

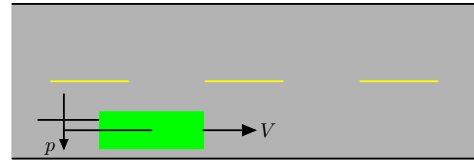
### 6.003: Signals and Systems

#### Feedback and Control

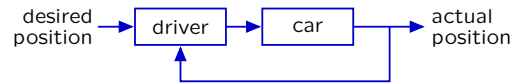
March 11, 2010

#### Feedback and Control

Feedback is pervasive in natural and artificial systems.

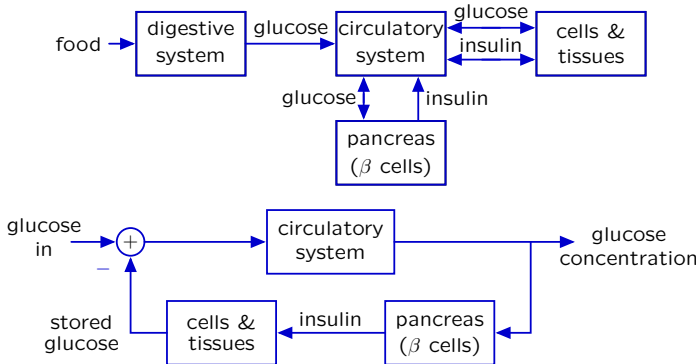


Turn steering wheel to stay centered in the lane.



#### Feedback and Control

Concentration of glucose in blood is highly regulated and remains nearly constant despite episodic ingestion and use.

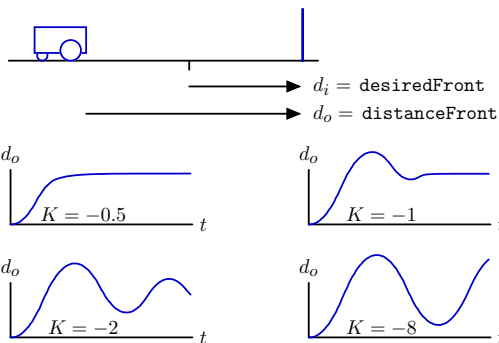


#### Today's goal

Use systems theory to gain insight into how to control a system.

#### Example: wallFinder System

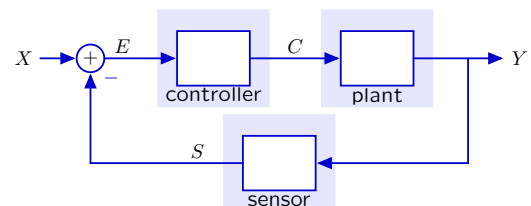
Approach a wall, stopping a desired distance  $d_i$  in front of it.



What causes these different types of responses?

#### Structure of a Control Problem

(Simple) Control systems have three parts.



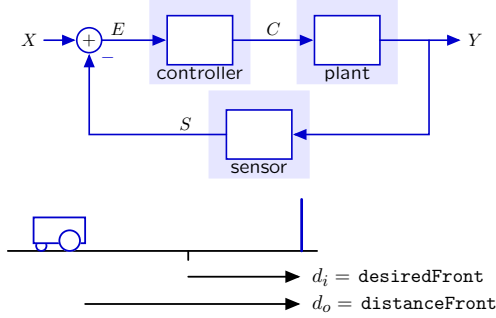
The **plant** is the system to be controlled.

The **sensor** measures the output of the plant.

The **controller** specifies a command  $C$  to the plant based on the **difference** between the input  $X$  and sensor output  $S$ .

**Analysis of wallFinder System**

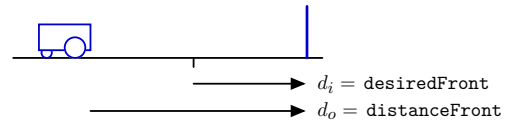
Cast wallFinder problem into control structure.



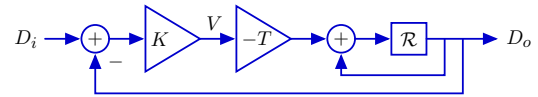
proportional controller:  $v[n] = Ke[n] = K(d_i[n] - d_s[n])$   
 locomotion:  $d_o[n] = d_o[n - 1] - Tv[n - 1]$   
 sensor with no delay:  $d_s[n] = d_o[n]$

**Analysis of wallFinder System: Block Diagram**

Visualize as block diagram.

