

# 6.003: Signals and Systems

## Fourier Transform

*April 6, 2010*

## Mid-term Examination #2

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Tomorrow, April 7, 7:30-9:30pm, 34-101.

No recitations tomorrow.

Coverage:     Lectures 1–15  
                  Recitations 1–15  
                  Homeworks 1–8

Homework 8 will not be collected or graded. Solutions are posted.

Closed book: 2 pages of notes ( $8\frac{1}{2} \times 11$  inches; front and back).

Designed as 1-hour exam; two hours to complete.

## Last Week: Fourier Series

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Representing periodic signals as sums of **sinusoids**.

→ new representations for systems as **filters**.

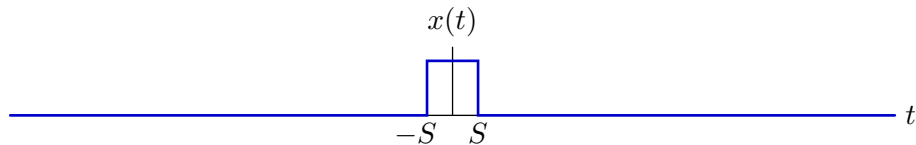
This week: generalize for aperiodic signals.

## Fourier Transform

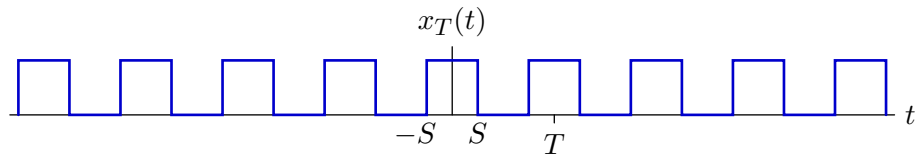
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An aperiodic signal can be thought of as periodic with infinite period.

Let  $x(t)$  represent an aperiodic signal.



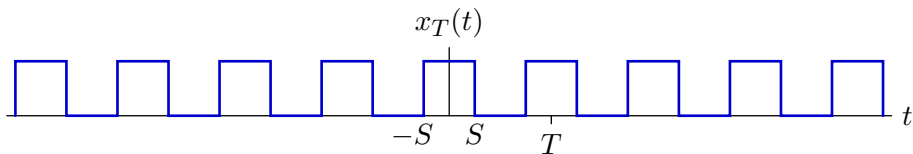
“Periodic extension”: 
$$x_T(t) = \sum_{k=-\infty}^{\infty} x(t + kT)$$



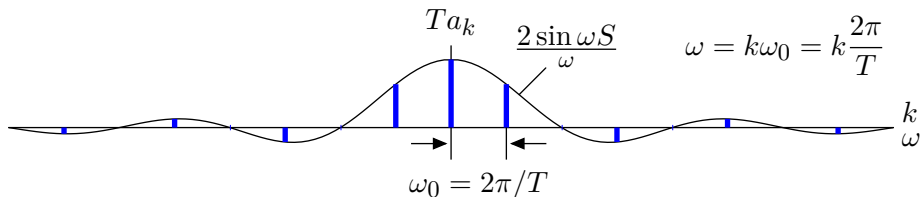
Then  $x(t) = \lim_{T \rightarrow \infty} x_T(t)$ .

# Fourier Transform

Represent  $x_T(t)$  by its Fourier series.

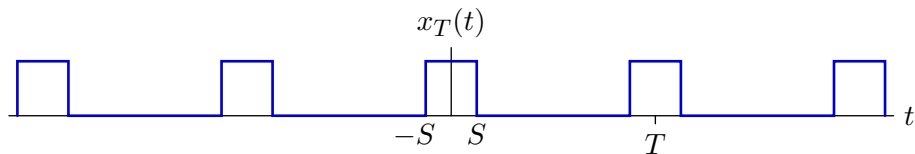


$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x_T(t) e^{-j\frac{2\pi}{T}kt} dt = \frac{1}{T} \int_{-S}^S e^{-j\frac{2\pi}{T}kt} dt = \frac{\sin \frac{2\pi kS}{T}}{\pi k} = \frac{2}{T} \frac{\sin \omega S}{\omega}$$

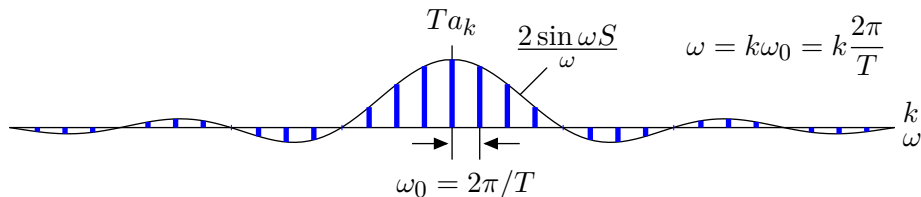


## Fourier Transform

Doubling period doubles # of harmonics in given frequency interval.

