

Today: Continuous Time
Fourier Transform (CTFT)

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CTFT is just the Laplace transform
on the $j\omega$ -axis.

$$x(t) \xleftrightarrow{\mathcal{L}} X(s) \qquad x(t) \xleftrightarrow{\mathcal{F}} X(j\omega)$$

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt.$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt.$$

So what's the big deal?

1) FT has a (useful) inverse transform.

2) We'll define the FT for some signals
where the integral does not converge. (is not finite).

Inverse transforms.

Recall inverse Laplace.

Eq: is an integral over complex domain.
we never used it.

Instead we did template matching.

$$x(t) = e^{-at} u(t) \rightarrow X(s) = \frac{1}{s+a}, \operatorname{Re}\{s\} > -a$$

$$X_2(s) = \frac{A}{s+p_1} + \frac{B}{s+p_2} + \frac{C}{s+p_3} + \dots$$

$$\Rightarrow x_2(t) = _ e^{-p_1 t} u(t) + _ e^{-p_2 t} u(t) + \dots$$

Inverse FT is easy!

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega.$$

Note the similarity to

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt.$$

flip ω -axis.

scale by 2π .

"Duality": relationships for time \rightarrow freq
also hold for freq \rightarrow time.

Inverse FT demotivates FT for signals which don't converge.

$$x(t) = 1.$$

$$X(s) = ? = \int_{-\infty}^{\infty} e^{-st} dt.$$

doesn't converge.

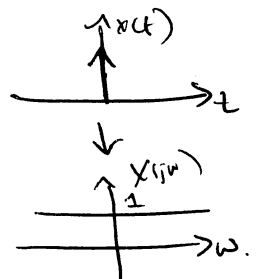
$$X(j\omega) = \int_{-\infty}^{\infty} e^{-j\omega t} dt = \begin{cases} \infty & \omega=0 \\ 0 & \omega \neq 0 \end{cases}$$

So doesn't converge,

but might be able to use $\delta(\omega)$.

Let's try $x(t) = \delta(t)$

$$X(j\omega) = \int_{-\infty}^{\infty} \delta(t) e^{-j\omega t} dt = e^{-j\omega \cdot 0} = 1 \quad \forall \omega.$$



By duality, expect δ in freq $\Rightarrow 1$ in time.

$$X(j\omega) = \delta(\omega).$$

$$x(t) = \frac{1}{2\pi} \int \delta(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} e^{j0t} = \frac{1}{2\pi}.$$

$$\text{So } x(t) = 1 \iff X(j\omega) = 2\pi \delta(\omega).$$

Try another:

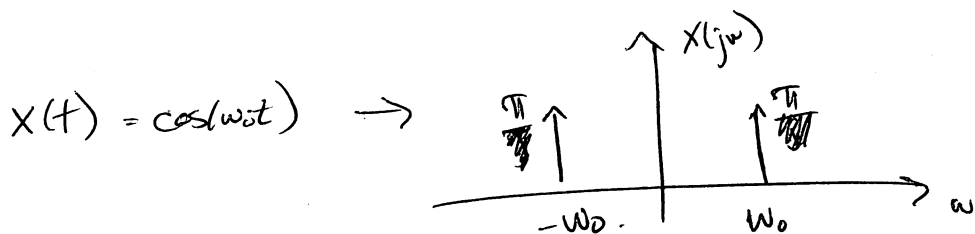
$$x(t) = \cos(\omega_0 t).$$

$$= \frac{1}{2} (e^{j\omega_0 t} + e^{-j\omega_0 t})$$

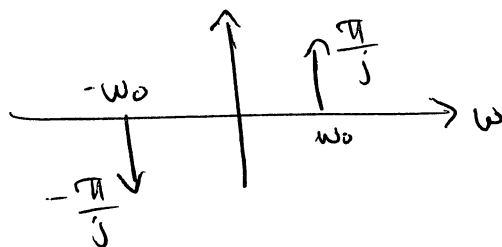
$$\begin{aligned} X(j\omega) &= \frac{1}{2} \int_{-\infty}^{\infty} e^{j\omega_0 t} e^{-j\omega t} dt + \frac{1}{2} \int_{-\infty}^{\infty} e^{-j\omega_0 t} e^{-j\omega t} dt \\ &= \frac{1}{2} \int_{-\infty}^{\infty} e^{j(\omega_0 - \omega)t} dt + \frac{1}{2} \int_{-\infty}^{\infty} e^{-j(\omega_0 + \omega)t} dt \end{aligned}$$

by argument above

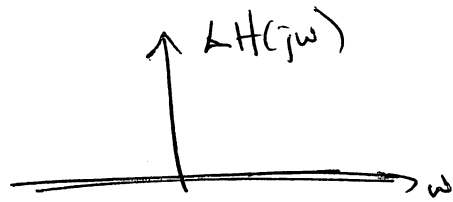
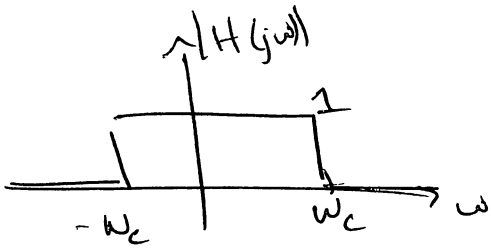
$$= \frac{1}{2} [2\pi \delta(\omega_0 - \omega) + 2\pi \delta(\omega_0 + \omega)]$$



