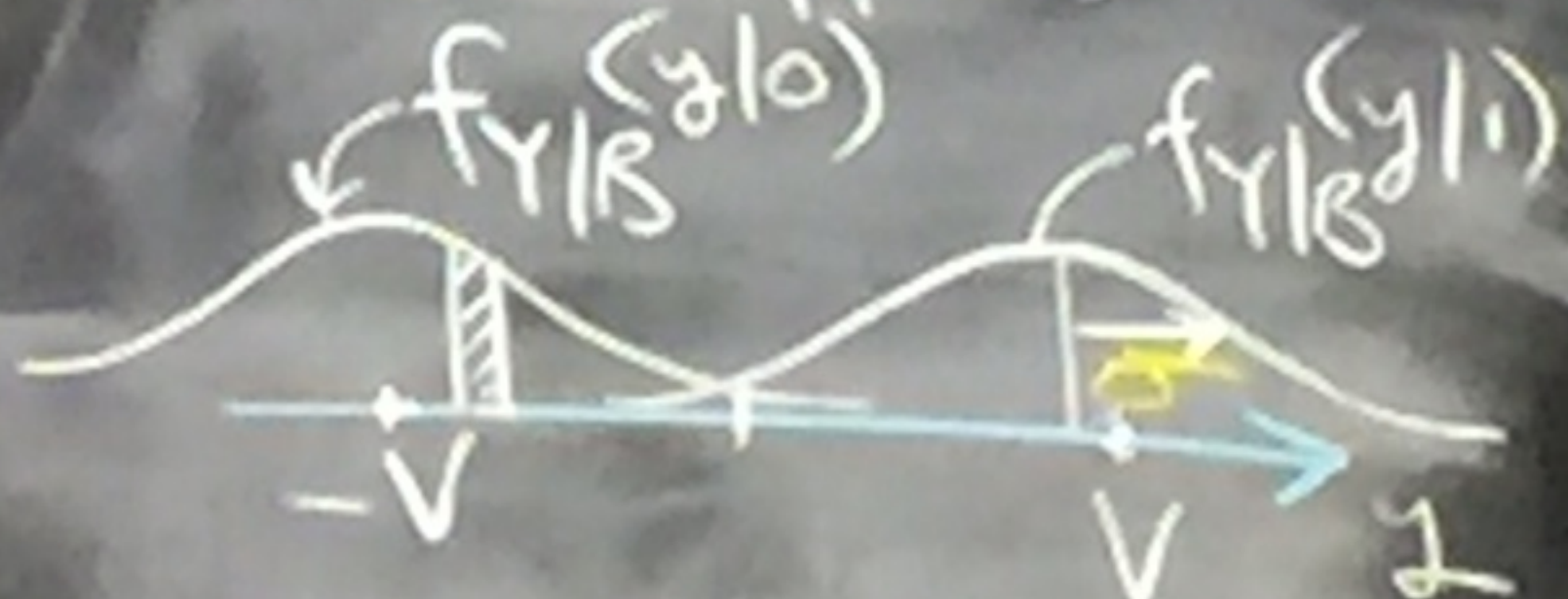
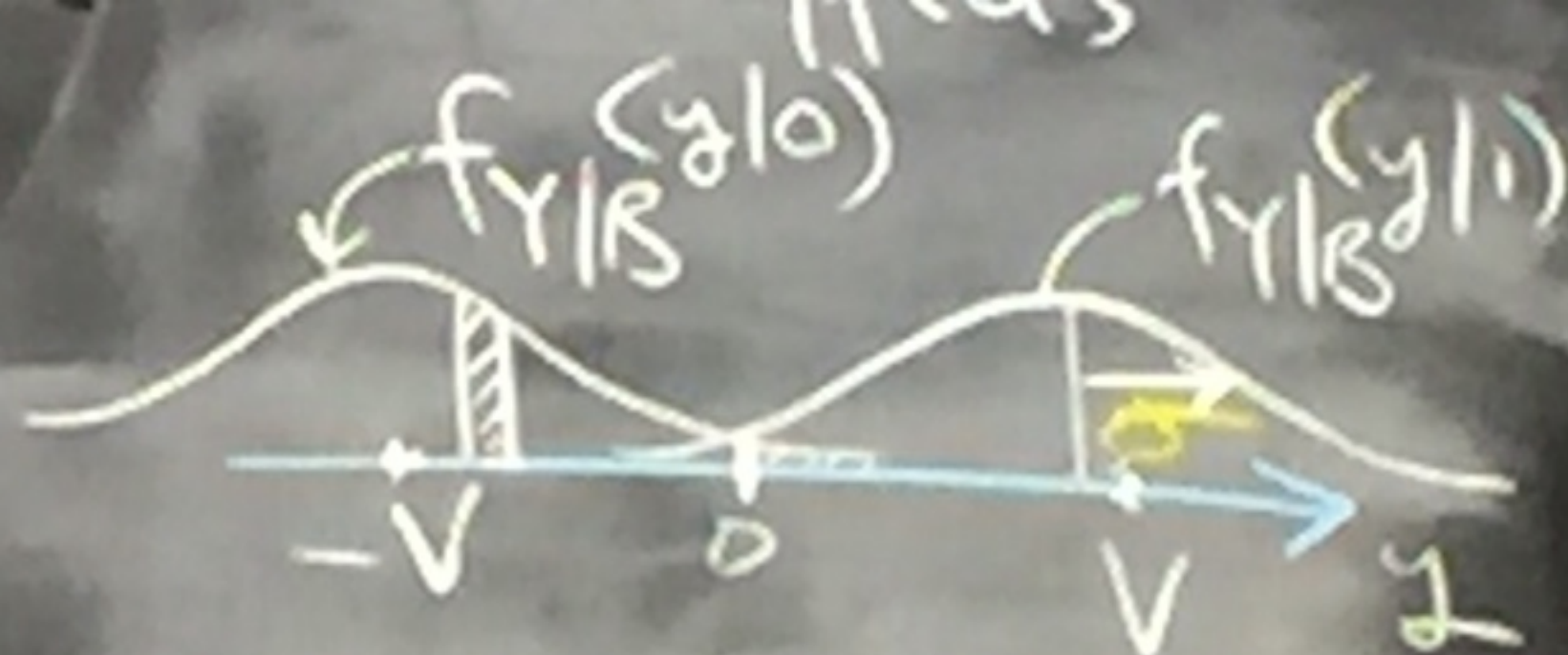
