# 6.003: Signals and Systems

**CT Frequency Response and Bode Plots** 

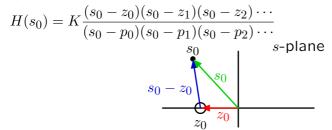
March 9, 2010

# Last Time

Complex exponentials are eigenfunctions of LTI systems.

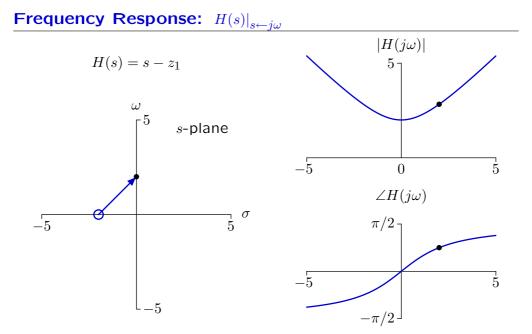
$$e^{s_0t} \longrightarrow H(s_0) e^{s_0t}$$

 $H(s_0)$  can be determined graphically using vectorial analysis.

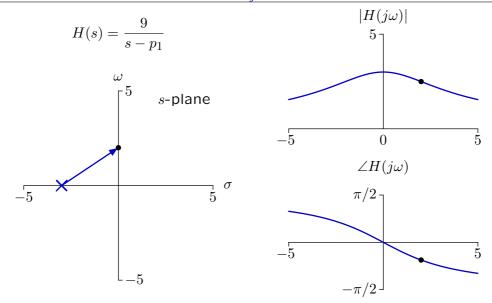


Response of an LTI system to an eternal cosine is an eternal cosine: same frequency, but scaled and shifted.

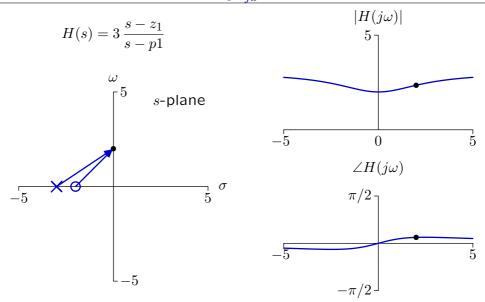
$$\cos(\omega_0 t) \longrightarrow H(s) \longrightarrow |H(j\omega_0)| \cos\left(\omega_0 t + \angle H(j\omega_0)\right)$$



# Frequency Response: $H(s)|_{s \leftarrow j\omega}$



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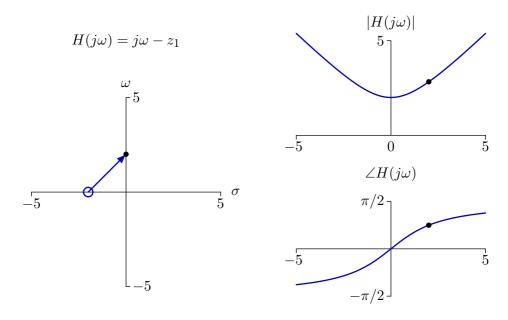


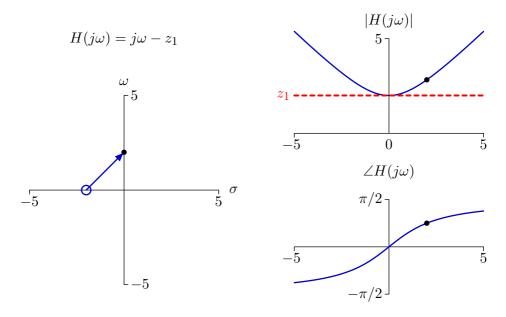
## **Poles and Zeros**

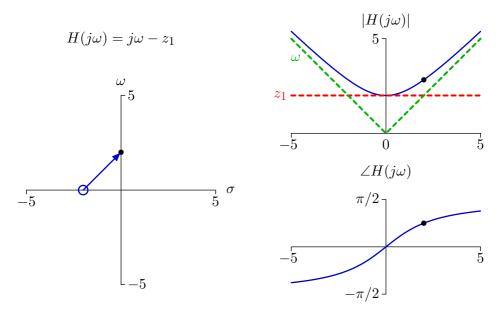
Thinking about systems as collections of poles and zeros is an important design concept.