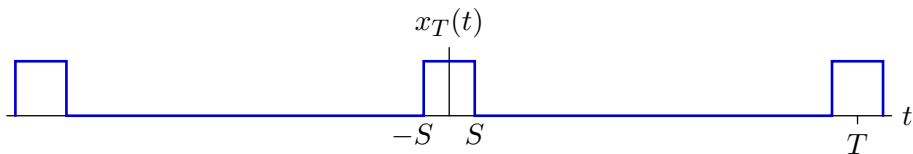


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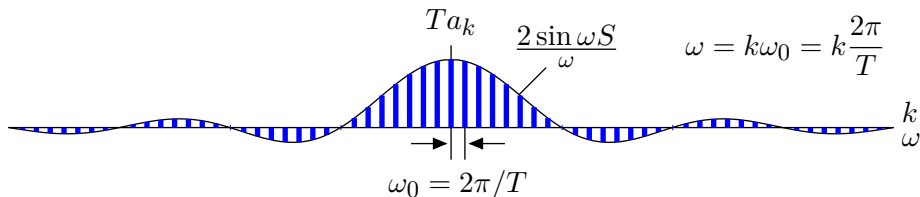


## Fourier Transform

As  $T \rightarrow \infty$ , discrete harmonic amplitudes  $\rightarrow$  a continuum  $E(\omega)$ .



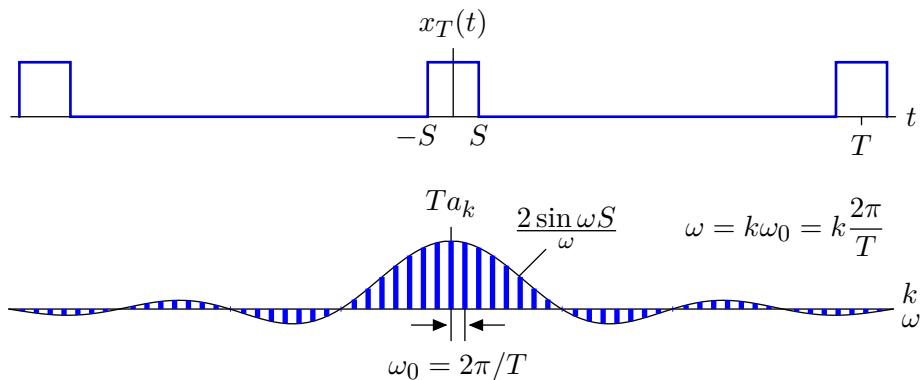
$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x_T(t) e^{-j\frac{2\pi}{T}kt} dt = \frac{1}{T} \int_{-S}^S e^{-j\frac{2\pi}{T}kt} dt = \frac{\sin \frac{2\pi kS}{T}}{\pi k} = \frac{2 \sin \omega S}{T \omega}$$



$$\lim_{T \rightarrow \infty} T a_k = \lim_{T \rightarrow \infty} \int_{-T/2}^{T/2} x(t) e^{-j\omega t} dt = \frac{2}{\omega} \sin \omega S = E(\omega)$$

## Fourier Transform

As  $T \rightarrow \infty$ , synthesis sum  $\rightarrow$  integral.



$$\lim_{T \rightarrow \infty} T a_k = \lim_{T \rightarrow \infty} \int_{-T/2}^{T/2} x(t) e^{-j\omega t} dt = \frac{2}{\omega} \sin \omega S = E(\omega)$$

$$x(t) = \sum_{k=-\infty}^{\infty} \underbrace{\frac{1}{T} E(\omega)}_{a_k} e^{j\frac{2\pi}{T} k t} = \sum_{k=-\infty}^{\infty} \frac{\omega_0}{2\pi} E(\omega) e^{j\omega t} \rightarrow \frac{1}{2\pi} \int_{-\infty}^{\infty} E(\omega) e^{j\omega t} d\omega$$



## Fourier Transform

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Replacing  $E(\omega)$  by  $X(j\omega)$  yields the Fourier transform relations.

$$E(\omega) = X(s)|_{s=j\omega} \equiv X(j\omega)$$

Fourier transform

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt \quad (\text{"analysis" equation})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega)e^{j\omega t} d\omega \quad (\text{"synthesis" equation})$$

## Fourier Transform

---

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Form is similar to that of Fourier series

→ provides alternate view of signal.

