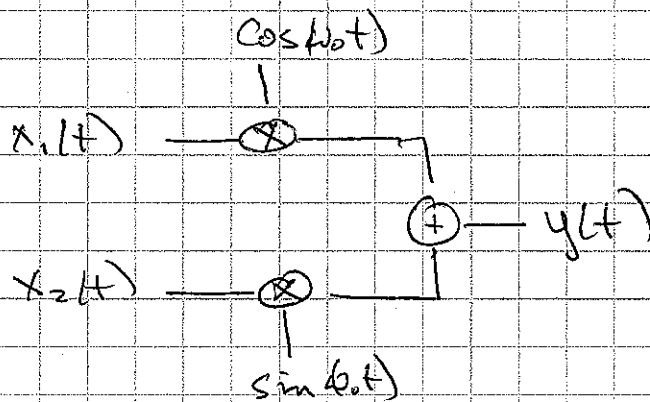


Example Quadrature modulation

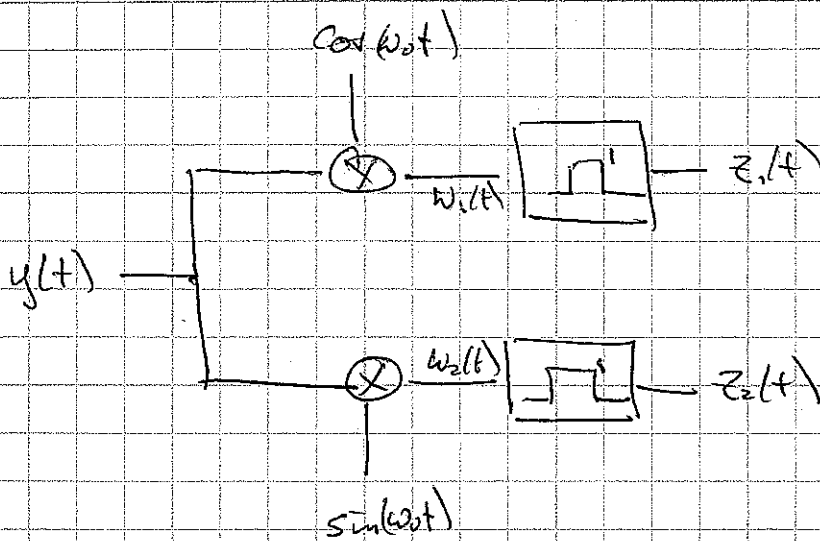
Basic idea is that cosine and sine functions are orthogonal, so that perhaps we might encode two signals instead of one.

Modulation scheme

$$y(t) = x_1(t) \cos(\omega_0 t) + \frac{1}{2} x_2(t) \sin(\omega_0 t)$$

$$Y(j\omega) = \frac{1}{2} \left[\overline{X_1(j(\omega + \omega_0))} + \overline{X_1(j(\omega - \omega_0))} \right] + \frac{1}{2j} \left[\overline{X_2(j(\omega + \omega_0))} - \overline{X_2(j(\omega - \omega_0))} \right]$$

Demodulation



$$w_1(t) = \cos(\omega_0 t) \left[x_1(t) \cos(\omega_0 t) + x_2(t) \sin(\omega_0 t) \right]$$

$$= \left[\frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} \right] \left[\frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} \right] x_1(t)$$

$$+ \left[\frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} \right] \left[\frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} \right] x_2(t)$$

$$= \frac{1}{4} \left[e^{2j\omega_0 t} + 2 + e^{-2j\omega_0 t} \right] x_1(t)$$

$$+ \frac{1}{4j} \left[e^{2j\omega_0 t} - e^{-2j\omega_0 t} \right] x_2(t)$$

So low pass filter of $w_1(t) \Rightarrow \frac{1}{2} x_1(t)$

$$w_2(t) = \sin(\omega_0 t) \left[x_1(t) \cos \omega_0 t + x_2(t) \sin \omega_0 t \right]$$

$$= \left[\frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} \right] \left[\frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} \right] x_1(t)$$

$$+ \left[\frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} \right] \left[\frac{e^{j\omega_0 t} - e^{-j\omega_0 t}}{2j} \right] x_2(t)$$

$$= \frac{1}{4j} \left[e^{2j\omega_0 t} - e^{-2j\omega_0 t} \right] x_1(t)$$

$$- \frac{1}{4} \left[e^{2j\omega_0 t} - 2 + e^{-2j\omega_0 t} \right] x_2(t)$$

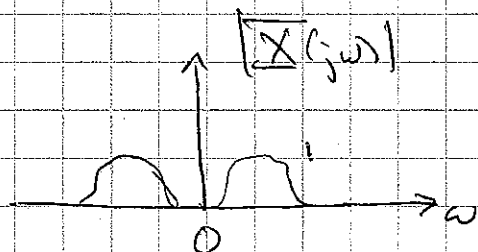
Low pass filtering $w_2(t) \Rightarrow \frac{1}{2} x_2(t)$