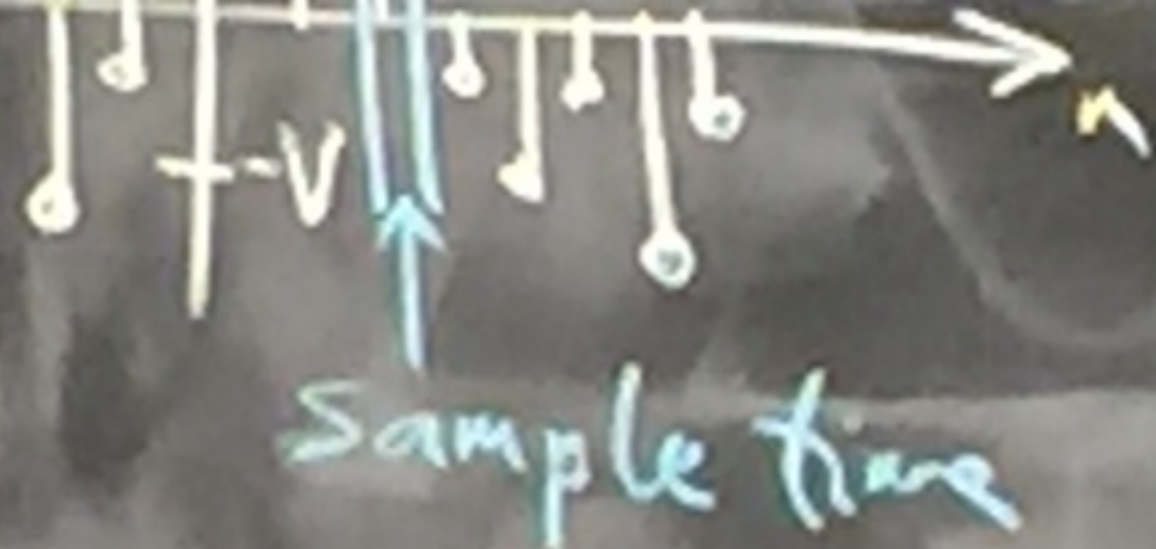


Assume no distortion  
Noise effects



