

Lecture 25

Multistage Amplifiers (II)

DC VOLTAGE AND CURRENT SOURCES

Outline

1. DC Voltage Sources
2. DC Current Sources and Sinks
3. Introduction to Differential Amplifiers

Reading Assignment:

Howe and Sodini, Chapter 9, Sections 9-3-9.4

Summary of Key Concepts

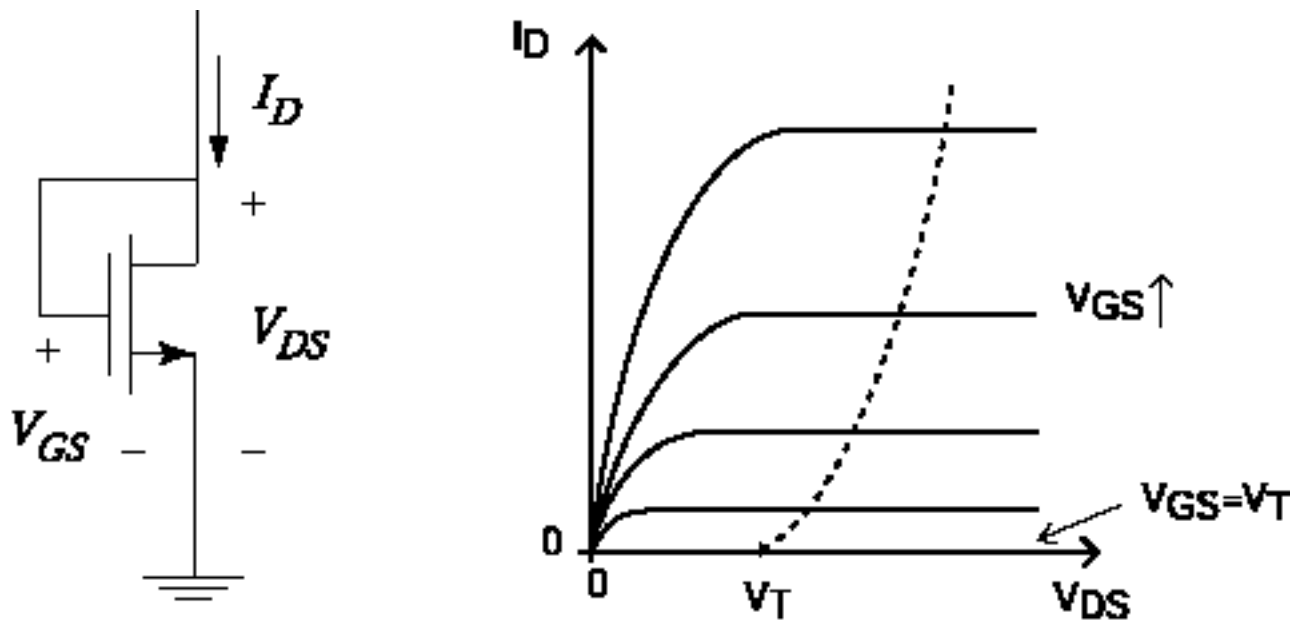
- *Voltage source* easily synthesized from *current source* using MOSFET in diode configuration
- *Current source* easily synthesized from *current source* using **current mirror** circuit.
- Multiple current sources and sinks with different magnitudes can be synthesized from a single current source.
- Voltage and current sources rely on the availability of well “matched” transistors in IC technology.
- In differential amplifiers, signals are represented by *difference* between two voltages

1. DC Voltage Sources

Characteristics of DC Voltage Sources :

- A well controlled output voltage
- Output voltage does not depend on current drawn from source \Rightarrow **Low Thevenin Resistance**

Consider a MOSFET connected in “diode configuration”



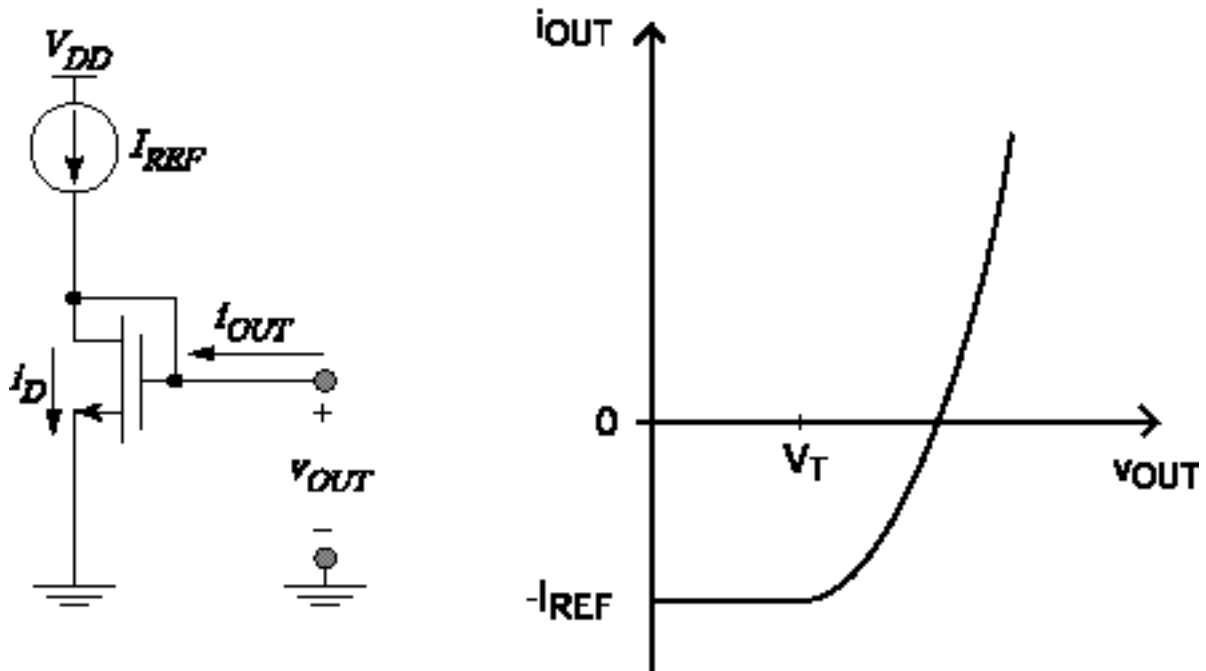
I-V characteristics:

$$I_D = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_{Tn})^2 = \frac{W}{2L} \mu_n C_{ox} (V_{DS} - V_{Tn})^2$$

Beyond the threshold voltage, the MOSFET looks like a “diode” with quadratic I-V characteristics

How does one synthesize a voltage source with this?