So, the scheme seems to work!

(23)

Example: Single-sideband modulation and demodulation

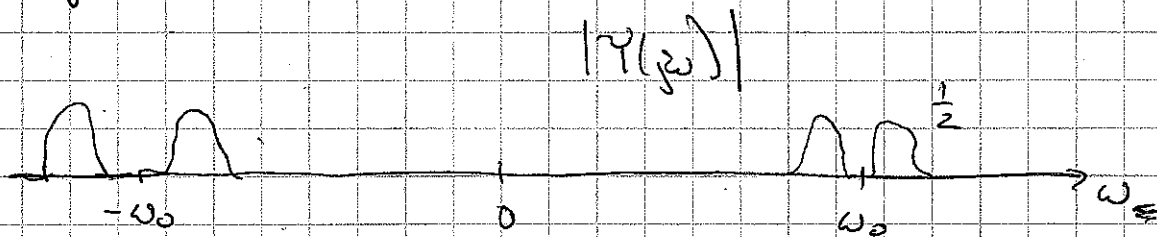
Thinking:



Transform of $x(t)$ has both positive and negative frequency components. We need both to construct the signal.

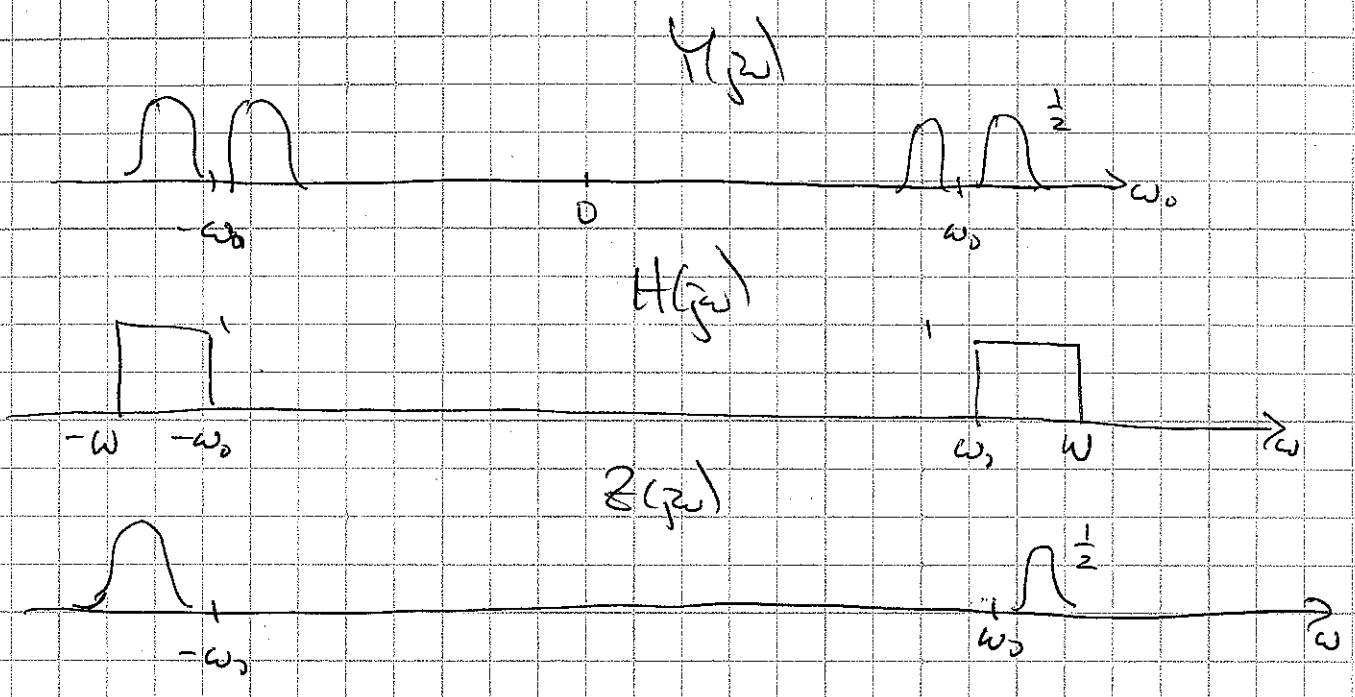
More thinking:

Consider transform of modulated $\cos(\omega_0 t) x(t) = y(t)$ signal.