## Relation between Fourier and Laplace Transforms

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If the Laplace transform of a signal exists and if the ROC includes the  $j\omega$  axis, then the Fourier transform is equal to the Laplace transform evaluated on the  $j\omega$  axis.

Laplace transform:

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

Fourier transform:

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt = H(s)|_{s=j\omega}$$

## Relation between Fourier and Laplace Transforms

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Fourier transform “inherits” properties of Laplace transform.

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Property	$x(t)$	$X(s)$	$X(j\omega)$
Linearity	$ax_1(t) + bx_2(t)$	$aX_1(s) + bX_2(s)$	$aX_1(j\omega) + bX_2(j\omega)$
Time shift	$x(t - t_0)$	$e^{-st_0}X(s)$	$e^{-j\omega t_0}X(j\omega)$
Time scale	$x(at)$	$\frac{1}{ a }X\left(\frac{s}{a}\right)$	$\frac{1}{ a }X\left(\frac{j\omega}{a}\right)$
Differentiation	$\frac{dx(t)}{dt}$	$sX(s)$	$j\omega X(j\omega)$
Multiply by $t$	$tx(t)$	$-\frac{d}{ds}X(s)$	$-\frac{1}{j}\frac{d}{d\omega}X(j\omega)$
Convolution	$x_1(t) * x_2(t)$	$X_1(s) \times X_2(s)$	$X_1(j\omega) \times X_2(j\omega)$

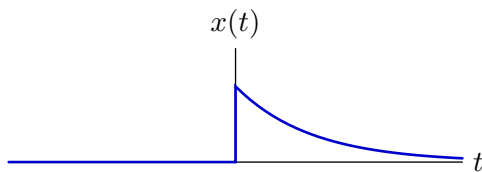
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## Relation between Fourier and Laplace Transforms

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There are also important differences.

Compare Fourier and Laplace transforms of  $x(t) = e^{-t}u(t)$ .



Laplace transform

$$X(s) = \int_{-\infty}^{\infty} e^{-t}u(t)e^{-st} dt = \int_0^{\infty} e^{-(s+1)t} dt = \frac{1}{1+s} ; \operatorname{Re}(s) > -1$$

a complex-valued function of **complex** domain.

Fourier transform

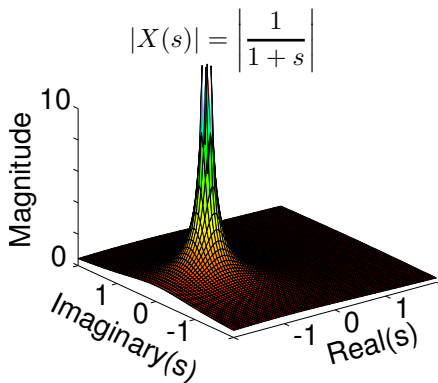
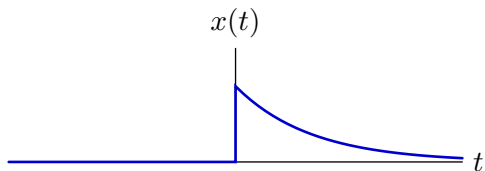
$$X(j\omega) = \int_{-\infty}^{\infty} e^{-t}u(t)e^{-j\omega t} dt = \int_0^{\infty} e^{-(j\omega+1)t} dt = \frac{1}{1+j\omega}$$

a complex-valued function of **real** domain.

## Laplace Transform

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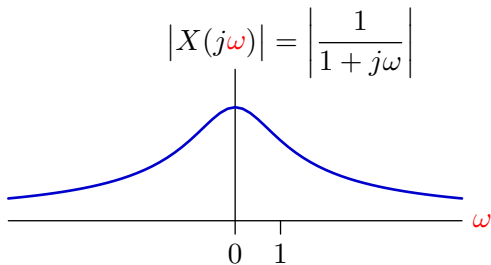
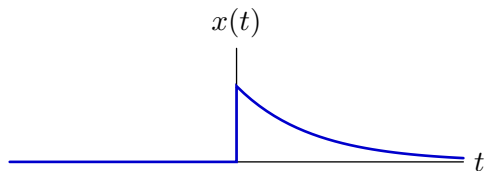
The Laplace transform maps a function of time  $t$  to a complex-valued function of complex-valued domain  $s$ .



