6.003: Signals and Systems

Fourier Representations

March 30, 2010

Mid-term Examination #2

Wednesday, April 7, 7:30-9:30pm, 34-101.

No recitations on the day of the exam.

Coverage: Lectures 1–15 Recitations 1–15 Homeworks 1–8

Homework 8 will not collected or graded. Solutions will be posted.

Closed book: 2 pages of notes $(8\frac{1}{2} \times 11 \text{ inches}; \text{ front and back}).$

Designed as 1-hour exam; two hours to complete.

Review sessions during open office hours.

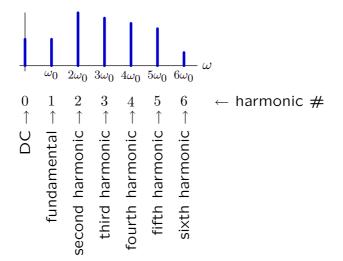
Conflict? Contact freeman@mit.edu before Friday, April 2, 5pm.

Fourier Representations

Fourier series represent signals in terms of sinusoids.

 \rightarrow leads to a new representation for systems as filters.

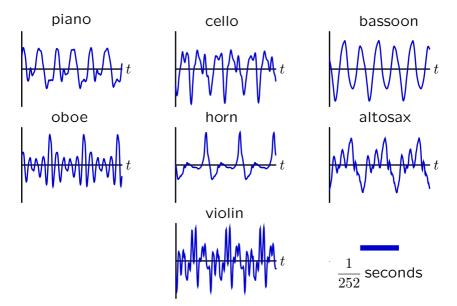
Representing signals by their harmonic components.



Musical Instruments

Harmonic content is natural way to describe some kinds of signals.

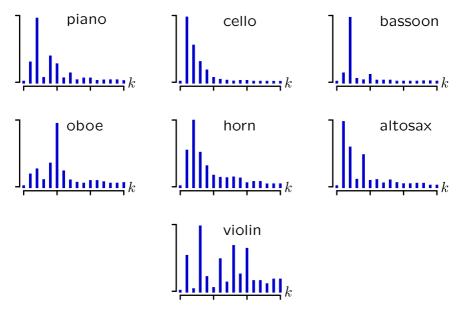
Ex: musical instruments (http://theremin.music.uiowa.edu/MIS)



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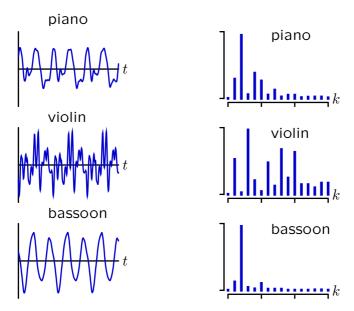
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Musical Instruments

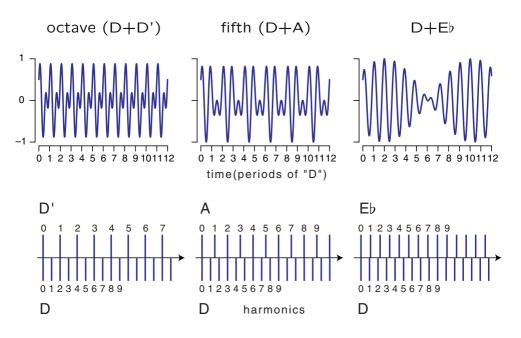
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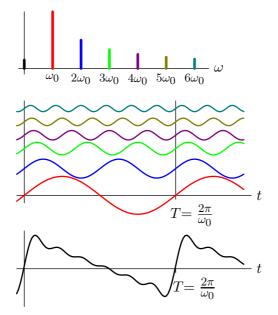
Harmonics

Harmonic structure determines consonance and dissonance.



Harmonic Representations

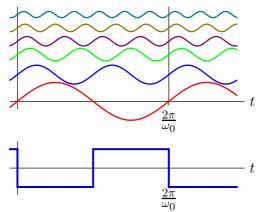
What signals can be represented by sums of harmonic components?



Only periodic signals: all harmonics of ω_0 are periodic in $T = 2\pi/\omega_0$.

Harmonic Representations

Is it possible to represent ALL periodic signals with harmonics? What about discontinuous signals?



Fourier claimed YES — even though all harmonics are continuous! Lagrange ridiculed the idea that a discontinuous signal could be written as a sum of continuous signals.