- simple: just a few numbers characterize entire system
- powerful: complete information about frequency response

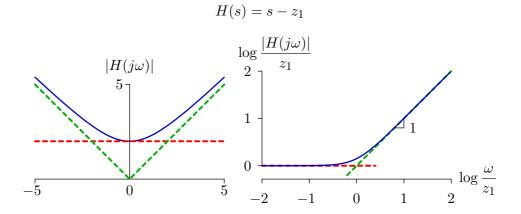
Today: poles, zeros, frequency responses, and Bode plots.





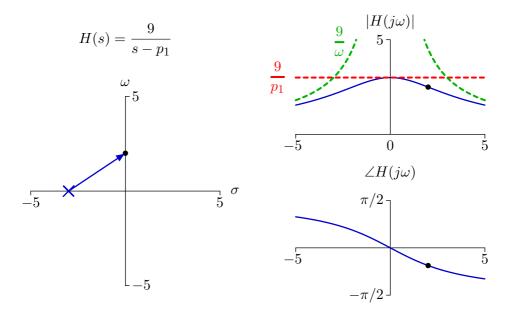


Two asymptotes provide a good approxmation on log-log axes.

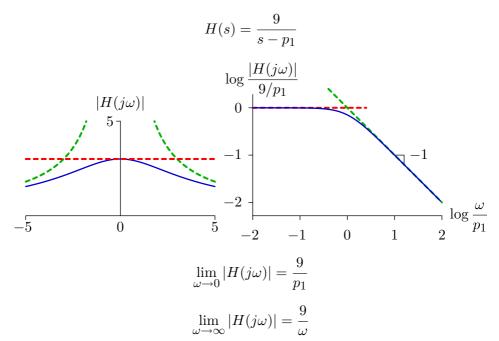


 $\lim_{\omega \to 0} |H(j\omega)| = z_1$ 

 $\lim_{\omega\to\infty}|H(j\omega)|=\omega$ 



Two asymptotes provide a good approxmation on log-log axes.



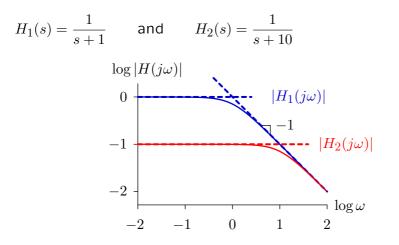
Compare log-log plots of the frequency-response magnitudes of the following system functions:

$$H_1(s) = \frac{1}{s+1}$$
 and  $H_2(s) = \frac{1}{s+10}$ 

The former can be transformed into the latter by

- 1. shifting horizontally
- 2. shifting and scaling horizontally
- 3. shifting both horizontally and vertically
- 4. shifting and scaling both horizontally and vertically
- 5. none of the above

Compare log-log plots of the frequency-response magnitudes of the following system functions:



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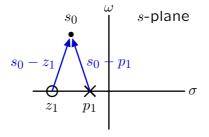
- 1. shifting horizontally
- 2. shifting and scaling horizontally
- 3. shifting both horizontally and vertically
- 4. shifting and scaling both horizontally and vertically
- 5. none of the above

no scaling in either vertical or horizontal directions!

# Asymptotic Behavior of More Complicated Systems

Constructing  $H(s_0)$ .

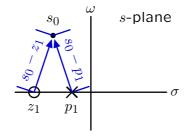
$$H(s_0) = K \quad \frac{\prod_{q=1}^Q (s_0 - z_q)}{\prod_{p=1}^P (s_0 - p_p)} \quad \leftarrow \text{ product of vectors for zeros}$$



### Asymptotic Behavior of More Complicated Systems

The magnitude of a product is the product of the magnitudes.