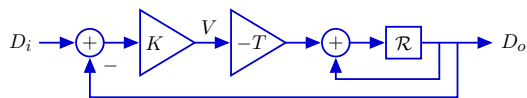
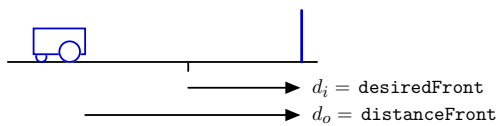


proportional controller:  $v[n] = Ke[n] = K(d_i[n] - d_s[n])$   
 locomotion:  $d_o[n] = d_o[n - 1] - Tv[n - 1]$   
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**Analysis of wallFinder System: System Function**

Solve.



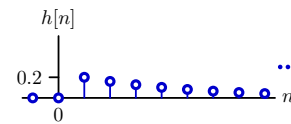
$$\frac{D_o}{D_i} = \frac{-KTR}{1 - R} = \frac{-KTR}{1 - R - KTR} = \frac{-KTR}{1 - (1 + KT)R}$$

**Analysis of wallFinder System: Poles**

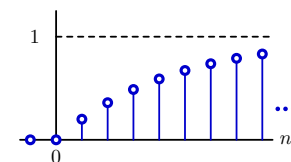
The system function contains a single **pole** at  $z = 1 + KT$ .

$$\frac{D_o}{D_i} = \frac{-KTR}{1 - (1 + KT)R}$$

Unit-sample response for  $KT = -0.2$ :



Unit-step response  $s[n]$  for  $KT = -0.2$ :



What determines the speed of the response? Could it be faster?

**Check Yourself**

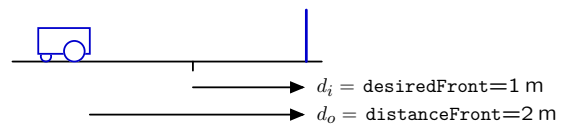
Find  $KT$  for fastest convergence of unit-sample response.

$$\frac{D_o}{D_i} = \frac{-KTR}{1 - (1 + KT)R}$$

1.  $KT = -2$
2.  $KT = -1$
3.  $KT = 0$
4.  $KT = 1$
5.  $KT = 2$
0. none of the above

**Analysis of wallFinder System**

The optimum gain  $K$  moves robot to desired position in **one** step.



$$KT = -1$$

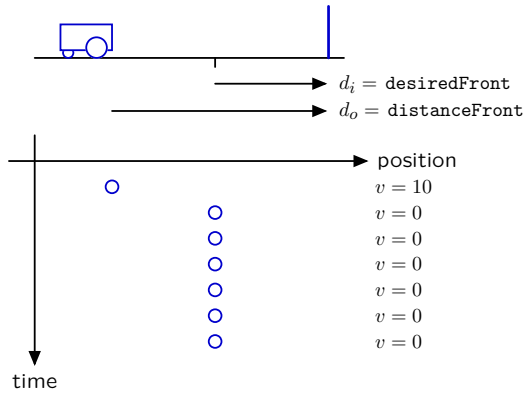
$$K = -\frac{1}{T} = -\frac{1}{1/10} = -10$$

$$v[n] = K(d_i[n] - d_o[n]) = -10(1 - 2) = 10 \text{ m/s}$$

exactly the right speed to get there in one step!

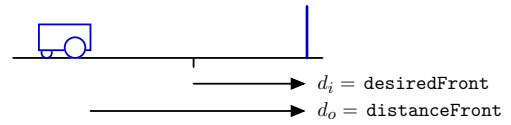
**Analyzing wallFinder: Space-Time Diagram**

The optimum gain  $K$  moves robot to desired position in **one** step.



**Analysis of wallFinder System: Adding Sensor Delay**

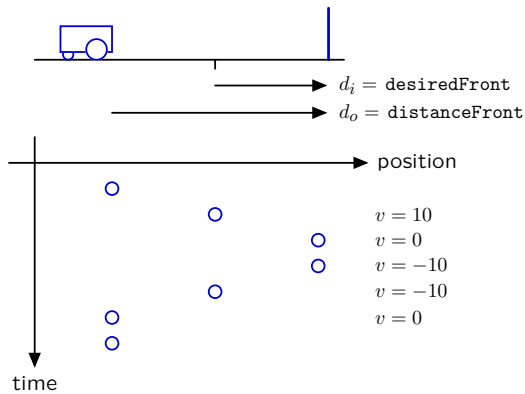
Adding delay tends to destabilize control systems.