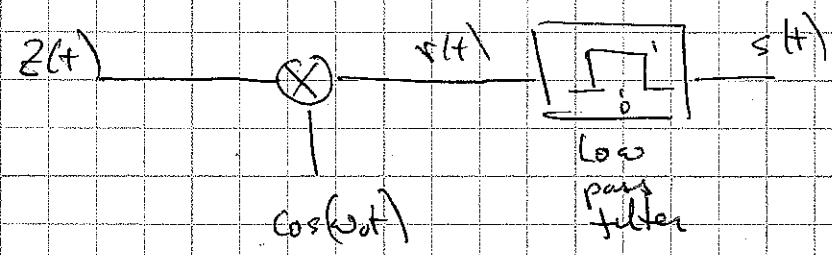


This one has $\frac{1}{2}$ copies, each with + and - frequency components. It is possible to eliminate half of the modulated signal!

Use band-pass filter



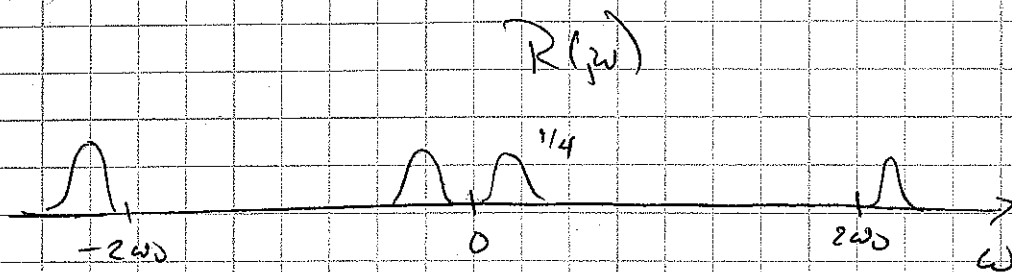
Demodulation



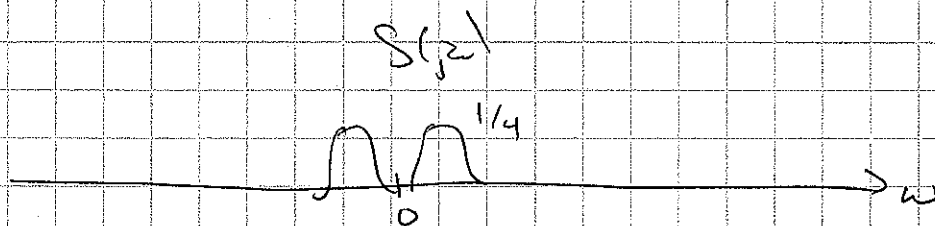
$$v(t) = z(t) \cos(\omega_0 t)$$

$$R(z) = \frac{1}{2\pi} Z(z) * \frac{1}{2} [2\pi(\omega - \omega_0) + 2\pi(\omega + \omega_0)]$$

(S2)



Next, low-pass filter to recover $s(t)$



So, it looks like $s(t) = \frac{1}{4} x(t)$, and the demodulation is successful.

What is $z(t)$?

$$z(t) = \int_{-\infty}^{\infty} z(j\omega) e^{j\omega t} \frac{d\omega}{2\pi}$$

$$= \int_{\omega_0}^{\omega} \overline{X}(j(\omega-\omega_0)) e^{j\omega t} \frac{d\omega}{2\pi}$$

$$+ \int_{-\omega}^{-\omega_0} \overline{X}(j(\omega+\omega_0)) e^{j\omega t} \frac{d\omega}{2\pi}$$

$$\overline{X}(j(\omega-\omega_0)) = \int_{-\infty}^{\infty} e^{-j(\omega-\omega_0)t} x(t) dt$$