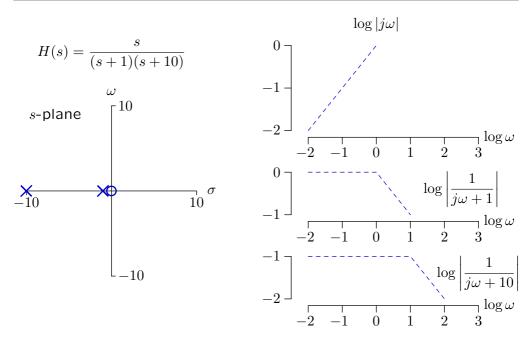
$$|H(s_0)| = \begin{vmatrix} K & \prod_{\substack{q=1\\P}}^{Q} (s_0 - z_q) \\ \prod_{p=1}^{P} (s_0 - p_p) \end{vmatrix} = |K| & \prod_{\substack{q=1\\P}}^{Q} |s_0 - z_q| \\ \prod_{p=1}^{P} |s_0 - p_p| \end{vmatrix}$$

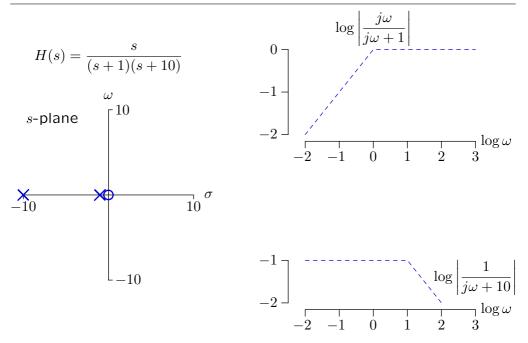


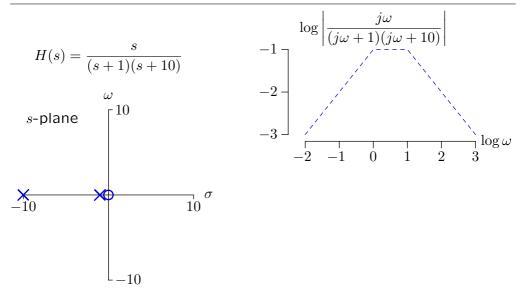
The log of the magnitude is a sum of logs.

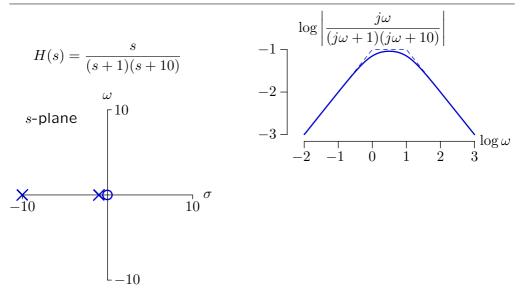
$$|H(s_0)| = \begin{vmatrix} K & \prod_{\substack{q=1\\P}}^{Q} (s_0 - z_q) \\ \prod_{p=1}^{P} (s_0 - p_p) \end{vmatrix} = |K| & \prod_{\substack{q=1\\P}}^{Q} |s_0 - z_q| \\ \prod_{p=1}^{P} |s_0 - p_p| \end{vmatrix}$$

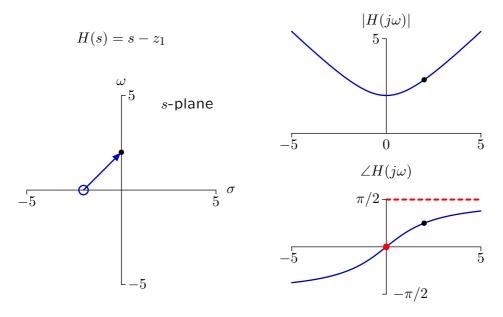
$$\log|H(j\omega)| = \log|K| + \sum_{q=1}^{Q} \log|j\omega - z_q| - \sum_{p=1}^{P} \log|j\omega - p_p|$$