## Fourier Transform

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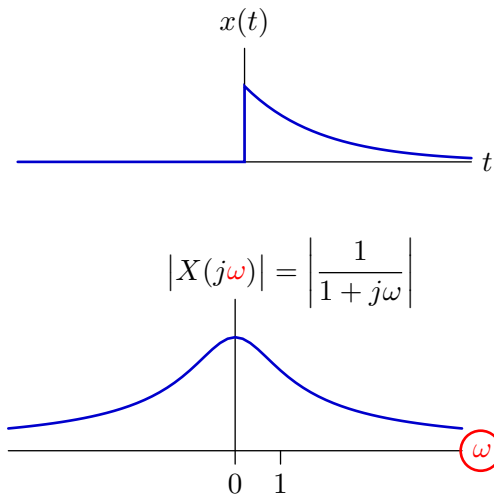
The Fourier transform maps a function of time  $t$  to a complex-valued function of real-valued domain  $\omega$ .



## Fourier Transform

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The Fourier transform maps a function of time  $t$  to a complex-valued function of real-valued domain  $\omega$ .

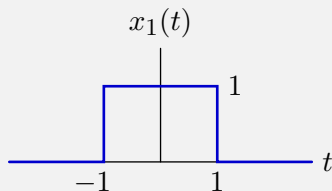


Frequency plots provide intuition that is difficult to otherwise obtain.



## Check Yourself

Find the Fourier transform of the following square pulse.

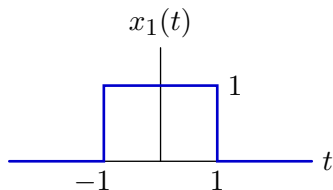


1.  $X_1(j\omega) = \frac{1}{\omega} (e^{\omega} - e^{-\omega})$
2.  $X_1(j\omega) = \frac{1}{\omega} \sin \omega$
3.  $X_1(j\omega) = \frac{2}{\omega} (e^{\omega} - e^{-\omega})$
4.  $X_1(j\omega) = \frac{2}{\omega} \sin \omega$
5. none of the above

## Fourier Transform

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Compare the Laplace and Fourier transforms of a square pulse.



Laplace transform:

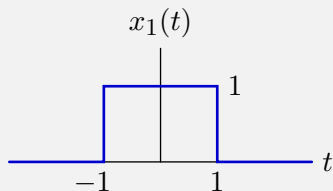
$$X_1(s) = \int_{-1}^1 e^{-st} dt = \left. \frac{e^{-st}}{-s} \right|_{-1}^1 = \frac{1}{s} (e^s - e^{-s}) \quad [\text{function of } s = \sigma + j\omega]$$

Fourier transform

$$X_1(j\omega) = \int_{-1}^1 e^{-j\omega t} dt = \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-1}^1 = \frac{2 \sin \omega}{\omega} \quad [\text{function of } \omega]$$

## Check Yourself

Find the Fourier transform of the following square pulse. 4

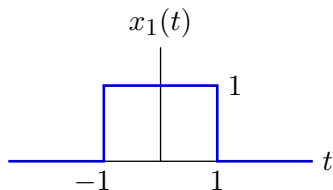


1.  $X_1(j\omega) = \frac{1}{\omega} (e^{\omega} - e^{-\omega})$
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4.  $X_1(j\omega) = \frac{2}{\omega} \sin \omega$
5. none of the above

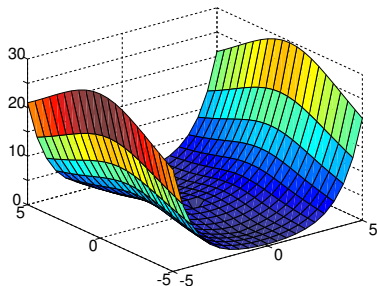
# Laplace Transform

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Laplace transform: complex-valued function of complex domain.



$$|X(s)| = \left| \frac{1}{s}(e^s - e^{-s}) \right|$$

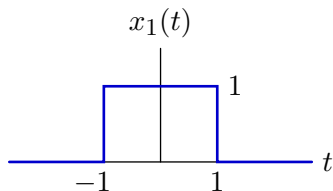


## Fourier Transform

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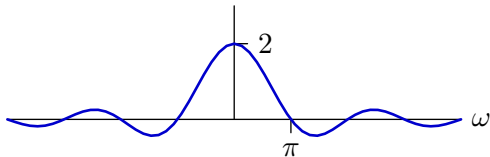
The Fourier transform is a function of real domain: frequency  $\omega$ .

