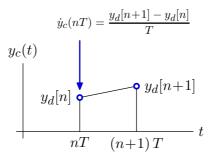
Approximate leaky-tank system using **forward** Euler approach.

Approximate continuous signals by discrete signals:

$$x_d[n] = x_c(nT)$$

$$y_d[n] = y_c(nT)$$

Approximate derivative at t = nT by looking **forward** in time:



# Forward Euler Approximation

Approximate leaky-tank system using forward Euler approach.

Substitute

$$x_d[n] = x_c(nT)$$

$$y_d[n] = y_c(nT)$$

$$\dot{y}_c(nT)\approx\frac{y_c\big((n+1)T\big)-y_c\big(nT\big)}{T}=\frac{y_d[n+1]-y_d[n]}{T}$$
 into the differential equation

$$\tau \dot{y}_c(t) = x_c(t) - y_c(t)$$

to obtain

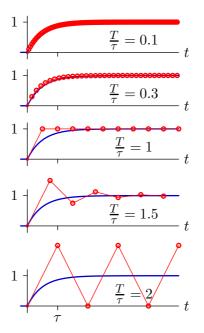
$$\frac{\tau}{T}\Big(y_d[n+1] - y_d[n]\Big) = x_d[n] - y_d[n].$$

Solve:

$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

# Forward Euler Approximation

Plot.



Why is this approximation badly behaved for large  $\frac{T}{\tau}$ ?

#### DT approximation:

$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

Find the DT pole.

1. 
$$z = \frac{T}{\tau}$$
 2.  $z = 1 - \frac{T}{\tau}$  3.  $z = \frac{\tau}{T}$  4.  $z = -\frac{\tau}{T}$ 

$$= \frac{\tau}{T} \qquad \qquad 4. \ \ z = -\frac{\tau}{T}$$

5. 
$$z = \frac{1}{1 + \frac{T}{\tau}}$$

DT approximation:

$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

Take the Z transform:

$$zY_d(z) - \left(1 - \frac{T}{\tau}\right)Y_d(z) = \frac{T}{\tau}X_d(z)$$

Solve for the system function:

$$H(z) = \frac{Y_d(z)}{X_d(z)} = \frac{\frac{1}{\tau}}{z - \left(1 - \frac{T}{\tau}\right)}$$

Pole at 
$$z = 1 - \frac{T}{\tau}$$
.

#### DT approximation:

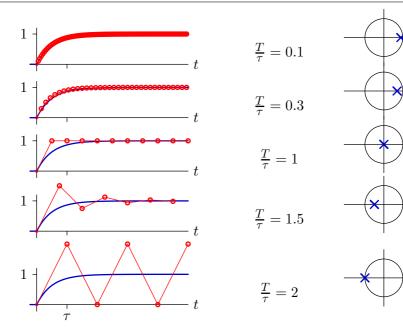
$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

#### Find the DT pole. 2

1. 
$$z = \frac{T}{\tau}$$
 2.  $z = 1 - \frac{T}{\tau}$ 

3. 
$$z = \frac{\tau}{T}$$
 4.  $z = -\frac{\tau}{T}$ 

5. 
$$z = \frac{1}{1 + \frac{T}{\tau}}$$



The CT pole was fixed  $(s=-\frac{1}{\tau})$ . Why is the DT pole changing?

Dependence of DT pole on  ${\it T}$  is generic property of forward Euler.

Dependence of DT pole on T is generic property of forward Euler.

Approach: make a systems model of forward Euler method.

CT block diagrams: adders, gains, and integrators:

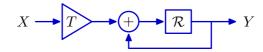
$$X \longrightarrow A \longrightarrow Y$$

Forward Euler approximation:

$$\frac{y[n+1] - y[n]}{T} = x[n]$$

Equivalent system:

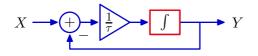
 $\dot{y}(t) = x(t)$ 



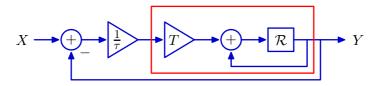
Forward Euler: substitute equivalent system for all integrators.