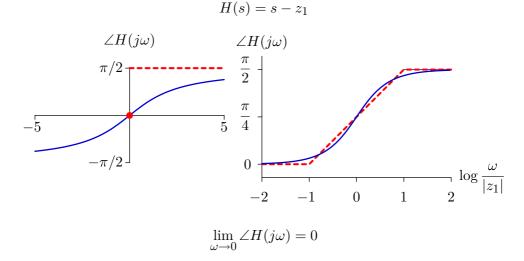




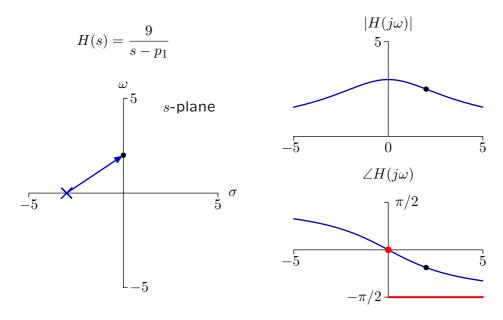




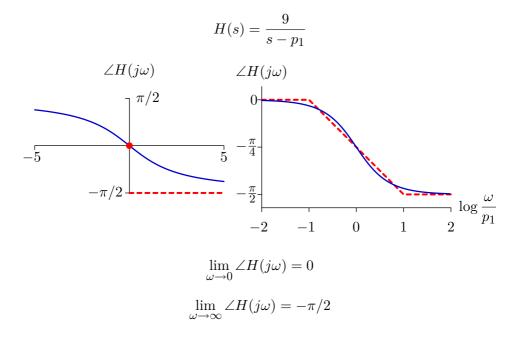
Three straight lines provide a good approximation versus log  $\omega$ .



 $\lim_{\omega\to\infty} \angle H(j\omega) = \pi/2$ 

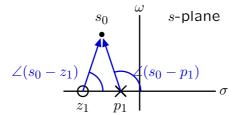


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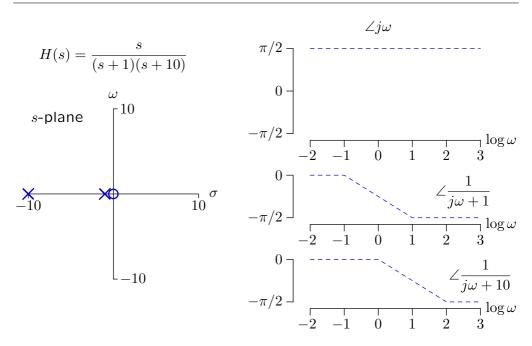


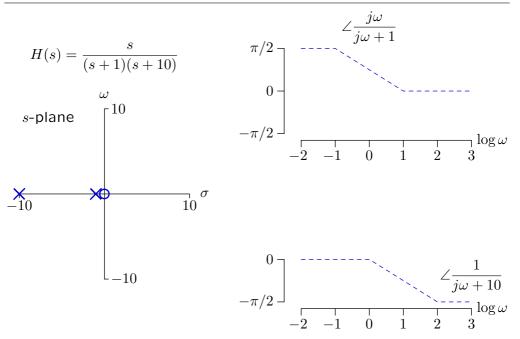
The angle of a product is the sum of the angles.

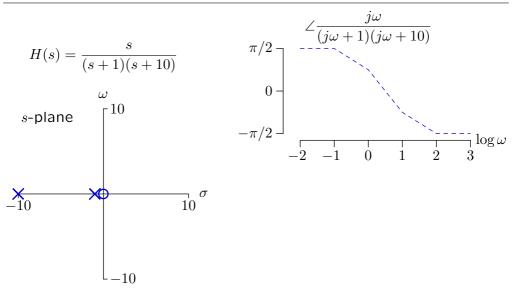
$$\angle H(s_0) = \angle \left( K \frac{\prod_{q=1}^{Q} (s_0 - z_q)}{\prod_{p=1}^{P} (s_0 - p_p)} \right) = \angle K + \sum_{q=1}^{Q} \angle (s_0 - z_q) - \sum_{p=1}^{P} \angle (s_0 - p_p)$$

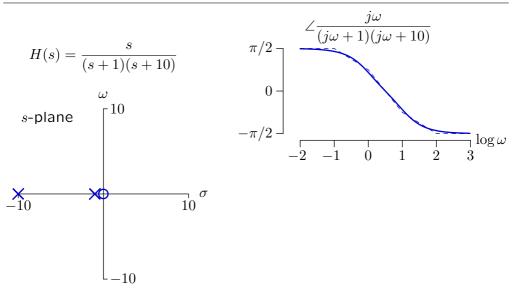


The angle of K can be 0 or  $\pi$  for systems described by linear differential equations with constant, real-valued coefficients.









### From Frequency Response to Bode Plot

The magnitude of  $H(j\omega)$  is a product of magnitudes.