Assume a current source is available



$V_{GS} = V_{DS}$ takes a value needed to sink current

$$I_D = I_{REF} + i_{OUT} = \frac{W}{2L} \mu_n C_{ox} (v_{OUT} - V_{Tn})^2$$

Then:

$$i_{OUT} = \frac{W}{2L} \mu_n C_{ox} (v_{OUT} - V_{Tn})^2 - I_{REF}$$

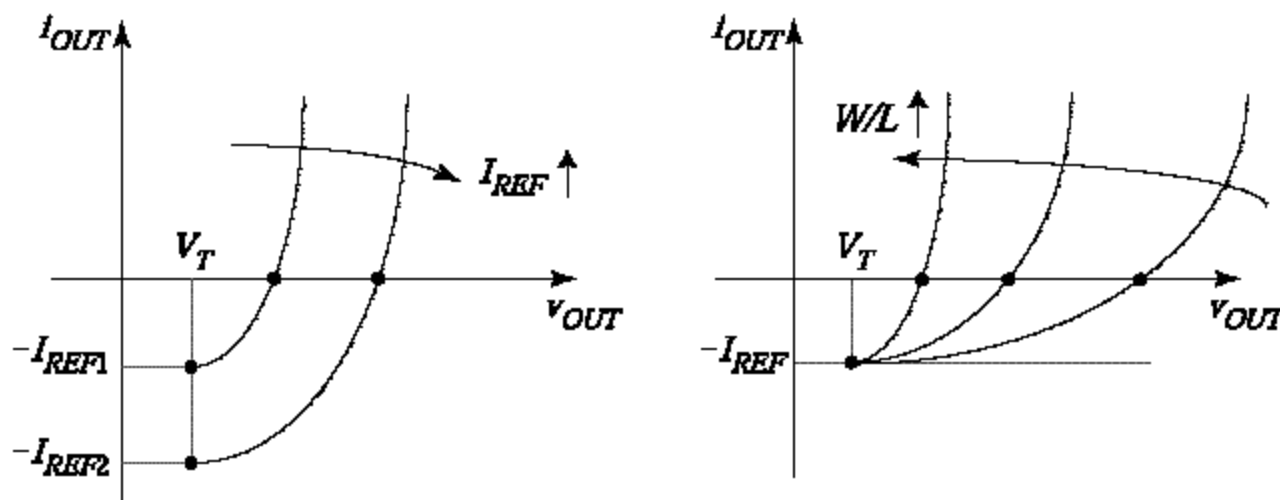
Solving for v_{OUT} :

$$v_{OUT} = V_{Tn} + \sqrt{\frac{I_{REF} + i_{OUT}}{\frac{W}{2L} \mu_n C_{ox}}}$$

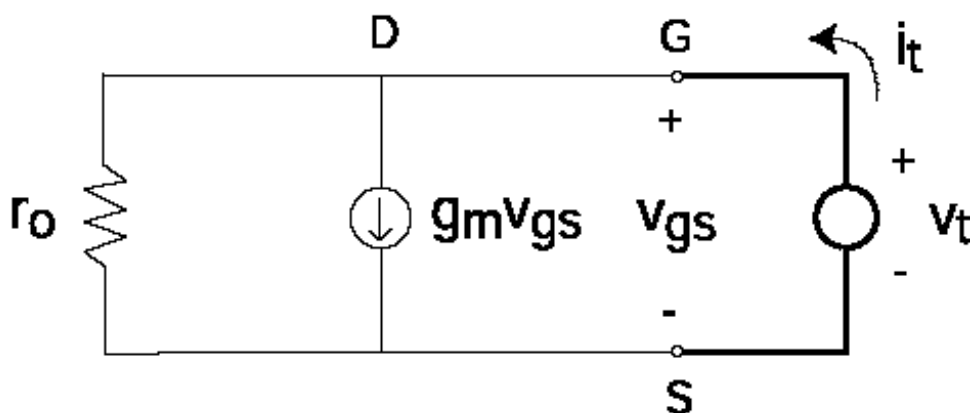
Synthesizing Voltage Sources (contd.)

v_{OUT} is a function of I_{REF} and W/L of MOSFET:

- $I_{REF} \uparrow \Rightarrow v_{OUT} \uparrow$
- $W/L \uparrow \Rightarrow v_{OUT} \downarrow$



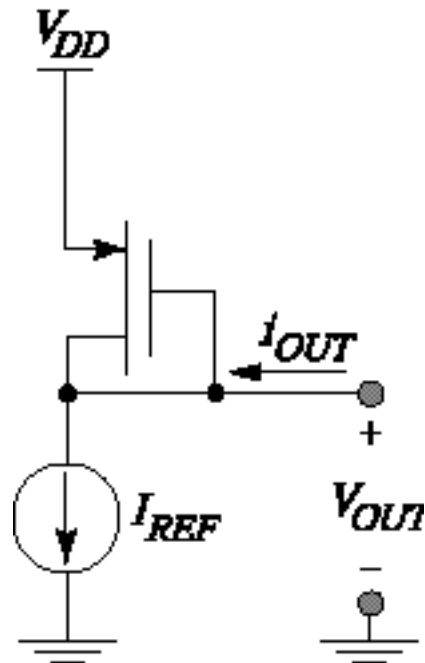
Small Signal Equivalent Circuit Model:



$$R_{out} = \frac{1}{g_m} \parallel r_o \approx \frac{1}{g_m}$$

R_{out} is small (good!)