We will assume the answer is YES and see if the answer makes sense.

Separating harmonic components

Underlying properties.

1. Multiplying two harmonics produces a new harmonic with the same fundamental frequency:

$$e^{jk\omega_0 t} \times e^{jl\omega_0 t} = e^{j(k+l)\omega_0 t}$$

2. The integral of a harmonic over any time interval with length equal to a period T is zero unless the harmonic is at DC:

$$\int_{t_0}^{t_0+T} e^{jk\omega_0 t} dt \equiv \int_T e^{jk\omega_0 t} dt = \begin{cases} 0, & k \neq 0\\ T, & k = 0 \end{cases}$$
$$= T\delta[k]$$

Separating harmonic components

Assume that x(t) is periodic in T and is composed of a weighted sum of harmonics of $\omega_0 = 2\pi/T$.

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt}$$

Then

$$\int_T x(t)e^{-jl\omega_0 t}dt = \int_T \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt}e^{-j\omega_0 lt}dt$$
$$= \sum_{k=-\infty}^{\infty} a_k \int_T e^{j\omega_0 (k-l)t}dt$$
$$= \sum_{k=-\infty}^{\infty} a_k T\delta[k-l] = Ta_l$$

Therefore

$$a_k = \frac{1}{T} \int_T x(t) e^{-j\omega_0 kt} dt \qquad = \frac{1}{T} \int_T x(t) e^{-j\frac{2\pi}{T}kt} dt$$

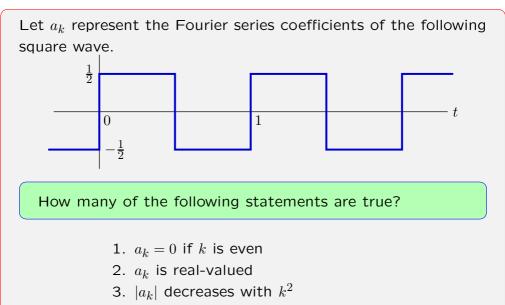
Determining harmonic components of a periodic signal.

$$a_k = \frac{1}{T} \int_T x(t) e^{-j\frac{2\pi}{T}kt} dt$$

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt}$$

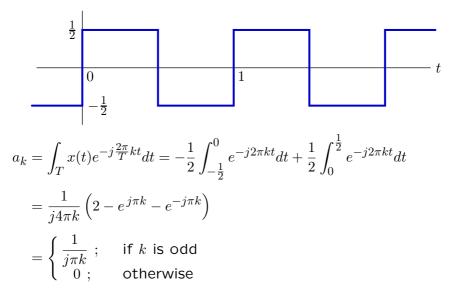
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("analysis" equation)



- 4. there are an infinite number of non-zero a_k
- 5. all of the above

Let a_k represent the Fourier series coefficients of the following square wave.

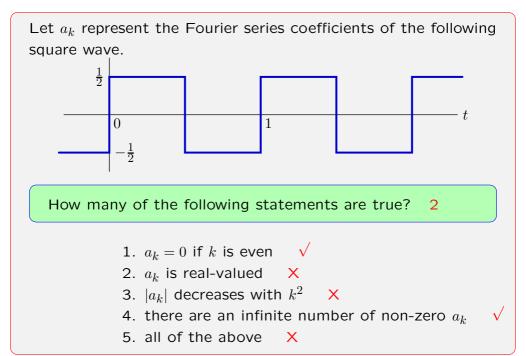


Let a_k represent the Fourier series coefficients of the following square wave.

$$a_k = \left\{ egin{array}{cc} rac{1}{j\pi k} \ ; & \mbox{if } k \ \mbox{is odd} \\ 0 \ ; & \mbox{otherwise} \end{array}
ight.$$

How many of the following statements are true?

- 1. $a_k = 0$ if k is even \checkmark
- 2. a_k is real-valued X
- 3. $|a_k|$ decreases with k^2 X
- 4. there are an infinite number of non-zero a_k
- 5. all of the above \times



Fourier Series Properties

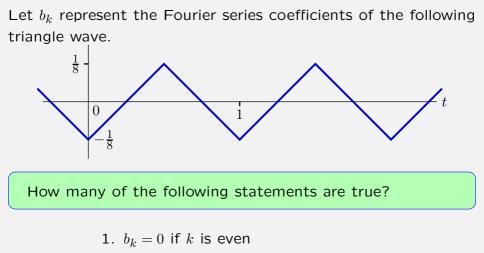
If a signal is differentiated in time, its Fourier coefficients are multiplied by $j\frac{2\pi}{T}k$.