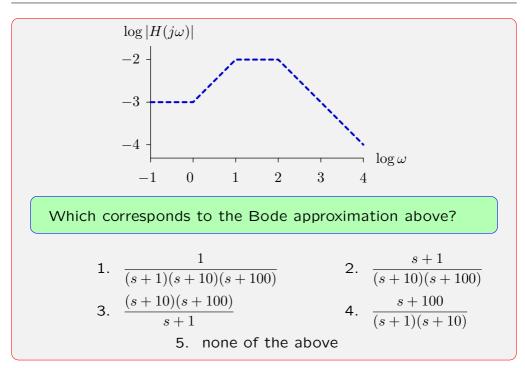
$$|H(j\omega)| = |K| \frac{\prod_{q=1}^{Q} |j\omega - z_q|}{\prod_{p=1}^{P} |j\omega - p_p|}$$

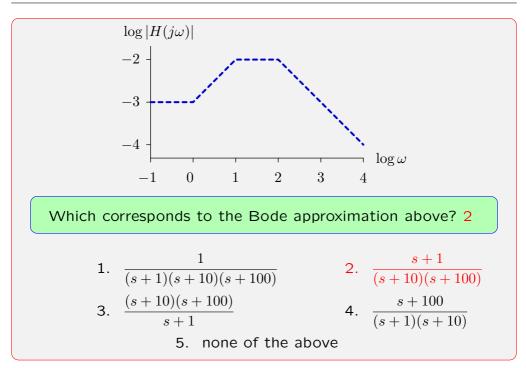
The log of the magnitude is a sum of logs.

$$\log|H(j\omega)| = \log|K| + \sum_{q=1}^{Q} \log|j\omega - z_q| - \sum_{p=1}^{P} \log|j\omega - p_p|$$

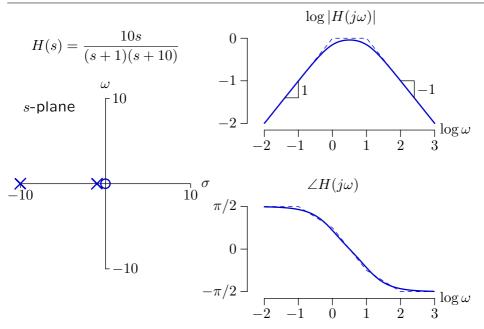
The angle of  $H(j\omega)$  is a sum of angles.

$$\angle H(j\omega) = \angle K + \sum_{q=1}^{Q} \angle (j\omega - z_q) - \sum_{p=1}^{P} \angle (j\omega - p_p)$$