$$\overline{X}(j(\omega+\omega_0)) = \int_{-\infty}^{\infty} e^{-j(\omega+\omega_0)t} x(t) dt$$

$$z(t) = \int_{\omega_0}^{\omega} \left(\int_{-\infty}^{\infty} dt' e^{-j(\omega-\omega_0)t'} x(t') \right) e^{j\omega t} \frac{d\omega}{2\pi}$$

$$+ \int_{-\omega}^{-\omega_0} \left(\int_{-\infty}^{\infty} dt' e^{-j(\omega+\omega_0)t'} x(t') \right) e^{j\omega t} \frac{d\omega}{2\pi}$$

$$= \int_{-\infty}^{\infty} x(t') e^{j\omega_0 t'} \left(\int_{\omega_0}^{\omega} e^{j\omega(t-t')} \frac{d\omega}{2\pi} \right) dt'$$

$$+ \int_{-\infty}^{\infty} x(t') e^{-j\omega_0 t'} \left(\int_{-\omega}^{-\omega_0} e^{j\omega(t-t')} \frac{d\omega}{2\pi} \right) dt'$$

$$\int_{-\omega_0}^{\omega} e^{j\omega(t-t')} \frac{d\omega}{2\pi} = \left. \frac{e^{j\omega(t-t')}}{2\pi j(t-t')} \right|_{-\omega_0}^{\omega}$$

$$= \frac{e^{j\omega(t-t')} - e^{j\omega_0(t-t')}}{2\pi j(t-t')}$$

$$\int_{-\omega}^{-\omega_0} e^{j\omega(t-t')} \frac{d\omega}{2\pi} = \left. \frac{e^{j\omega(t-t')}}{2\pi j(t-t')} \right|_{-\omega}^{-\omega_0}$$

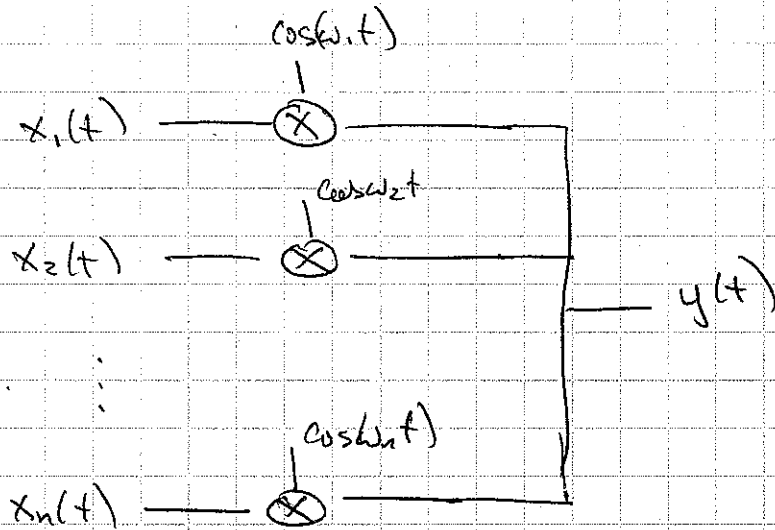
$$= \frac{e^{-j\omega_0(t-t')} - e^{-j\omega(t-t')}}{2\pi j(t-t')}$$

$$z(t) = \int_{-\infty}^{\infty} x(t') e^{j\omega_0 t} \left[\frac{e^{j\omega(t-t')} - e^{j\omega_0(t-t')}}{2\pi j(t-t')} \right] dt'$$

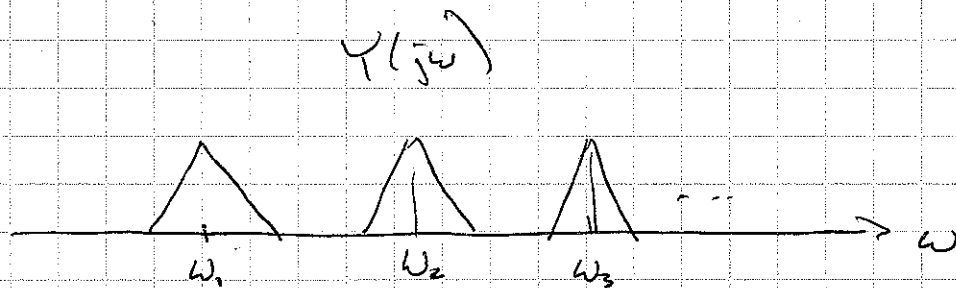
$$+ \int_{-\infty}^{\infty} x(t') e^{-j\omega_0 t} \left[\frac{e^{-j\omega_0(t-t')} - e^{-j\omega(t-t')}}{2\pi j(t-t')} \right] dt'$$