Proof: Let

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt}$$

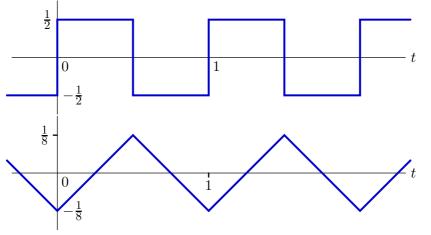
then

$$\dot{x}(t) = \dot{x}(t+T) = \sum_{k=-\infty}^{\infty} \left(j \frac{2\pi}{T} k \, a_k \right) e^{j \frac{2\pi}{T} k t}$$



- 2. b_k is real-valued
- 3. $|b_k|$ decreases with k^2
- 4. there are an infinite number of non-zero b_k
- 5. all of the above

The triangle waveform is the integral of the square wave.



Therefore the Fourier coefficients of the triangle waveform are $\frac{1}{j2\pi k}$ times those of the square wave.

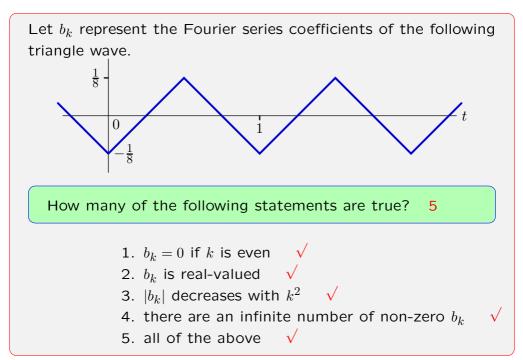
$$b_k = rac{1}{jk\pi} imes rac{1}{j2\pi k} = rac{-1}{2k^2\pi^2} \; ; \; k \; {
m odd}$$

Let \boldsymbol{b}_k represent the Fourier series coefficients of the following triangle wave.

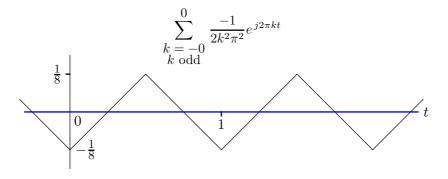
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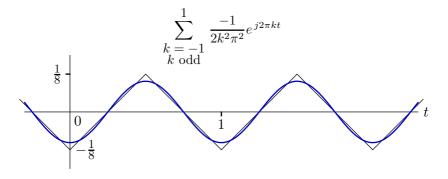
- 1. $b_k = 0$ if k is even \checkmark
- 2. b_k is real-valued \checkmark
- 3. $|b_k|$ decreases with $k^2 \sqrt{}$
- 4. there are an infinite number of non-zero b_k
- 5. all of the above \checkmark



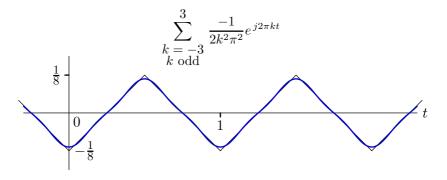
One can visualize convergence of the Fourier Series by incrementally adding terms.



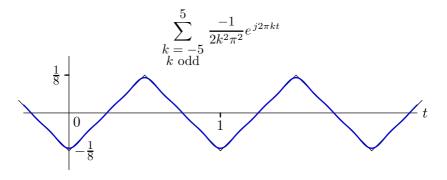
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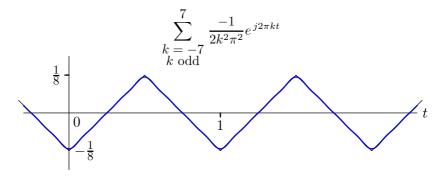
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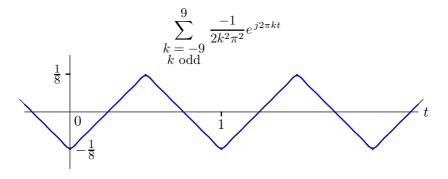
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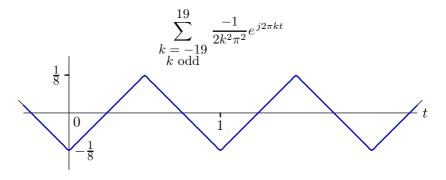
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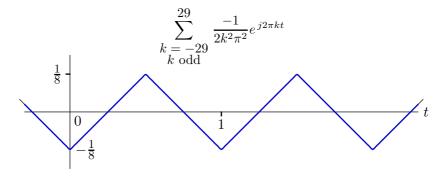
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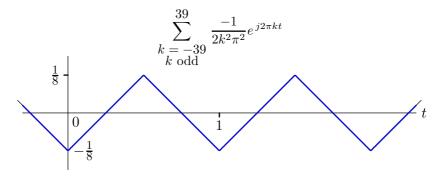