Time representation:



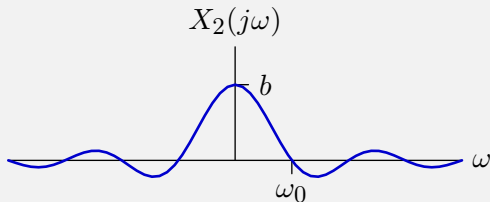
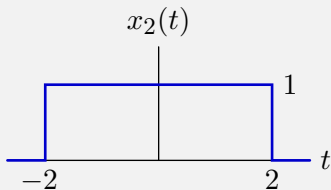
Frequency representation:

$$X_1(j\omega) = \frac{2 \sin \omega}{\omega}$$



## Check Yourself

Signal  $x_2(t)$  and its Fourier transform  $X_2(j\omega)$  are shown below.



Which is true?

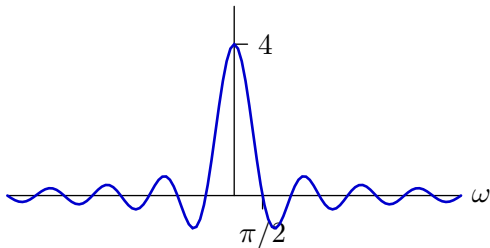
1.  $b = 2$  and  $\omega_0 = \pi/2$
2.  $b = 2$  and  $\omega_0 = 2\pi$
3.  $b = 4$  and  $\omega_0 = \pi/2$
4.  $b = 4$  and  $\omega_0 = 2\pi$
5. none of the above

## Check Yourself

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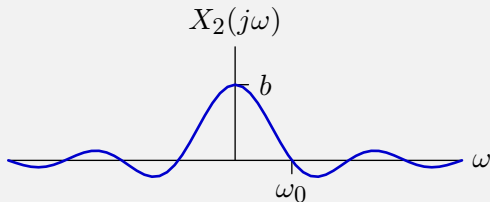
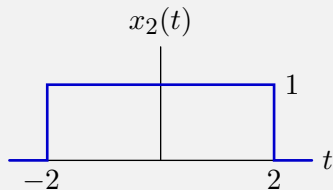
Find the Fourier transform.

$$X_2(j\omega) = \int_{-2}^2 e^{-j\omega t} dt = \left. \frac{e^{-j\omega t}}{-j\omega} \right|_{-2}^2 = \frac{2 \sin 2\omega}{\omega} = \frac{4 \sin 2\omega}{2\omega}$$



## Check Yourself

Signal  $x_2(t)$  and its Fourier transform  $X_2(j\omega)$  are shown below.



Which is true? **3**

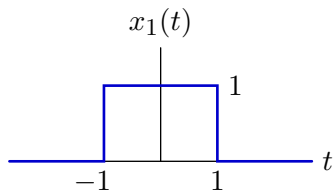
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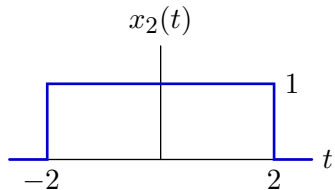
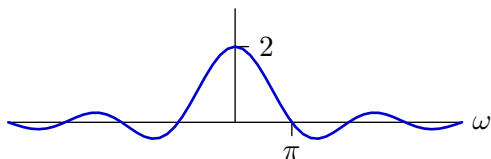
# Fourier Transforms

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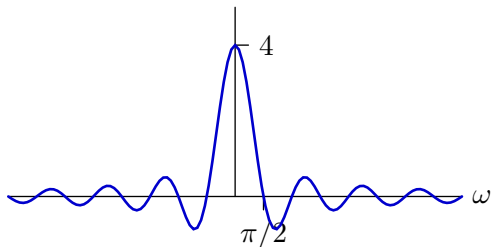
Stretching time compresses frequency.



$$X_1(j\omega) = \frac{2 \sin \omega}{\omega}$$



$$X_2(j\omega) = \frac{4 \sin 2\omega}{2\omega}$$



## Check Yourself

---

Stretching time compresses frequency.

Find a general scaling rule.

Let  $x_2(t) = x_1(at)$ .

If time is stretched in going from  $x_1$  to  $x_2$ , is  $a > 1$  or  $a < 1$ ?

## Check Yourself

---

Stretching time compresses frequency.

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Let  $x_2(t) = x_1(at)$ .

If time is stretched in going from  $x_1$  to  $x_2$ , is  $a > 1$  or  $a < 1$ ?

$$x_2(2) = x_1(1)$$

$$x_2(t) = x_1(at)$$

Therefore  $a = 1/2$ , or more generally,  $a < 1$ .

## Check Yourself

---

Stretching time compresses frequency.

Find a general scaling rule.

Let  $x_2(t) = x_1(at)$ .

If time is stretched in going from  $x_1$  to  $x_2$ , is  $a > 1$  or  $a < 1$ ?