Example: Multiplexing

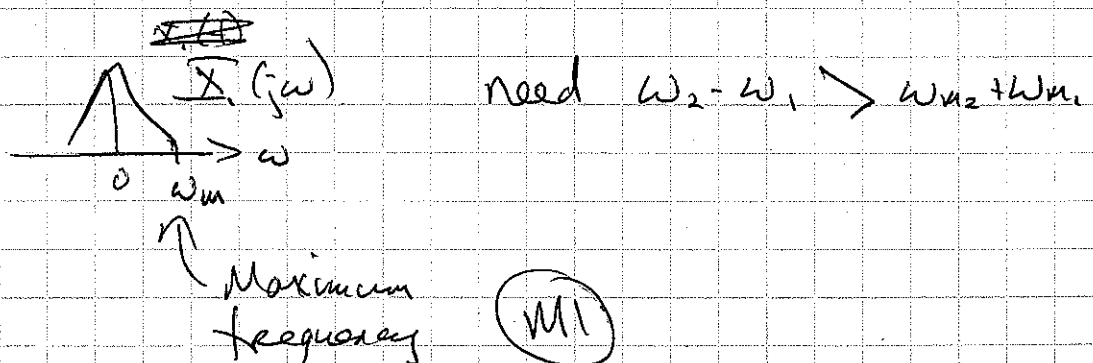
First toy:



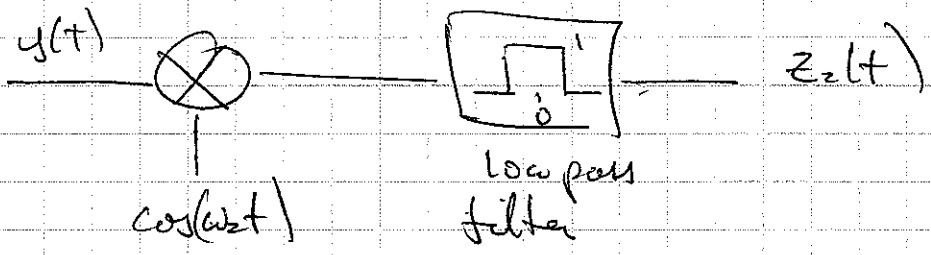
What do things look like in frequency?



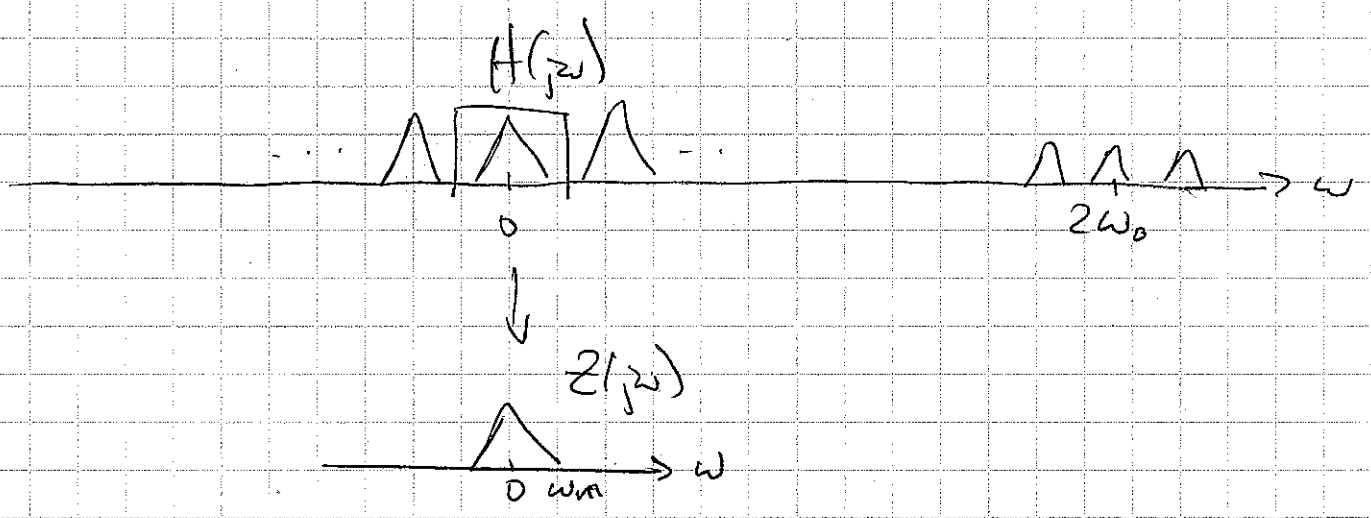
Thinking about frequency separation



What about de-modulation?