PMOS voltage source



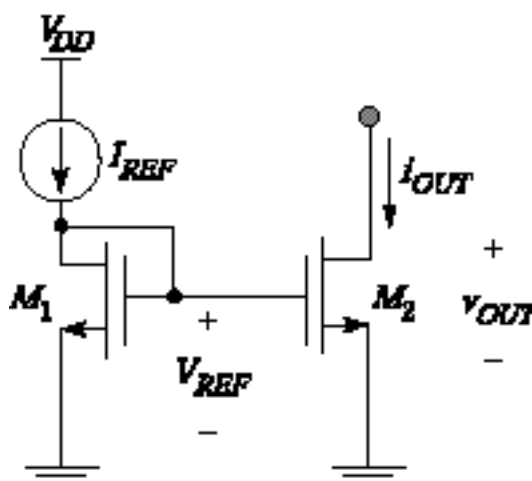
Same operation and characteristics as NMOS voltage source. PMOS needs to be larger to attain the same R_{out} .

3. DC Current Sources and Sinks

Characteristics of Current Sources

- A well controlled output current
- Supplied current does not depend on output voltage
⇒ *High Thevenin Resistance*

Connect a voltage source to the gate of another MOSFET:



$$I_{OUT} \approx \frac{1}{2} \left(\frac{W}{L} \right)_2 \mu_n C_{ox} (V_{REF} - V_{Tn})^2$$
$$I_{REF} \approx \frac{1}{2} \left(\frac{W}{L} \right)_1 \mu_n C_{ox} (V_{REF} - V_{Tn})^2$$

Then:

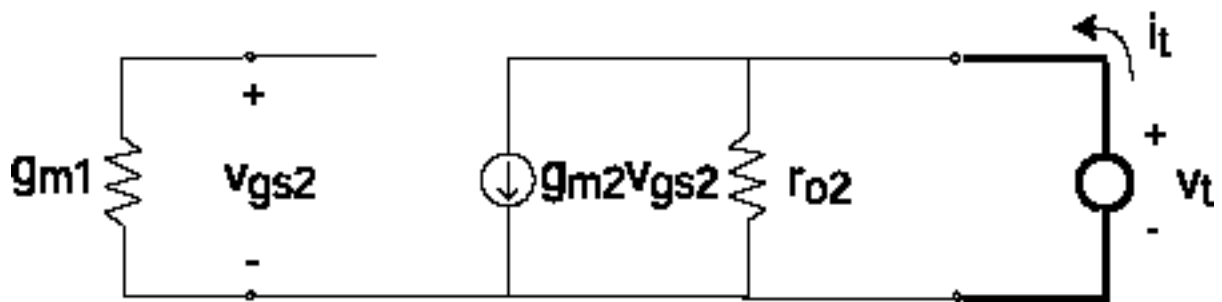
$$I_{OUT} = I_{REF} \frac{\left(\frac{W}{L} \right)_2}{\left(\frac{W}{L} \right)_1}$$

I_{OUT} scales with I_{REF} by W/L ratios of two MOSFETs
⇒ *Current Mirror Circuit*

Well “matched” transistors important.

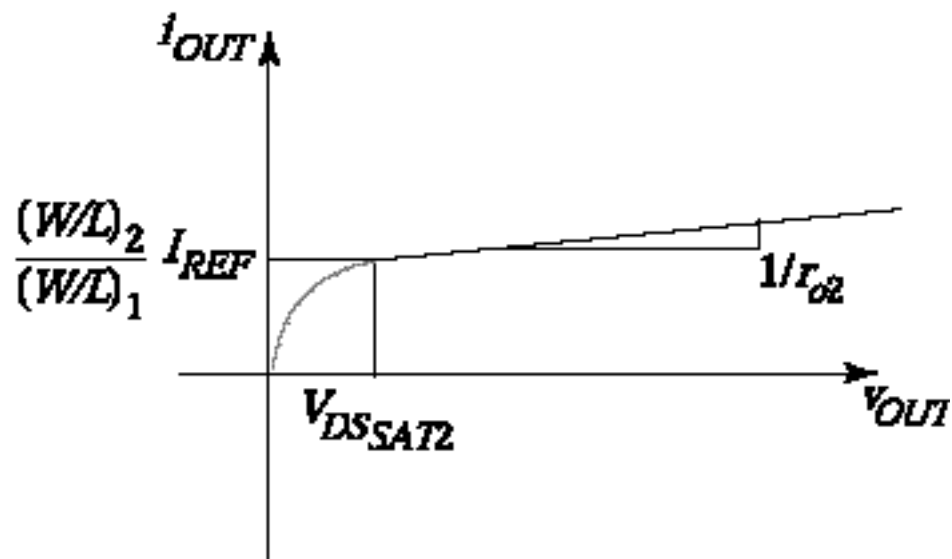
DC Current Sources and Sinks (contd.)

Small Signal Equivalent Circuit Model:



$$R_{out2} = r_{o2}$$

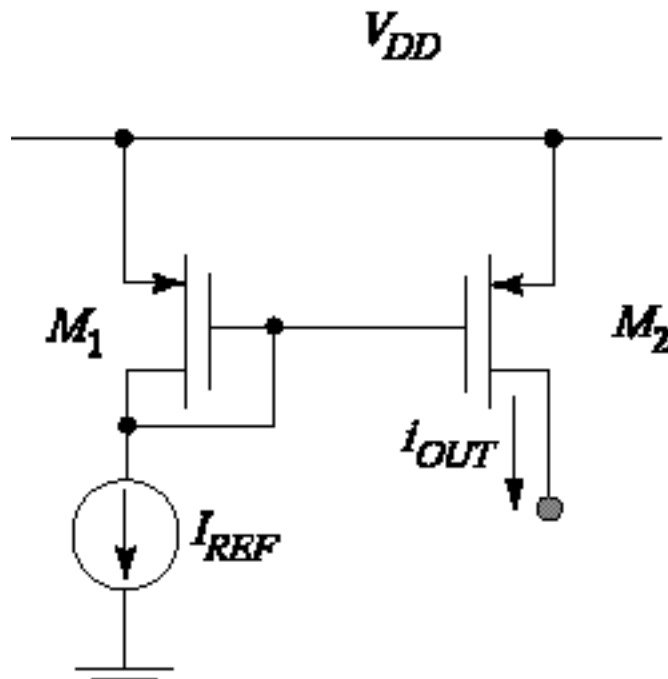
I-V characteristics of NMOS current source:



PMOS Current Source

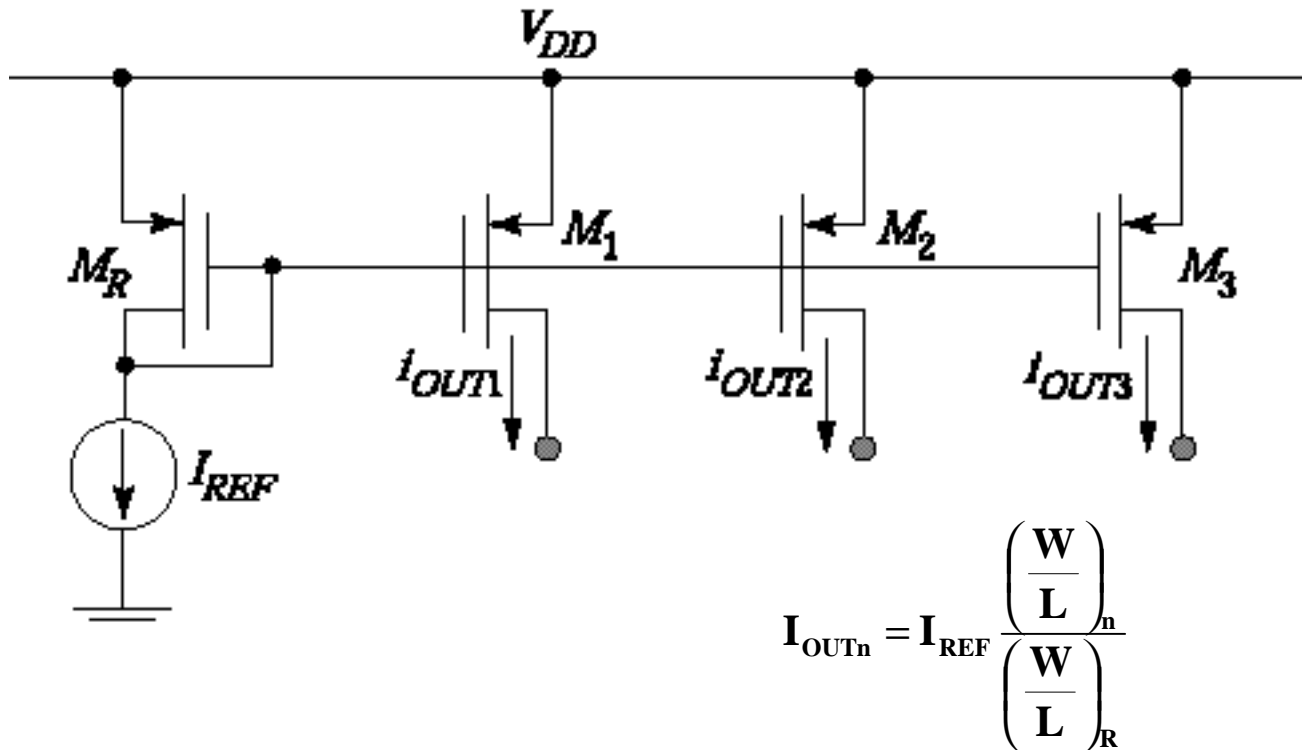
- NMOS current source sinks current to ground
- PMOS current source sources current from positive supply

PMOS Current Mirror:

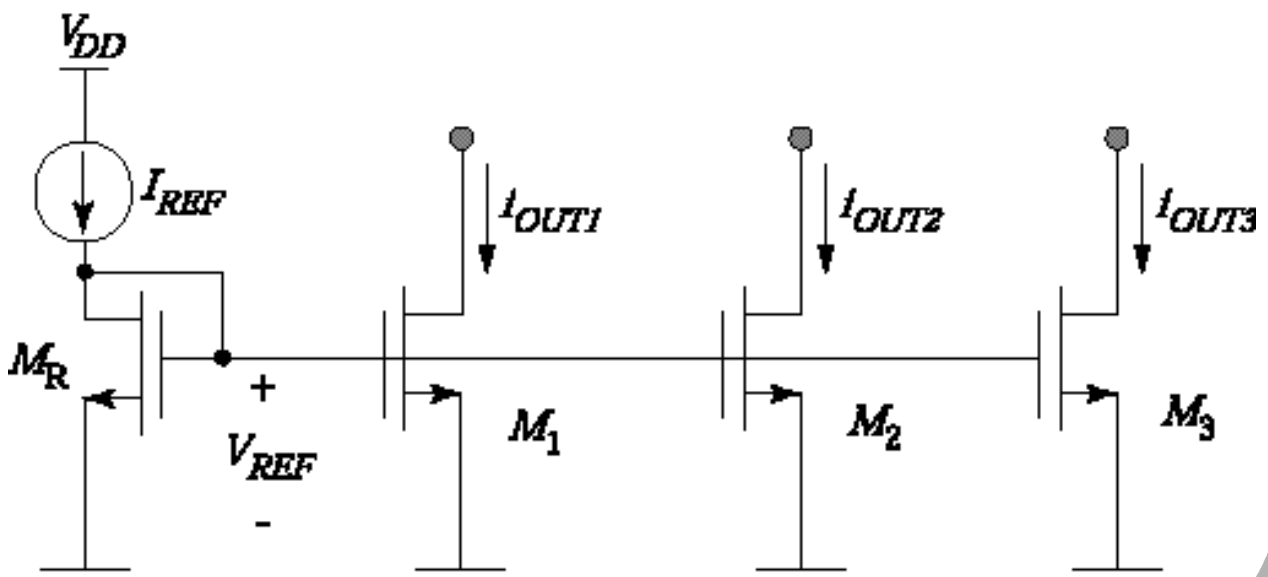


3. Multiple Current Sources

Since there is no DC gate current in MOSFET, we can tie up multiple current mirrors to single current source:

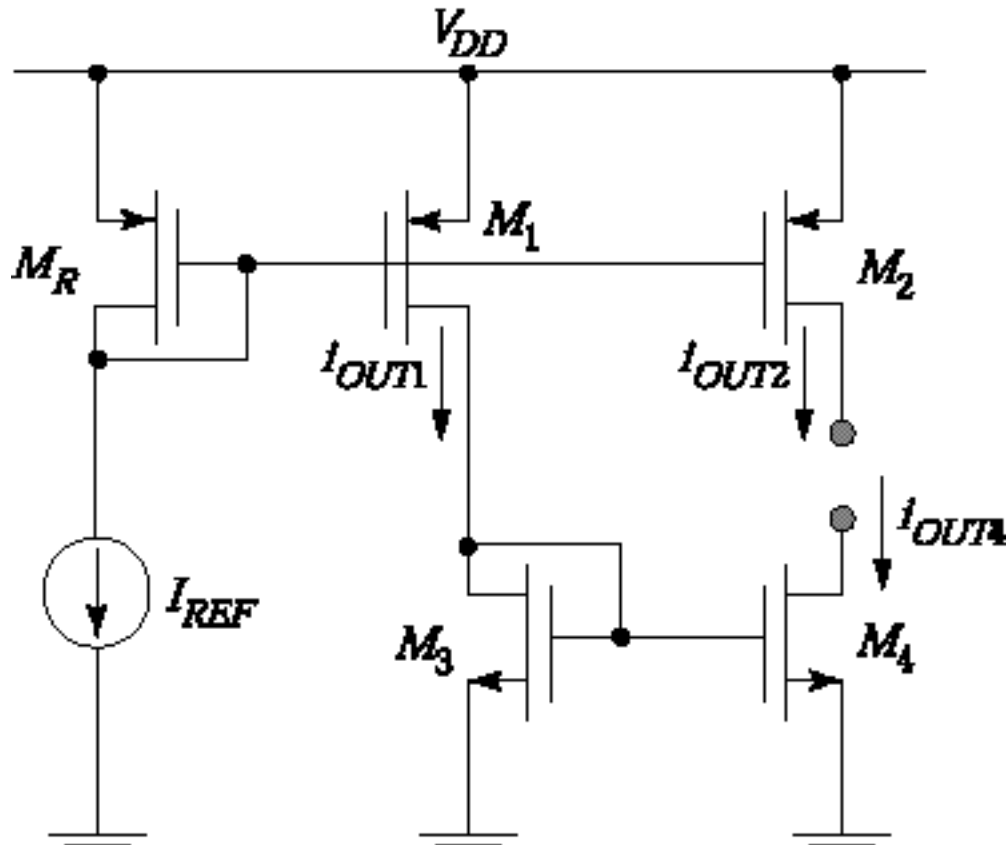


Similar idea with NMOS current sinks:



Multiple Current Sources and Sinks

Often, in a given circuit, we need current sources and sinks. We can build them all out of a single current source.



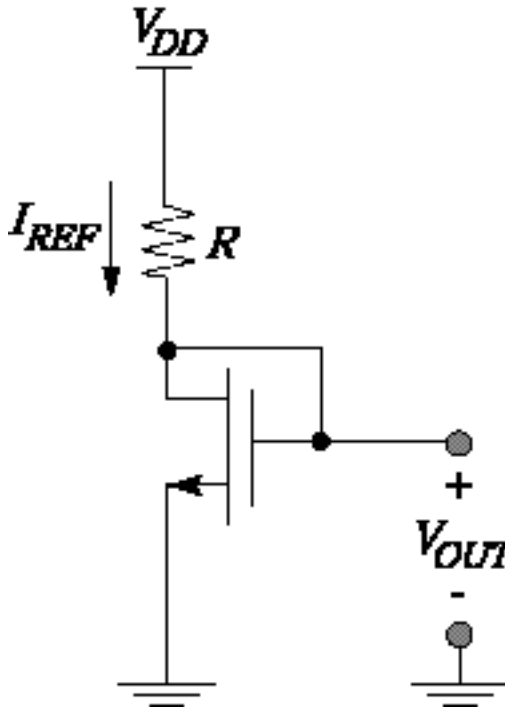
$$I_{OUT1} = I_{REF} \frac{\left(\frac{W}{L}\right)_1}{\left(\frac{W}{L}\right)_R}$$

$$I_{OUT2} = I_{REF} \frac{\left(\frac{W}{L}\right)_2}{\left(\frac{W}{L}\right)_R}$$

$$I_{OUT4} = I_{OUT1} \frac{\left(\frac{W}{L}\right)_4}{\left(\frac{W}{L}\right)_3} = I_{REF} \frac{\left(\frac{W}{L}\right)_4 \left(\frac{W}{L}\right)_1}{\left(\frac{W}{L}\right)_3 \left(\frac{W}{L}\right)_R}$$

Generating I_{REF}

Simple circuit:



$$I_{REF} = \frac{V_{DD} - V_{OUT}}{R}$$

$$V_{OUT} = V_{Tn} + \sqrt{\frac{I_{REF}}{\frac{W}{2L} \mu_n C_{ox}}}$$

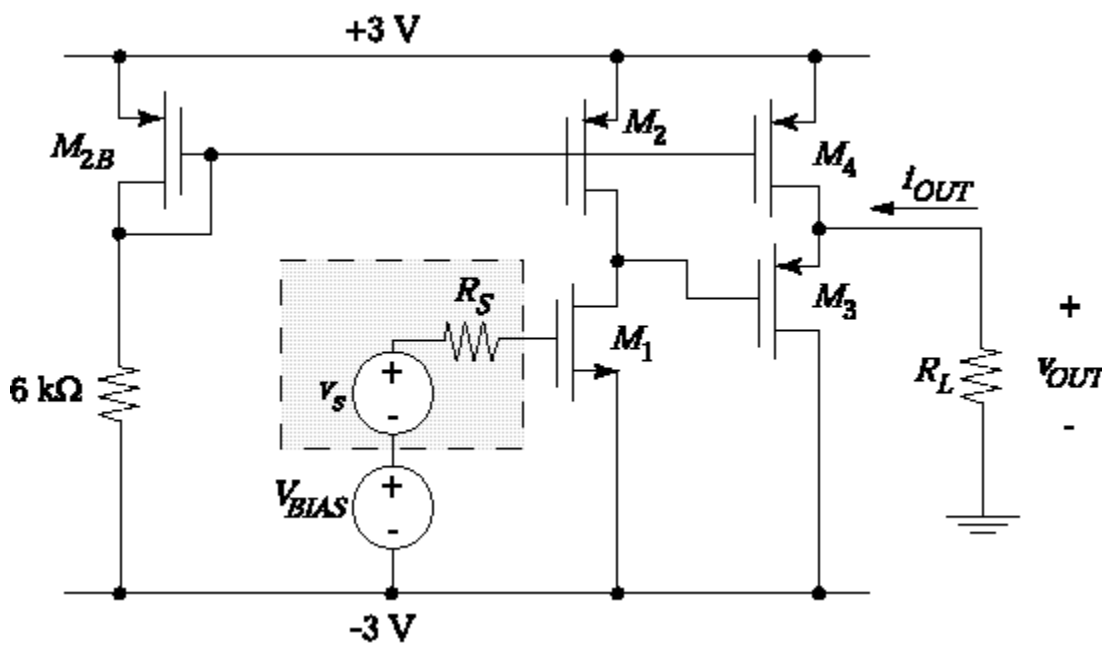
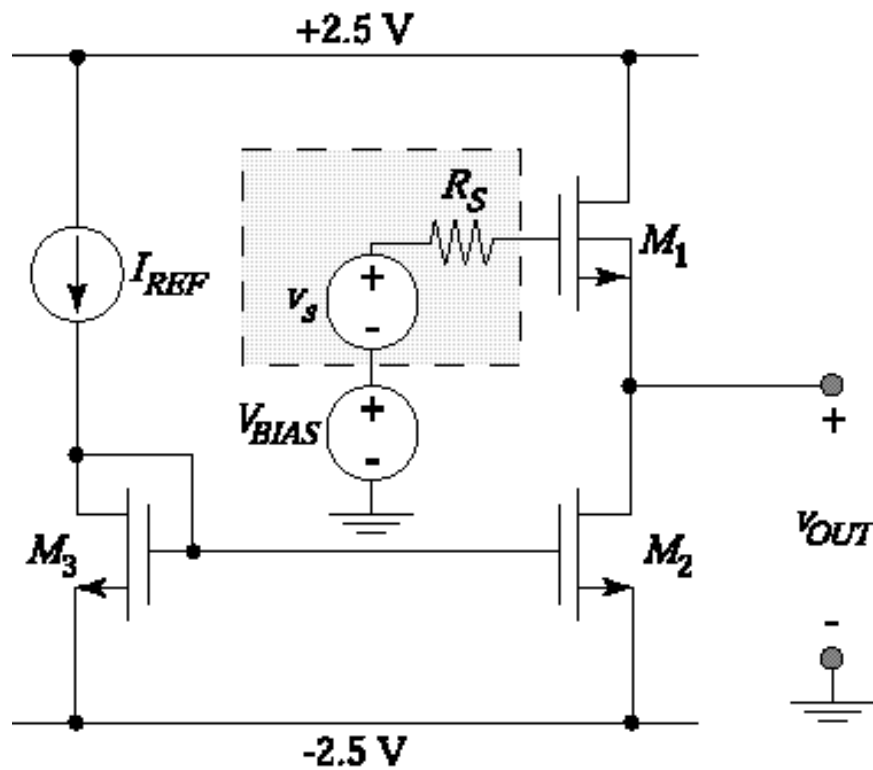
For large W/L :
$$I_{REF} \approx \frac{V_{DD} - V_{Tn}}{R}$$

- Advantages
 - I_{REF} set by value of resistor
- Disadvantages
 - V_{DD} also affects I_{REF} .
 - V_{Tn} and R are functions of temperature $\Rightarrow I_{REF}(T)$.

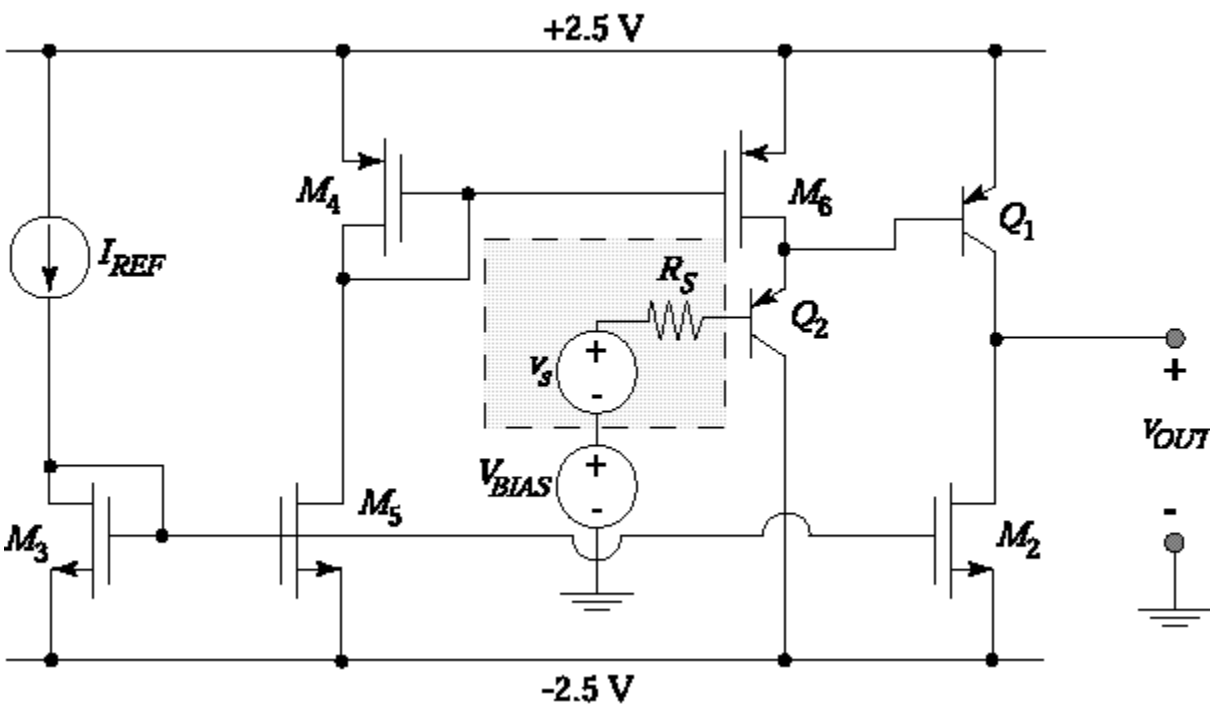
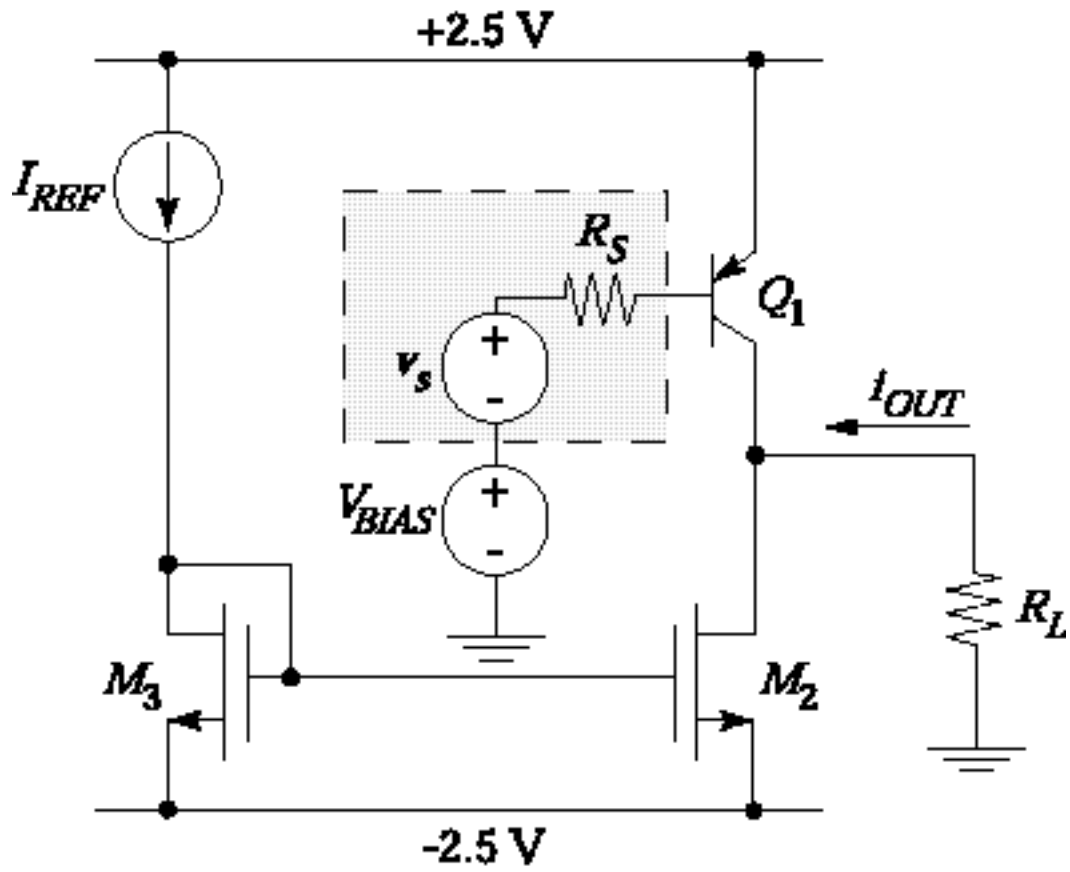
In the real world, more sophisticated circuits are used to generate I_{REF} that are V_{DD} and T independent.