$a < 1$

## Fourier Transforms

---

Find a general scaling rule.

Let  $x_2(t) = x_1(at)$ .

$$X_2(j\omega) = \int_{-\infty}^{\infty} x_2(t)e^{-j\omega t} dt = \int_{-\infty}^{\infty} x_1(at)e^{-j\omega t} dt$$

Let  $\tau = at$  ( $a > 0$ ).

$$X_2(j\omega) = \int_{-\infty}^{\infty} x_1(\tau)e^{-j\omega\tau/a} \frac{1}{a} d\tau = \frac{1}{a} X_1\left(\frac{j\omega}{a}\right)$$

If  $a < 0$  the sign of  $d\tau$  would change along with the limits of integration. In general,

$$x_1(at) \leftrightarrow \frac{1}{|a|} X_1\left(\frac{j\omega}{a}\right).$$

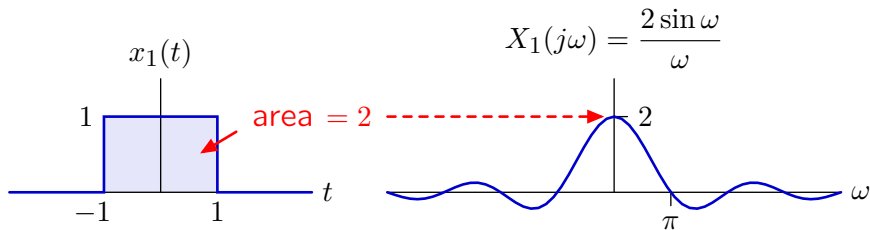
If time is stretched ( $a < 1$ ) then frequency is compressed and amplitude increases (preserving area).

## Moments

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The value of  $X(j\omega)$  at  $\omega = 0$  is the integral of  $x(t)$  over time  $t$ .

$$X(j\omega)|_{\omega=0} = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt = \int_{-\infty}^{\infty} x(t)e^{j0t} dt = \int_{-\infty}^{\infty} x(t) dt$$

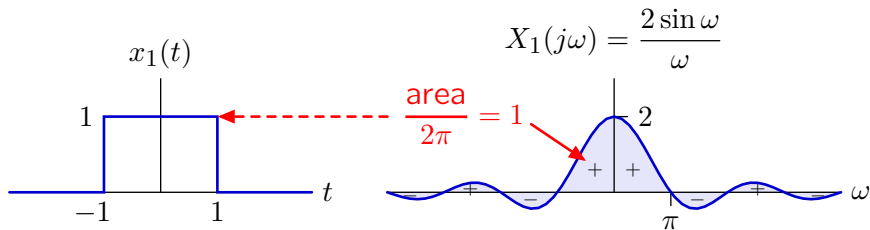


## Moments

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The value of  $x(0)$  is the integral of  $X(j\omega)$  divided by  $2\pi$ .

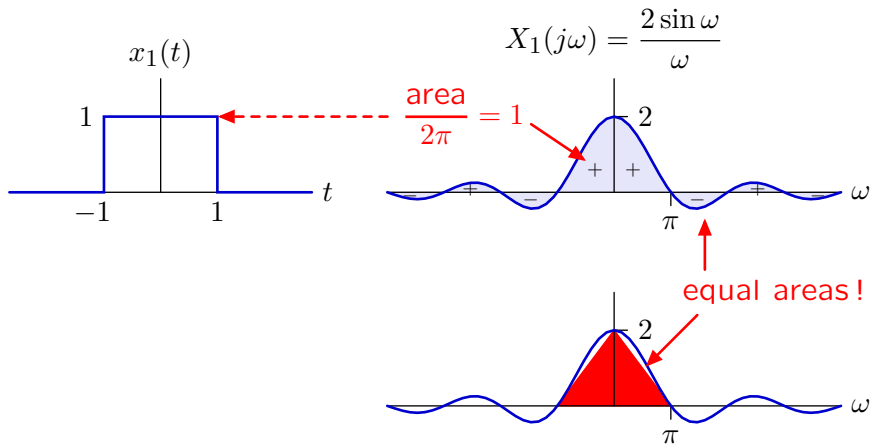
$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) d\omega$$



## Moments

The value of  $x(0)$  is the integral of  $X(j\omega)$  divided by  $2\pi$ .