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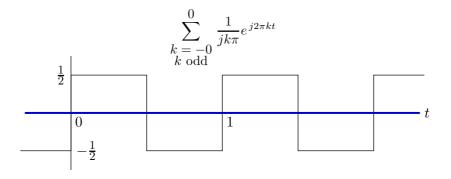
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Example: triangle waveform

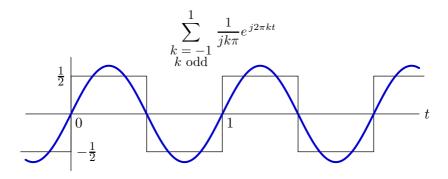


Fourier series representations of functions with discontinuous slopes converge toward functions with discontinuous slopes.

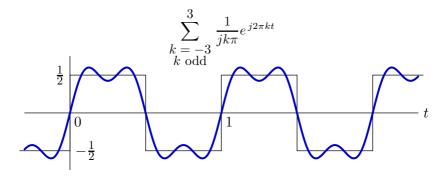
One can visualize convergence of the Fourier Series by incrementally adding terms.



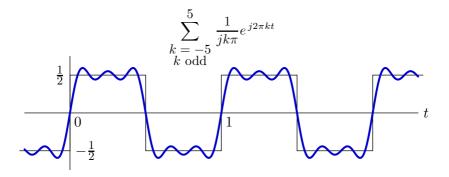
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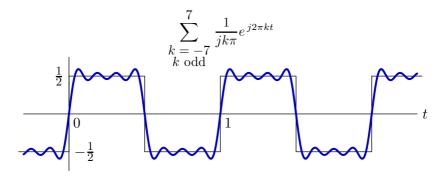
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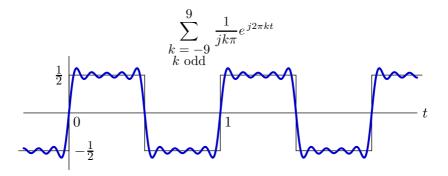
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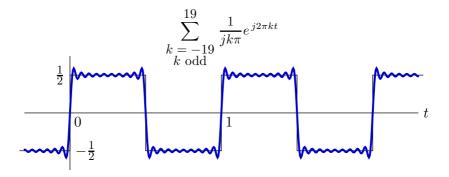
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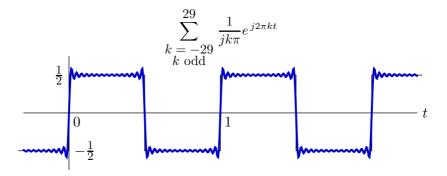
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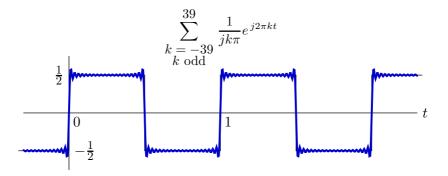
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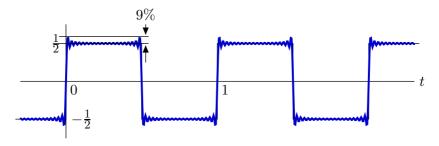
One can visualize convergence of the Fourier Series by incrementally adding terms.



One can visualize convergence of the Fourier Series by incrementally adding terms.



Partial sums of Fourier series of discontinuous functions "ring" near discontinuities: Gibb's phenomenon.



This ringing results because the magnitude of the Fourier coefficients is only decreasing as $\frac{1}{k}$ (while they decreased as $\frac{1}{k^2}$ for the triangle). You can decrease (and even eliminate the ringing) by decreasing the magnitudes of the Fourier coefficients at higher frequencies.