proportional controller:  $v[n] = Ke[n] = K(d_i[n] - d_s[n])$   
 locomotion:  $d_o[n] = d_o[n-1] - Tv[n-1]$   
 sensor **with delay**:  $d_s[n] = d_o[n-1]$

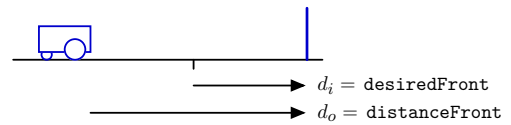
**Analysis of wallFinder System: Adding Sensor Delay**

Adding delay tends to destabilize control systems.

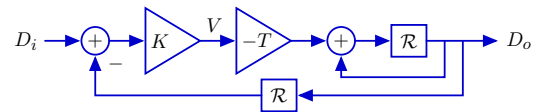


**Analysis of wallFinder System: Block Diagram**

Incorporating sensor delay in block diagram.

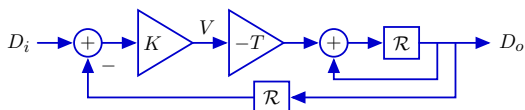


proportional controller:  $v[n] = Ke[n] = K(d_i[n] - d_s[n])$   
 locomotion:  $d_o[n] = d_o[n-1] - Tv[n-1]$   
 sensor with no delay:  $d_s[n] = d_o[n-1]$



**Check Yourself**

Find the system function  $H = \frac{D_o}{D_i}$ .



1.  $\frac{KTR}{1-R}$
2.  $\frac{-KTR}{1+R-KTR^2}$
3.  $\frac{KTR}{1-R} - KTR$
4.  $\frac{-KTR}{1-R-KTR^2}$
5. none of the above

**Analyzing wallFinder: Poles**

Substitute  $\mathcal{R} \rightarrow \frac{1}{z}$  in the system functional to find the poles.

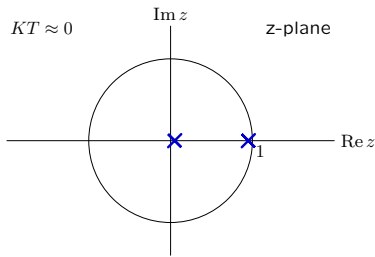
The poles are then the roots of the denominator.

$$z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT}$$

**Feedback and Control: Poles**

If  $KT$  is small, the poles are at  $z \approx -KT$  and  $z \approx 1 + KT$ .

$$z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT} \approx \frac{1}{2} \pm \sqrt{\left(\frac{1}{2} + KT\right)^2} = 1 + KT, -KT$$

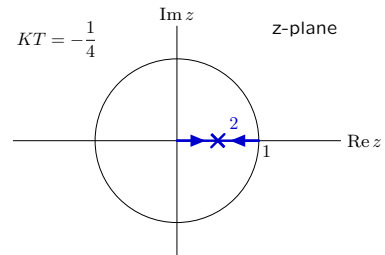


Pole near 0 generates fast response.  
 Pole near 1 generates slow response.  
 Slow mode (pole near 1) dominates the response.

**Feedback and Control: Poles**

As  $KT$  becomes more negative, the poles move toward each other and collide at  $z = \frac{1}{2}$  when  $KT = -\frac{1}{4}$ .

$$z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT} = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 - \frac{1}{4}} = \frac{1}{2}, \frac{1}{2}$$

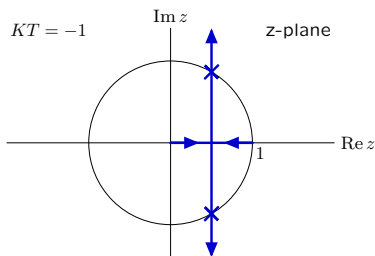


Persistent responses decay. The system is stable.

**Feedback and Control: Poles**

If  $KT < -1/4$ , the poles are complex.