## Example: leaky tank system

Started with leaky tank system:



Replace integrator with forward Euler rule:



Write system functional:

$$\frac{Y}{X} = \frac{\frac{T}{\tau} \frac{\mathcal{R}}{1 - \mathcal{R}}}{1 + \frac{T}{\tau} \frac{\mathcal{R}}{1 - \mathcal{R}}} = \frac{\frac{T}{\tau} \mathcal{R}}{1 - \mathcal{R} + \frac{T}{\tau} \mathcal{R}} = \frac{\frac{T}{\tau} \mathcal{R}}{1 - \left(1 - \frac{T}{\tau}\right) \mathcal{R}}$$

Equivalent to system we previously developed:

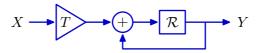
$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

#### Model of Forward Euler Method

Replace every integrator in the CT system

$$X \longrightarrow A \longrightarrow Y$$

with the forward Euler model:



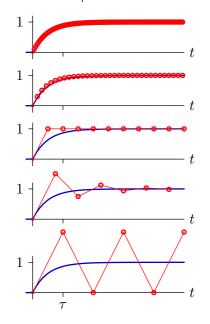
Substitute the DT operator for A:

$$\mathcal{A} = \frac{1}{s} \to \frac{T\mathcal{R}}{1 - \mathcal{R}} = \frac{\frac{T}{z}}{1 - \frac{1}{z}} = \frac{T}{z - 1}$$

Forward Euler maps  $s \to \frac{z-1}{T}$ .

Or equivalently: z = 1 + sT.

Pole at  $z=1-\frac{T}{\tau}=1+sT$ .



$$\frac{T}{\tau} = 0.1$$



$$\frac{T}{\tau} = 1$$

$$\frac{T}{\tau} = 1.5$$









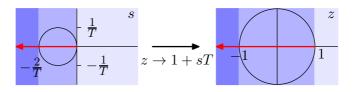




# Forward Euler: Mapping CT poles to DT poles

Forward Euler Map:

$$\begin{array}{ccc} s & \rightarrow & z = 1 + sT \\ 0 & 1 \\ -\frac{1}{T} & 0 \\ -\frac{2}{T} & -1 \end{array}$$



DT stability: CT pole must be inside circle of radius  $\frac{1}{T}$  at  $s=-\frac{1}{T}.$ 

$$-\frac{2}{T} < -\frac{1}{\tau} < 0 \quad \rightarrow \quad \frac{T}{\tau} < 2$$

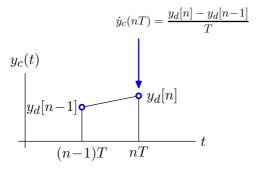
## **Backward Euler Approximation**

We can do a similar analysis of the **backward** Euler method.

Approximate continuous signals by discrete signals:

$$x_d[n] = x_c(nT)$$
$$y_d[n] = y_c(nT)$$

Approximate derivative at t = nT by looking **backward** in time:



# **Backward Euler Approximation**

We can do a similar analysis of the backward Euler method.

Substitute

$$\begin{split} x_d[n] &= x_c(nT) \\ y_d[n] &= y_c(nT) \\ \dot{y}_c(nT) &\approx \frac{y_c\big(nT\big) - y_c\big((n-1)T\big)}{T} = \frac{y_d[n] - y_d[n-1]}{T} \end{split}$$

into the differential equation

$$\tau \dot{y}_c(t) = x_c(t) - y_c(t)$$

to obtain

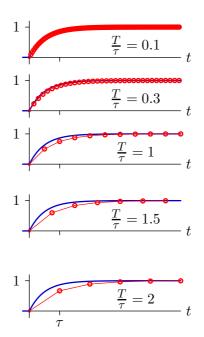
$$\frac{\tau}{T}\Big(y_d[n] - y_d[n-1]\Big) = x_d[n] - y_d[n].$$

Solve:

$$\left(1 + \frac{T}{\tau}\right) y_d[n] - y_d[n-1] = \frac{T}{\tau} x_d[n]$$

# **Backward Euler Approximation**

Plot.



This approximation is better behaved. Why?