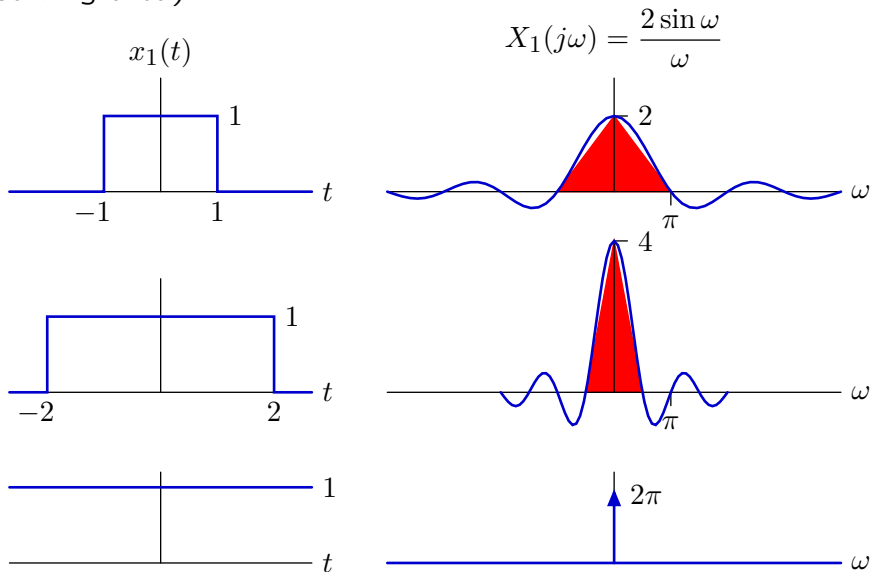
$$x(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) d\omega$$





## Stretching to the Limit

Stretching time compresses frequency and increases amplitude (preserving area).



New way to think about an impulse!

## Fourier Transform

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One of the most useful features of the Fourier transform (and Fourier series) is the simple “inverse” Fourier transform.

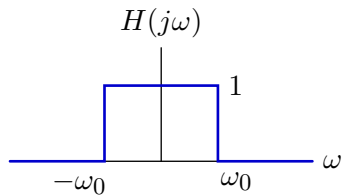
$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt \quad (\text{Fourier transform})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega)e^{j\omega t} d\omega \quad (\text{“inverse” Fourier transform})$$

## Inverse Fourier Transform

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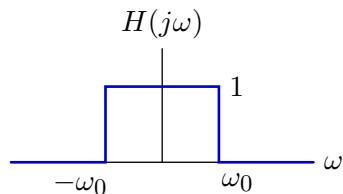
Find the impulse response of an “ideal” low pass filter.



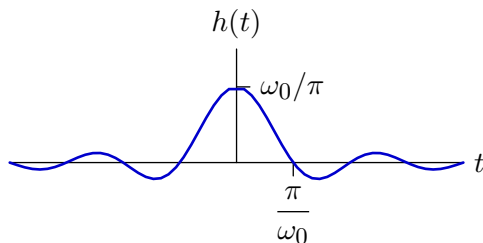
## Inverse Fourier Transform

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Find the impulse response of an “ideal” low pass filter.



$$h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(j\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\omega_0}^{\omega_0} e^{j\omega t} d\omega = \frac{1}{2\pi} \left. \frac{e^{j\omega t}}{jt} \right|_{-\omega_0}^{\omega_0} = \frac{\sin \omega_0 t}{\pi t}$$



This result is not so easily obtained without inverse relation.

## Fourier Transform

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The Fourier transform and its inverse have very similar forms.

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt \quad (\text{Fourier transform})$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega)e^{j\omega t} d\omega \quad (\text{"inverse" Fourier transform})$$

Convert one to the other by

- $t \rightarrow \omega$
- $\omega \rightarrow -t$
- scale by  $2\pi$

# Duality

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The Fourier transform and its inverse have very similar forms.

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega)e^{j\omega t} d\omega$$

Two differences:

- minus sign: flips time axis (or equivalently, frequency axis)
- divide by  $2\pi$  (or multiply in the other direction)

$$x_1(t) = f(t) \leftrightarrow X_1(j\omega) = g(\omega)$$

$\omega \rightarrow t$   $t \rightarrow \omega$  ; flip ;  $\times 2\pi$


$$x_2(t) = g(t) \leftrightarrow X_2(j\omega) = 2\pi f(-\omega)$$

## Duality

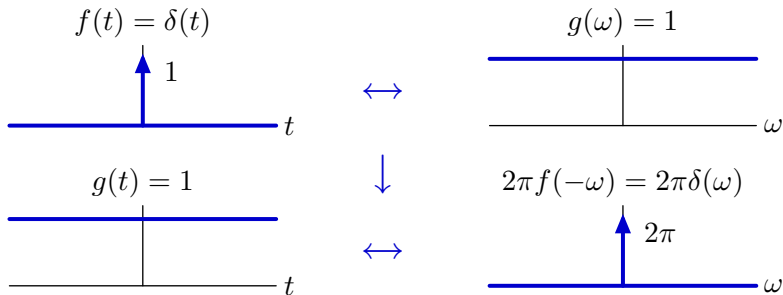
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Using duality to find new transform pairs.