Can now understand more complex circuits?

Examples:



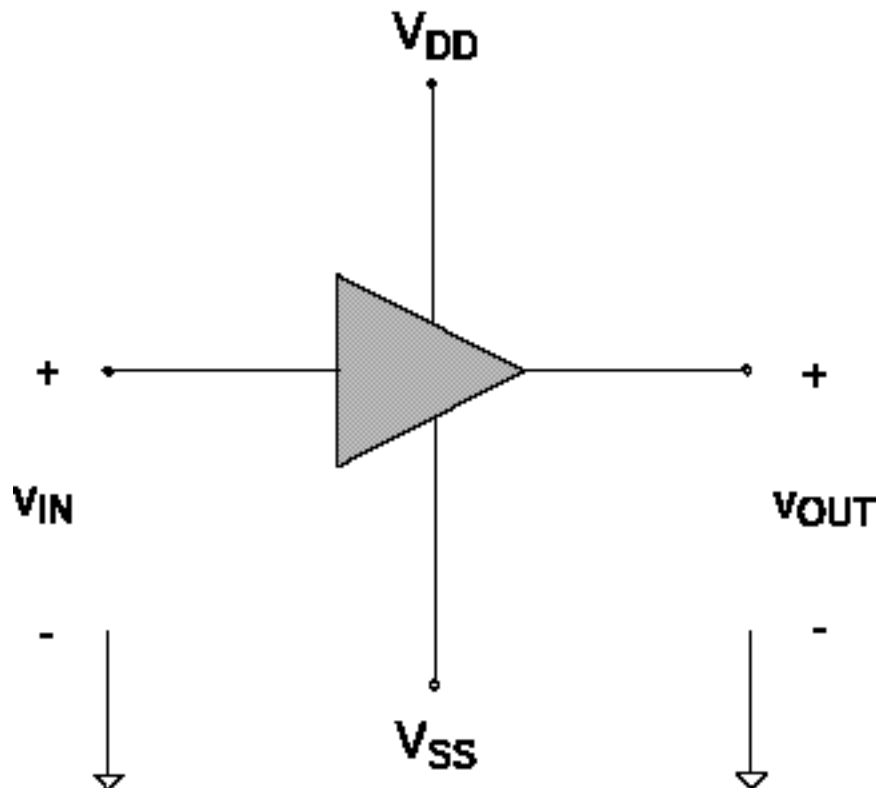
Can now understand more complex circuits?



3. Introduction to Differential Amplifiers

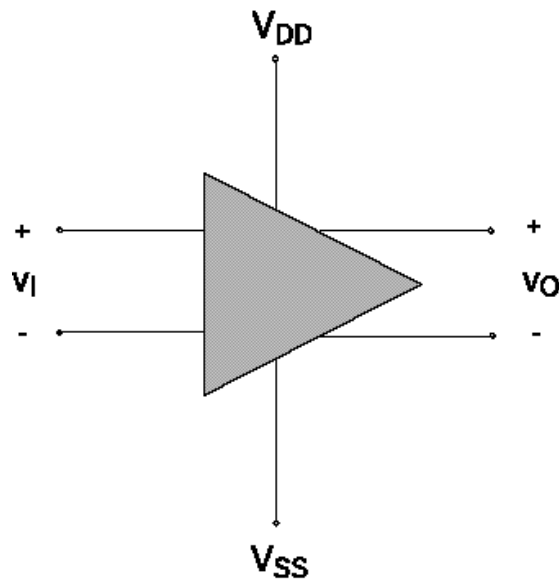
Two problems found in single-transistor amplifier stages are:

- Bias and gain sensitivity to device parameters (μC_{ox} , V_T)
 - Sensitivity can be mitigated but often at a price in terms of performance or cost (gain, power, device area, etc.)
- Vulnerability to ground and power supply noise
 - In dense IC's there is cross-talk, 60 Hz coupling, substrate noise, etc.