Fourier Series: Summary

Fourier series represent periodic signals as sums of sinusoids.

- valid for an extremely large class of periodic signals
- valid even for discontinuous signals such as square wave

However, convergence as *#* harmonics increases can be complicated.

Filtering

The output of an LTI system is a "filtered" version of the input.

Input: Fourier series \rightarrow sum of complex exponentials.

$$x(t) = x(t+T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt}$$

Complex exponentials: eigenfunctions of LTI systems.

$$e^{j\frac{2\pi}{T}kt} \to H(j\frac{2\pi}{T}k)e^{j\frac{2\pi}{T}kt}$$

Output: same eigenfunctions, amplitudes/phases set by system.

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt} \to y(t) = \sum_{k=-\infty}^{\infty} a_k H(j\frac{2\pi}{T}k) e^{j\frac{2\pi}{T}kt}$$

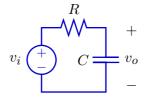
Filtering

Notion of a filter.

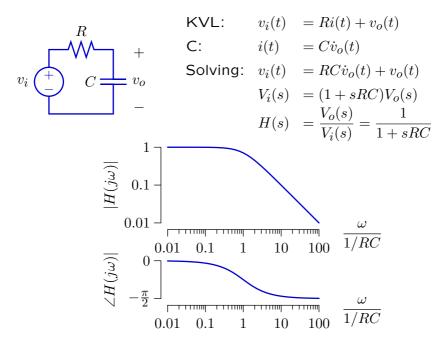
LTI systems

- cannot create new frequencies.
- can scale magnitudes and shift phases of existing components.

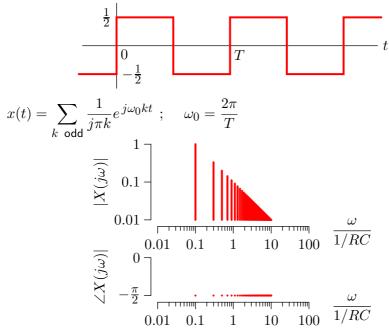
Example: Low-Pass Filtering with an RC circuit



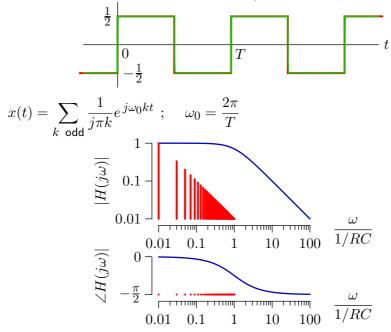
Calculate the frequency response of an RC circuit.



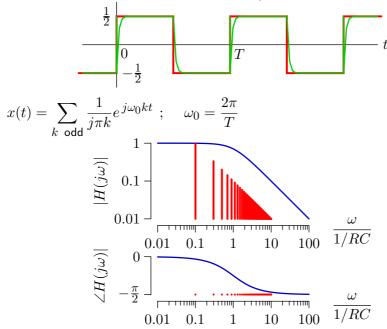
Let the input be a square wave.



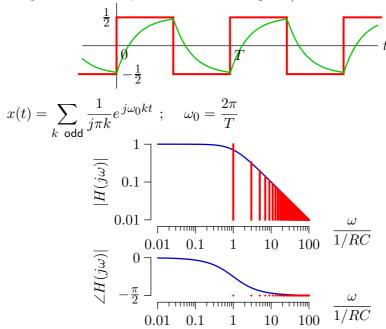
Low frequency square wave: $\omega_0 \ll 1/RC$.



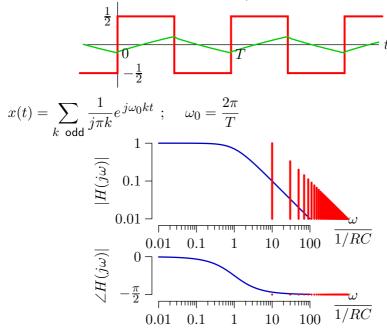
Higher frequency square wave: $\omega_0 < 1/RC$.



Still higher frequency square wave: $\omega_0 = 1/RC$.



High frequency square wave: $\omega_0 > 1/RC$.



Fourier Series: Summary

Fourier series represent signals by their frequency content.

Representing a signal by its frequency content is useful for many signals, e.g., music.

Fourier series motivate a new representation of a system as a filter.