Look at it in frequency

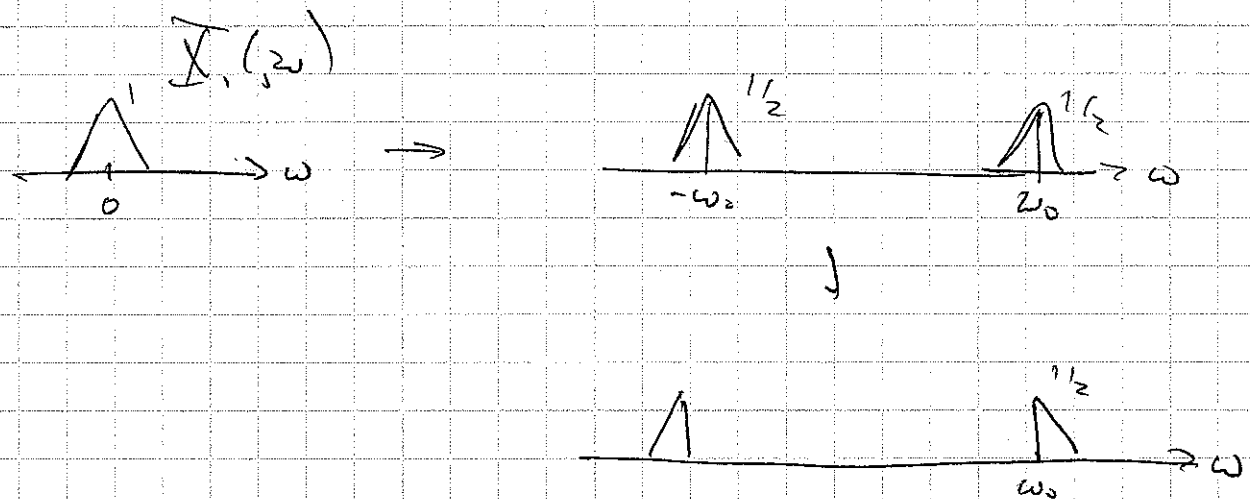


Simple ~~general~~ demodulation scheme that we have been using seems to work ok here.

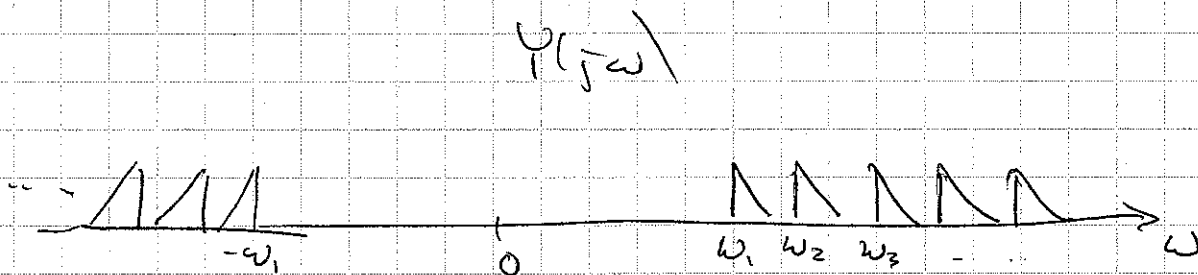
M2

Can we fit in more signals into available bandwidth?

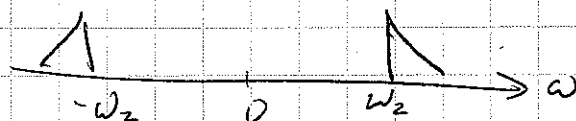
Make use of single side-band modulation



Then add to other single side band signals



Demodulation - first filter at high frequency to get



then ~~the~~ modulate with $\cos(\omega_c t)$ and low-pass filter.

(M3)