Introduction (contd.)

Solution : represent relevant signal by the difference between two voltages

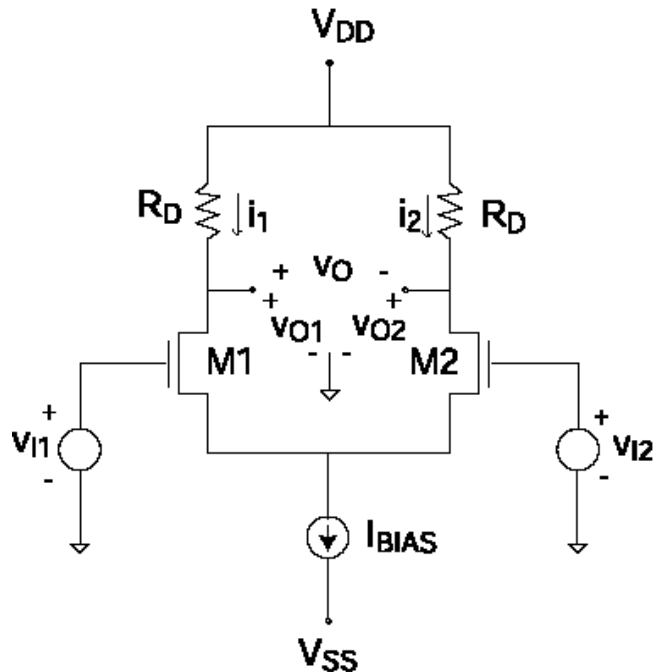


Differential Amplifier:

- Amplifies **difference** between two voltages
- Rejects components **common** to both voltages

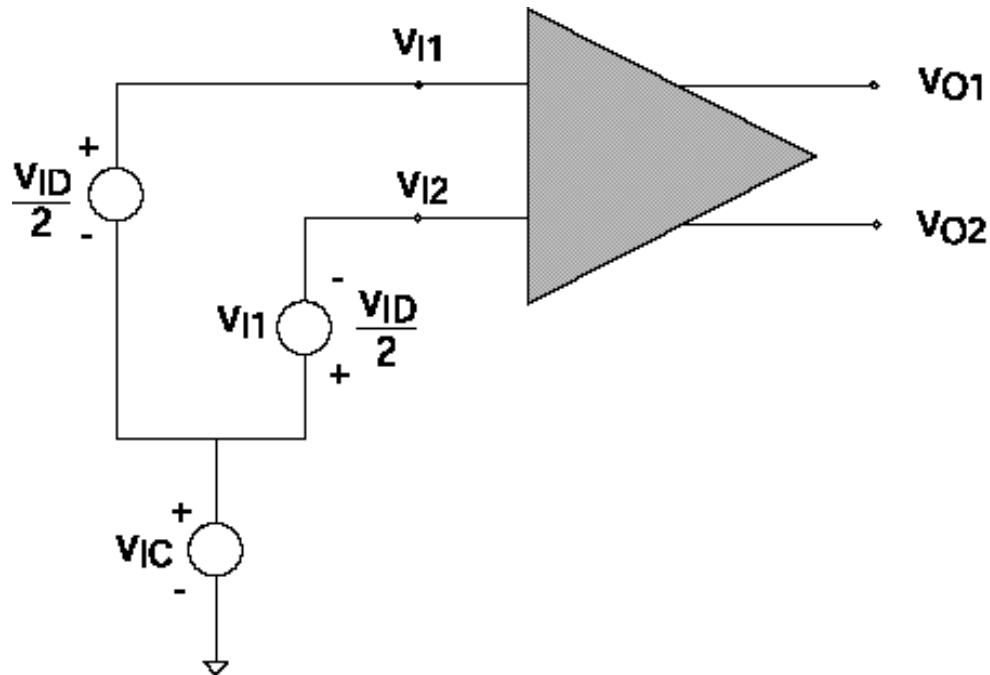
MOSFET Differential Amplifier

Basic Configuration



- v_O responds to difference between v_I 's
 - If $v_{I1} = v_{I2} \Rightarrow$ symmetry $\Rightarrow v_{O1} = v_{O2} \Rightarrow v_O = 0$
 - If $v_{I1} > v_{I2} \Rightarrow$ M1 conducts more than M2 $\Rightarrow i_1 > i_2 \Rightarrow v_{O1} < v_{O2} \Rightarrow v_O < 0$
- v_O insensitive to common mode signals:
 - If both v_{O1} and v_{O2} move in sync, symmetry is preserved $\Rightarrow v_O$ unchanged
 - If ground V_{DD} or V_{SS} have noise, symmetry preserved $\Rightarrow v_O$ unchanged
 - If V_T or μC_{ox} change, symmetry preserved $\Rightarrow v_O$ unchanged.
- Need precise device matching

Differential-mode and Common-mode signals



Distinguish between common-mode and differential-mode:

$$v_{I1} = v_{IC} + \frac{v_{ID}}{2}, \quad v_{I2} = v_{IC} - \frac{v_{ID}}{2}$$

Then:

$$v_{ID} = v_{I1} - v_{I2}, \quad v_{IC} = \frac{v_{I1} + v_{I2}}{2}$$

Similarly at the output:

$$v_{O1} = v_{OC} + \frac{v_{OD}}{2}, \quad v_{O2} = v_{OC} - \frac{v_{OD}}{2}$$

Then:

$$v_{OD} = v_{O1} - v_{O2}, \quad v_{OC} = \frac{v_{O1} + v_{O2}}{2}$$

What did we learn today?

Summary of Key Concepts

- *Voltage source* easily synthesized from *current source* using MOSFET in diode configuration
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- Voltage and current sources rely on the availability of well “matched” transistors in IC technology.
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