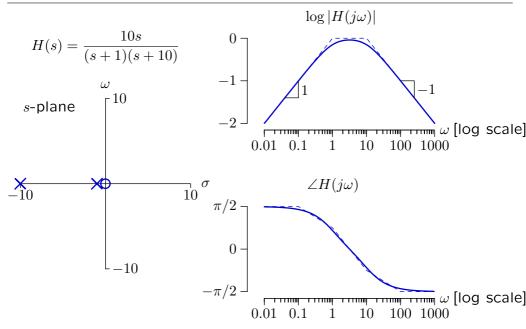




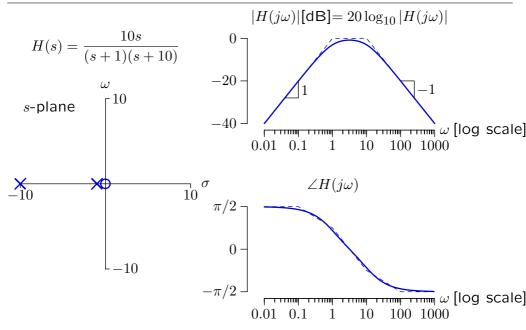
# Bode Plot: dB



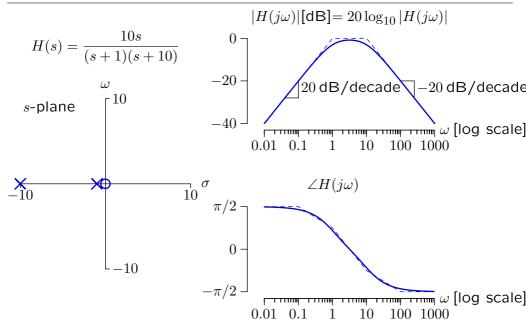
## Bode Plot: dB



### Bode Plot: dB

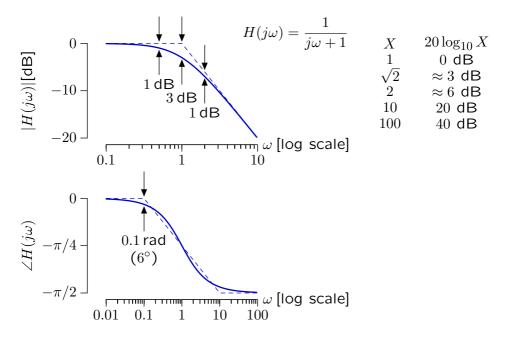


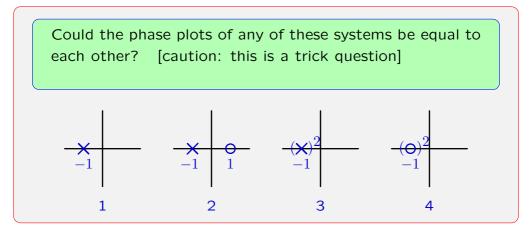
### Bode Plot: dB

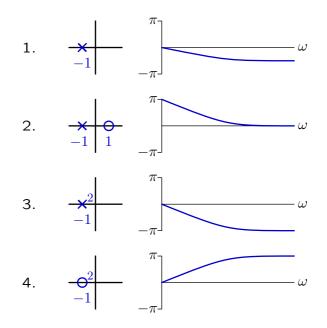


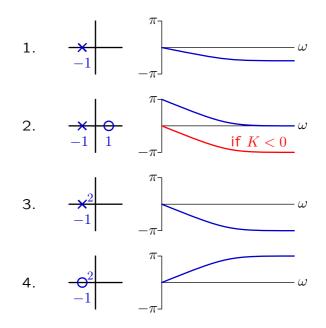
### **Bode Plot: Accuracy**

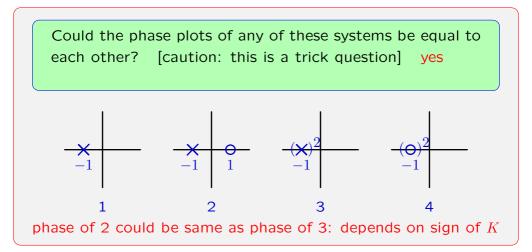
The straight-line approximations are surprisingly accurate.











$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

$$\frac{s}{\omega_0} \text{ plane} \int \sqrt{1 - \left(\frac{1}{2Q}\right)^2} \int \sqrt{1 - \left(\frac{1}{2}\right)^2} \int \sqrt{1 - \left(\frac{1}{2}\right)^2}$$

$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

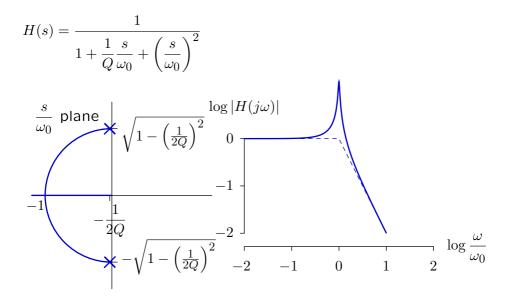
$$\frac{s}{\omega_0} \text{ plane} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} 0 \frac{\log|H(j\omega)|}{-1 - \frac{1}{2Q}} - \frac{1}{2} - \frac{1}{2}$$

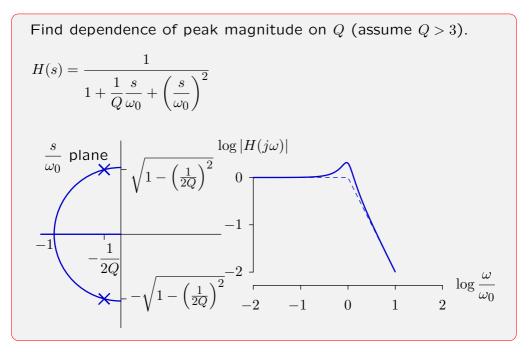
$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

$$\stackrel{s}{\xrightarrow{\omega_0}} \text{plane} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} 0 \frac{\log|H(j\omega)|}{1 - \left(\frac{1}{2Q}\right)^2} 0 \frac{1}{\sqrt{1 - \left(\frac{1}{2Q}\right)^2}} \frac{\log|H(j\omega)|}{1 - \left(\frac{1}{2Q}\right)^2} \frac{1}{\sqrt{1 - \left(\frac{1}{2Q}\right)^2}} \frac{\log|W(j\omega)|}{\sqrt{1 - \left(\frac{1}{2Q}\right)^2} - 2} \frac{\log \frac{\omega}{\omega_0}}{1 - 1} \frac{\log \frac{\omega}{\omega_0}}{1 - 2} \log \frac{\omega}{\omega_0}}$$