What about $x(t) = \sin(\omega_0 t)$



Ideal LPF



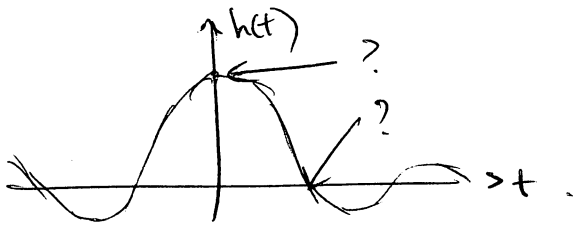
what happens if you push cos / sin through?

if $\omega_0 < \omega_c$, come through unaffected.

if $\omega_0 > \omega_c$, zero out.

What is $h(t)$ for ideal LPF?

Know from duality?



Hard to get from a physical system!

$$h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(j\omega) e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega t} d\omega$$

$$= \frac{1}{2\pi} \left[\frac{1}{j t} e^{j\omega t} \right]_{-\omega_c}^{\omega_c}$$

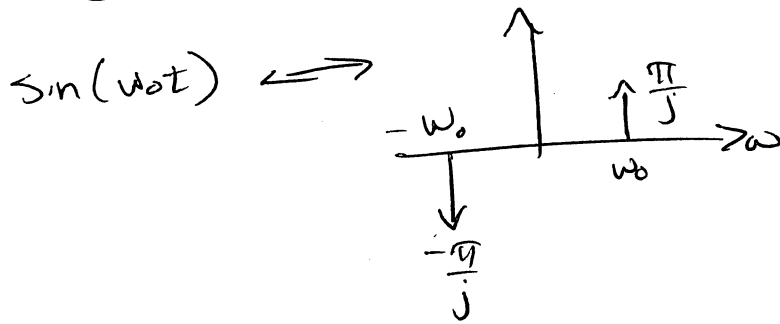
$$= \frac{1}{2\pi j t} \left[e^{j\omega_c t} - e^{-j\omega_c t} \right] = \frac{\sin(\omega_c t)}{\pi t}$$

$h(t) = 0$ for $t = k \frac{\pi}{\omega_c}$ Integer k , $k \neq 0$.

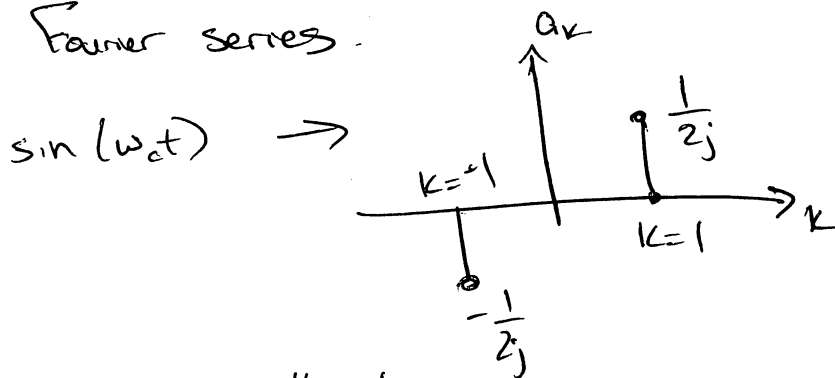
$$h(0) = \frac{\omega_c}{\pi} \quad \left[\sin(\omega_c t) \approx \omega_c t \text{ at zero} \right] \Rightarrow \frac{\sin(\omega_c t)}{\pi t} \approx \frac{\omega_c t}{\pi t} = \frac{\omega_c}{\pi}$$

or $\frac{1}{2\pi}$ - area in $H(j\omega)$.

Something very cool happened. Did you notice.



Recall Fourier series.



This is generally true.

Relationship to Fourier Series.

$$X(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\omega_0 t}$$

$$X(j\omega) = \sum_{k=-\infty}^{\infty} a_k \int_{-\infty}^{\infty} e^{jk\omega_0 t} e^{-j\omega t} dt$$

$$= \sum_{k=-\infty}^{\infty} a_k \int_{-\infty}^{\infty} e^{j(k\omega_0 - \omega)t} dt$$

$$= \sum_{k=-\infty}^{\infty} a_k 2\pi \delta(k\omega_0 - \omega)$$