DT approximation:

$$\left(1 + \frac{T}{\tau}\right) y_d[n] - y_d[n-1] = \frac{T}{\tau} x_d[n]$$

Find the DT pole.

1. 
$$z = \frac{T}{\tau}$$
 2.  $z = 1 - \frac{T}{\tau}$  3.  $z = \frac{\tau}{T}$  4.  $z = -\frac{\tau}{T}$ 

$$= \frac{1}{T}$$
 4.  $z = -\frac{1}{T}$ 

5. 
$$z = \frac{1}{1 + \frac{T}{\tau}}$$

DT approximation:

$$\left(1 + \frac{T}{\tau}\right) y_d[n] - y_d[n-1] = \frac{T}{\tau} x_d[n]$$

Take the Z transform:

$$\left(1 + \frac{T}{\tau}\right)Y_d(z) - z^{-1}Y_d(z) = \frac{T}{\tau}X_d(z)$$

Find the system function:

$$H(z) = \frac{Y_d(z)}{X_d(z)} = \frac{\frac{T}{\tau}z}{\left(1 + \frac{T}{\tau}\right)z - 1}$$

Pole at 
$$z = \frac{1}{1 + \frac{T}{2}}$$
.

#### DT approximation:

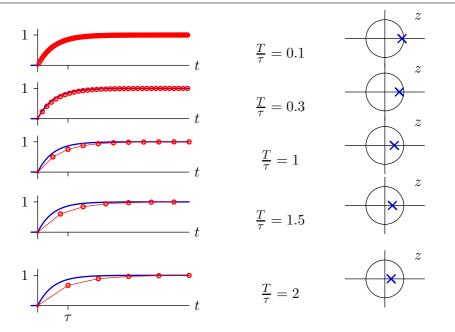
$$y_d[n+1] - \left(1 - \frac{T}{\tau}\right) y_d[n] = \frac{T}{\tau} x_d[n]$$

## Find the DT pole. 5

1. 
$$z = \frac{T}{\tau}$$
 2.  $z = 1 - \frac{T}{\tau}$ 

3. 
$$z = \frac{\tau}{T}$$
 4.  $z = -\frac{\tau}{T}$ 

5. 
$$z = \frac{1}{1 + T}$$



Why is this approximation better behaved?

Make a systems model of backward Euler method.

CT block diagrams: adders, gains, and integrators:

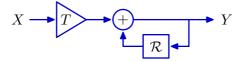
$$X \longrightarrow A \longrightarrow Y$$

$$\dot{y}(t) = x(t)$$

Backward Euler approximation:

$$\frac{y[n] - y[n-1]}{T} = x[n]$$

Equivalent system:



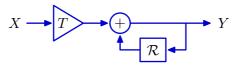
Backward Euler: substitute equivalent system for all integrators.

#### Model of Backward Euler Method

Replace every integrator in the CT system

$$X \longrightarrow A \longrightarrow Y$$

with the backward Euler model:

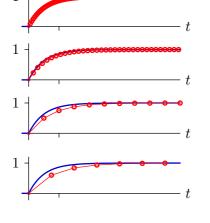


Substitute the DT operator for A:

$$\mathcal{A} = \frac{1}{s} \to \frac{T}{1 - \mathcal{R}} = \frac{T}{1 - \frac{1}{z}}$$

Backward Euler maps  $z \to \frac{1}{1-sT}$ .

Pole at 
$$z = \frac{1}{1 + \frac{T}{\tau}} = \frac{1}{1 - sT}$$
.



$$\frac{T}{\tau} = 0.1$$





$$\frac{T}{\tau} = 1.5$$











# Backward Euler: Mapping CT poles to DT poles

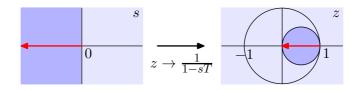
Backward Euler Map:

$$s \rightarrow z = \frac{1}{1-sT}$$

$$0 \qquad 1$$

$$-\frac{1}{T} \qquad \frac{1}{2}$$

$$-\frac{2}{T} \qquad \frac{1}{3}$$

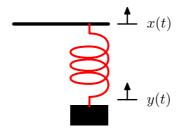


The entire left half-plane maps inside a circle with radius  $\frac{1}{2}$  at  $z=\frac{1}{2}$ . If CT system is stable, then DT system is also stable.