$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

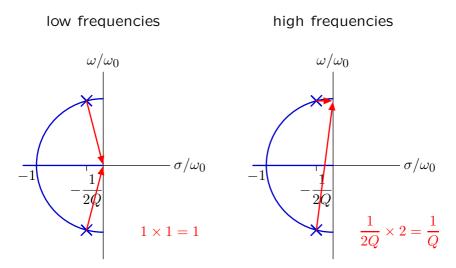
$$\frac{s}{\omega_0} \text{ plane} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} 0 \frac{1}{\sqrt{1 - \left(\frac{1}{2Q}\right)^2}} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} 0 \frac{1}{\sqrt{1 - \left(\frac{1}{2Q}\right)^2}} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} \sqrt{1 - \left(\frac{1}{2Q}\right)$$



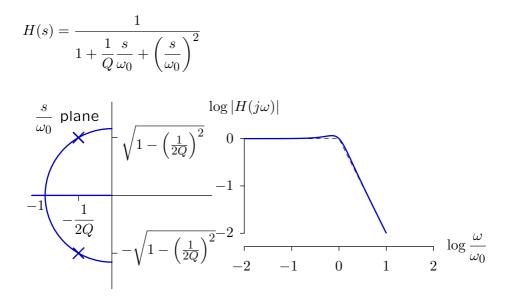


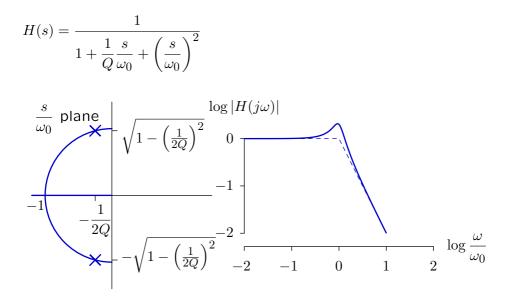
Find dependence of peak magnitude on Q (assume Q > 3).

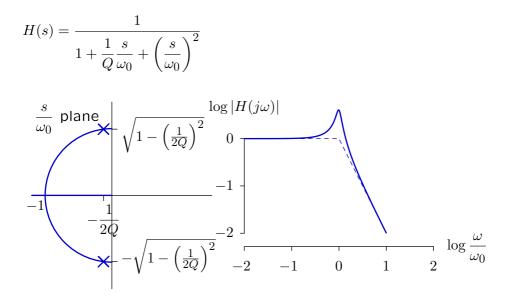
Analyze with vectors.

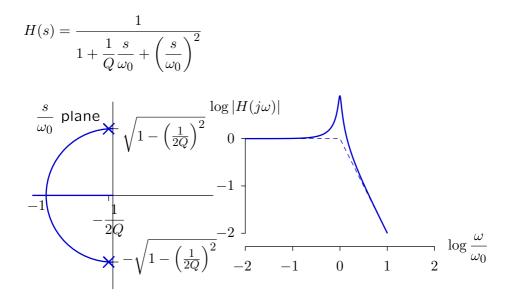


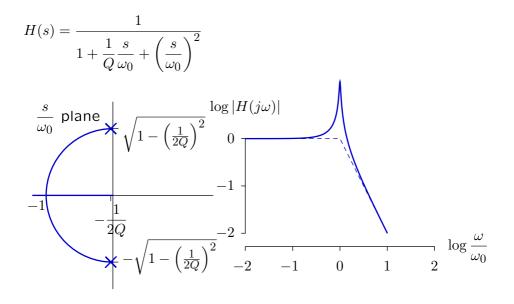
Peak magnitude increases with Q!





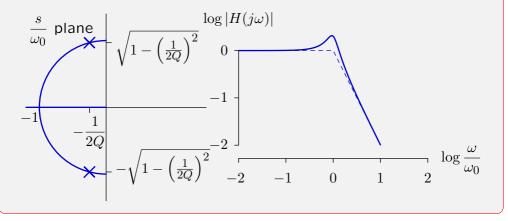






Estimate the "3dB bandwidth" of the peak (assume Q > 3).