$$x_1(t) = f(t) \leftrightarrow X_1(j\omega) = g(\omega)$$

$\omega \rightarrow t$    $t \rightarrow \omega$  ; flip ;  $\times 2\pi$

$$x_2(t) = g(t) \leftrightarrow X_2(j\omega) = 2\pi f(-\omega)$$

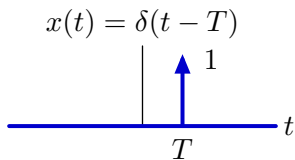


The function  $g(t) = 1$  does not have a Laplace transform!

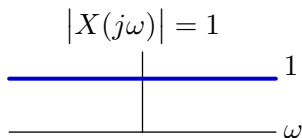
## More Impulses

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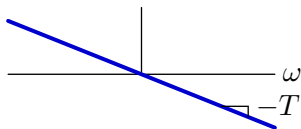
Fourier transform of delayed impulse:  $\delta(t - T) \leftrightarrow e^{-j\omega T}$ .



$$X(j\omega) = \int_{-\infty}^{\infty} \delta(t - T)e^{-j\omega t} dt = e^{-j\omega T}$$



$$\angle X(j\omega) = -\omega T$$





## Eternal Sinusoids

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Using duality to find the Fourier transform of an eternal sinusoid.

$$\begin{array}{ccc} \delta(t - T) & \leftrightarrow & e^{-j\omega T} \\ \omega \rightarrow t & \begin{array}{c} \text{red lines crossing} \\ \text{with arrows pointing to } \omega \rightarrow t \text{ and } t \rightarrow \omega \end{array} & t \rightarrow \omega ; \text{ flip ; } \times 2\pi \\ e^{-jtT} & \leftrightarrow & 2\pi\delta(\omega + T) \\ T \rightarrow \omega_0 : & & \\ e^{-j\omega_0 t} & \leftrightarrow & 2\pi\delta(\omega + \omega_0) \end{array}$$

$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt} \quad \begin{array}{c} \text{CTFS} \\ \longleftrightarrow \end{array} \quad \{a_k\}$$

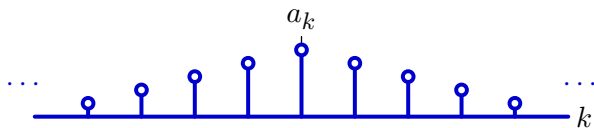
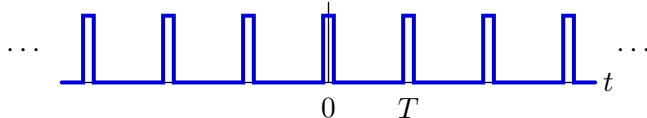
$$x(t) = x(t + T) = \sum_{k=-\infty}^{\infty} a_k e^{j\frac{2\pi}{T}kt} \quad \begin{array}{c} \text{CTFT} \\ \longleftrightarrow \end{array} \quad \sum_{k=-\infty}^{\infty} 2\pi a_k \delta\left(\omega - \frac{2\pi}{T}k\right)$$

## Relation between Fourier Transform and Fourier Series

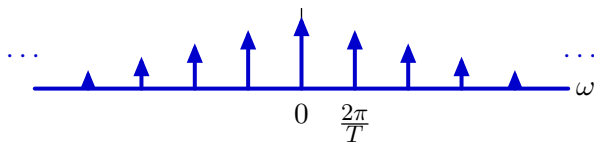
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Each term in the Fourier series is replaced by an impulse.

$$x(t) = \sum_{k=-\infty}^{\infty} x_p(t - kT)$$



$$X(j\omega) = \sum_{k=-\infty}^{\infty} 2\pi a_k \delta\left(\omega - k\frac{2\pi}{T}\right)$$



## Mid-term Examination #2

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Tomorrow, April 7, 7:30-9:30pm, 34-101.

No recitations tomorrow.

Coverage:     Lectures 1–15  
                  Recitations 1–15  
                  Homeworks 1–8

Homework 8 will not be collected or graded. Solutions are posted.

Closed book: 2 pages of notes ( $8\frac{1}{2} \times 11$  inches; front and back).

Designed as 1-hour exam; two hours to complete.