## Masses and Springs, Forwards and Backwards

In Homework 2, you investigated three numerical approximations to a mass and spring system:

- forward Euler
- backward Euler
- centered method



## **Trapezoidal Rule**

The trapezoidal rule uses centered differences.

Approximate CT signals at points between samples:

$$y_c((n-\frac{1}{2})T) = \frac{y_d[n] + y_d[n-1]}{2}$$

Approximate derivatives at points between samples:

$$\dot{y}_c \left( (n - \frac{1}{2})T \right) = \frac{y_d[n] - y_d[n-1]}{T}$$

$$y_{c}\left(\left(n-\frac{1}{2}\right)T\right) = \frac{y_{d}[n] + y_{d}[n-1]}{2}$$

$$\dot{y}_{c}\left(\left(n-\frac{1}{2}\right)T\right) = \frac{y_{d}[n] - y_{d}[n-1]}{T}$$

$$y_{c}(t)$$

$$y_{d}[n-1] \qquad y_{d}[n]$$

$$(n-1)T \qquad nT$$

## **Trapezoidal Rule**

The trapezoidal rule uses centered differences.

$$\dot{y}(t) = x(t)$$

Trapezoidal rule:

$$\frac{y[n] - y[n-1]}{T} = \frac{x[n] + x[n-1]}{2}$$

Z transform:

$$H(z) = \frac{Y(s)}{X(s)} = \frac{T}{2} \left( \frac{1+z^{-1}}{1-z^{-1}} \right) = \frac{T}{2} \left( \frac{z+1}{z-1} \right)$$

Map:

$$\mathcal{A} = \frac{1}{s} \to \frac{T}{2} \left( \frac{z+1}{z-1} \right)$$

Trapezoidal rule maps 
$$z \to \frac{1 + \frac{sI}{2}}{1 - \frac{sT}{2}}$$
.

# Trapezoidal Rule: Mapping CT poles to DT poles

Trapezoidal Map:

$$s \rightarrow z = \frac{1 + \frac{sT}{2}}{1 - \frac{sT}{2}}$$

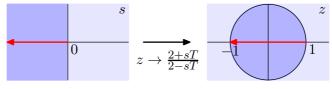
$$0 \qquad 1$$

$$-\frac{1}{T} \qquad \frac{1}{3}$$

$$-\frac{2}{T} \qquad 0$$

$$-\infty \qquad -1$$

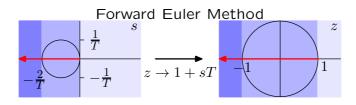
$$j\omega \qquad \frac{2 + j\omega T}{2 - j\omega T}$$

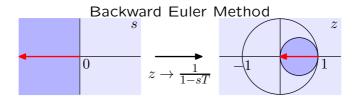


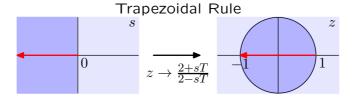
The entire left-half plane maps inside the unit circle.

The  $j\omega$  axis maps onto the unit circle

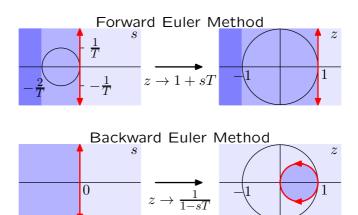
# Mapping s to z: Leaky-Tank System

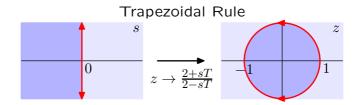




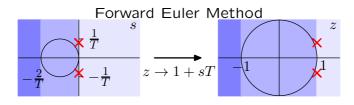


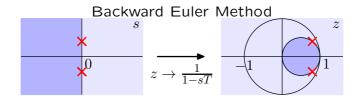
# Mapping s to z: Mass and Spring System

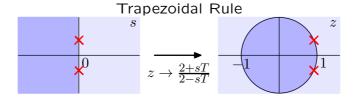




# Mapping s to z: Mass and Spring System







#### **Concept Map**

Relations between CT and DT representations.