Let  $\omega_l$  (or  $\omega_h$ ) represent the lowest (or highest) frequency for which the magnitude is greater than the peak value divided by  $\sqrt{2}$ . The 3dB bandwidth is then  $\omega_h - \omega_l$ .

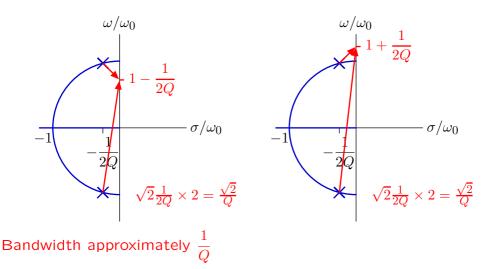


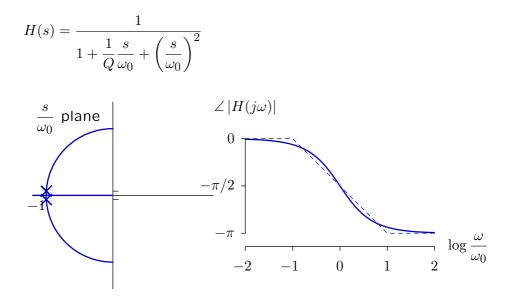
Estimate the "3dB bandwidth" of the peak (assume Q > 3).

Analyze with vectors.

low frequencies

high frequencies





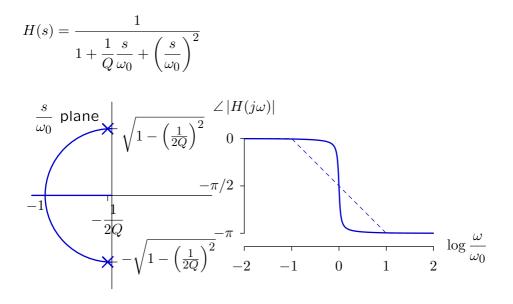
$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

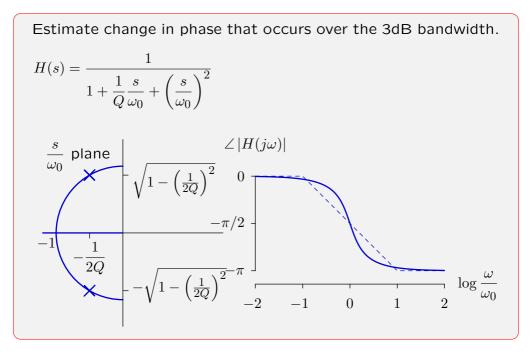
$$\frac{s}{\omega_0} \text{ plane} \int \sqrt{1 - \left(\frac{1}{2Q}\right)^2} 0 \int \sqrt{1 - \left(\frac{1}{2Q}\right)^2} \int \sqrt{1 - \left(\frac{1}{2Q}\right)^2}$$

$$H(s) = \frac{1}{1 + \frac{1}{Q}\frac{s}{\omega_0} + \left(\frac{s}{\omega_0}\right)^2}$$

$$\stackrel{s}{\xrightarrow{\omega_0}} \text{ plane} \sqrt{1 - \left(\frac{1}{2Q}\right)^2} \xrightarrow{\omega_0} |H(j\omega)|$$

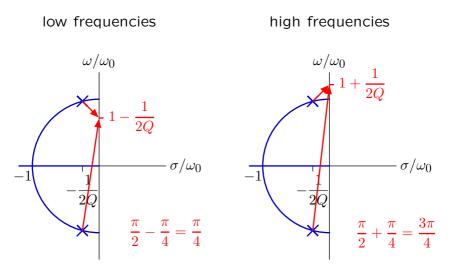
$$-\pi/2 - \frac{1}{2Q} - \pi/2 - \frac{1}{2Q} - \pi/2 - \frac{1}{2Q} - \pi/2 - \frac{1}{2Q$$





Estimate change in phase that occurs over the 3dB bandwidth.

Analyze with vectors.



Change in phase approximately  $\frac{\pi}{2}$ .

# Summary

The frequency response of a system can be quickly determined using Bode plots.

Bode plots are constructed from sections that correspond to single poles and single zeros.

Responses for each section simply sum when plotted on logarithmic coordinates.