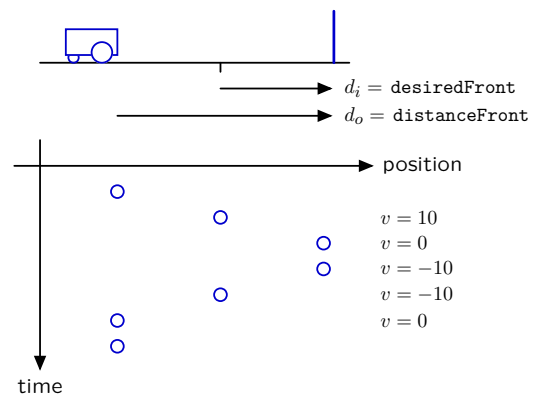
$$z = \frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT} = \frac{1}{2} \pm j\sqrt{-KT - \left(\frac{1}{2}\right)^2}$$



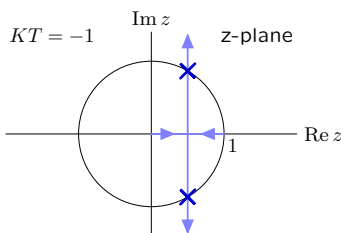
Complex poles → oscillations.

**Same oscillation we saw earlier!**

Adding delay tends to destabilize control systems.



**Check Yourself**

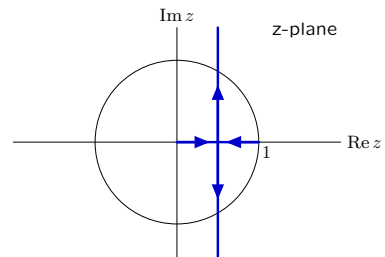


What is the period of the oscillation?

- 1. 1    2. 2    3. 3
- 4. 4    5. 6    0. none of above

**Feedback and Control: Poles**

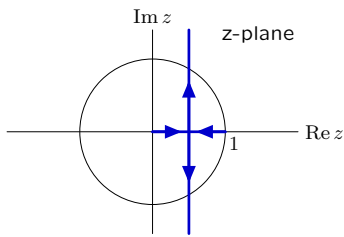
The closed loop poles depend on the gain.



If  $KT : 0 \rightarrow -\infty$ : then  $z_1, z_2 : 0, 1 \rightarrow \frac{1}{2}, \frac{1}{2} \rightarrow \frac{1}{2} \pm j\infty$

**Check Yourself**

Find  $KT$  for fastest response.



closed-loop poles

$$\frac{1}{2} \pm \sqrt{\left(\frac{1}{2}\right)^2 + KT}$$

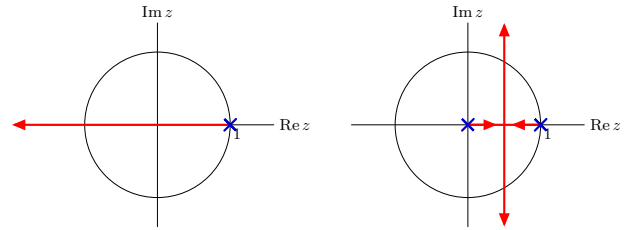
- 1. 0
- 2.  $-\frac{1}{4}$
- 3.  $-\frac{1}{2}$
- 4. -1
- 5.  $-\infty$
- 0. none of above

**Destabilizing Effect of Delay**

Adding delay in the feedback loop makes it more difficult to stabilize.

Ideal sensor:  $d_s[n] = d_o[n]$

More realistic sensor (with delay):  $d_s[n] = d_o[n - 1]$



Fastest response without delay: single pole at  $z = 0$ .

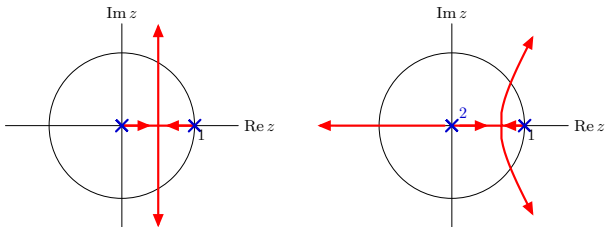
Fastest response with delay: double pole at  $z = \frac{1}{2}$ . **much slower!**

**Destabilizing Effect of Delay**

Adding more delay in the feedback loop is even worse.

More realistic sensor (with delay):  $d_s[n] = d_o[n - 1]$

Even more delay:  $d_s[n] = d_o[n - 2]$



Fastest response with delay: double pole at  $z = \frac{1}{2}$ .

Fastest response with more delay: double pole at  $z = 0.682$ .

→ **even slower**

**Feedback and Control: Summary**

Feedback is an elegant way to design a control system.

Stability of a feedback system is determined by its dominant pole.

Delays tend to decrease the stability of a feedback system.