Bit



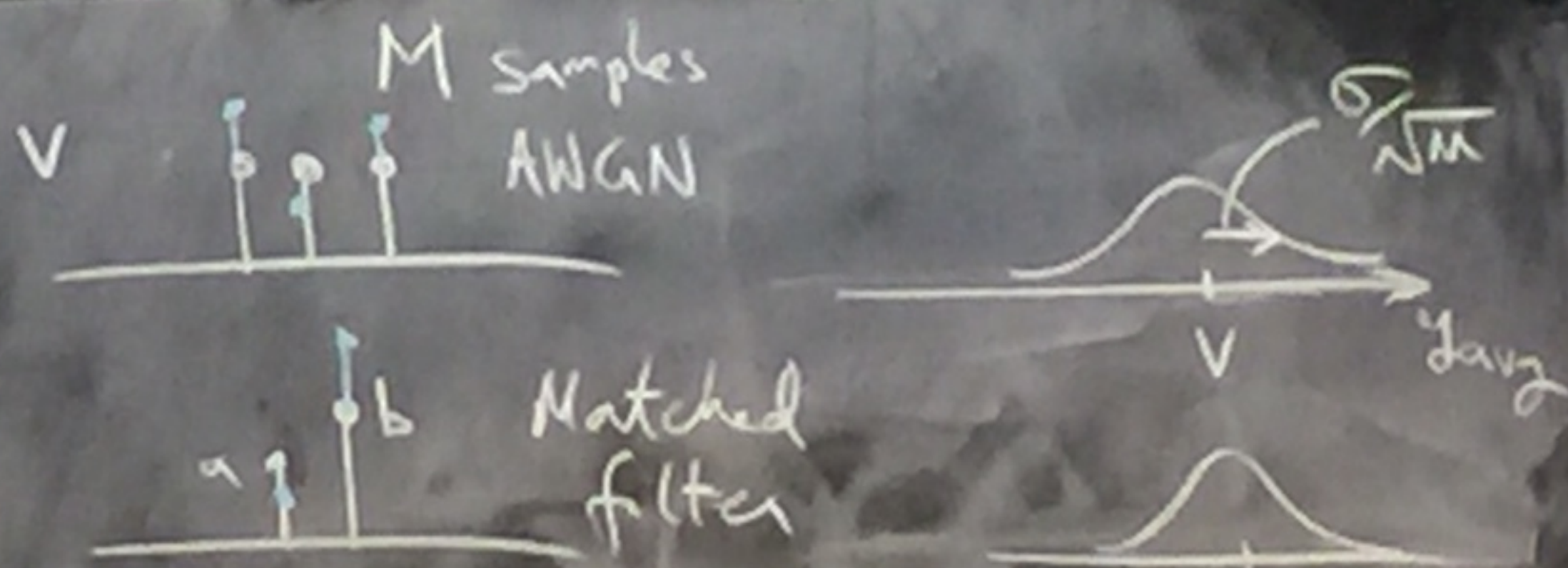
Bit flip prob  $p$  :

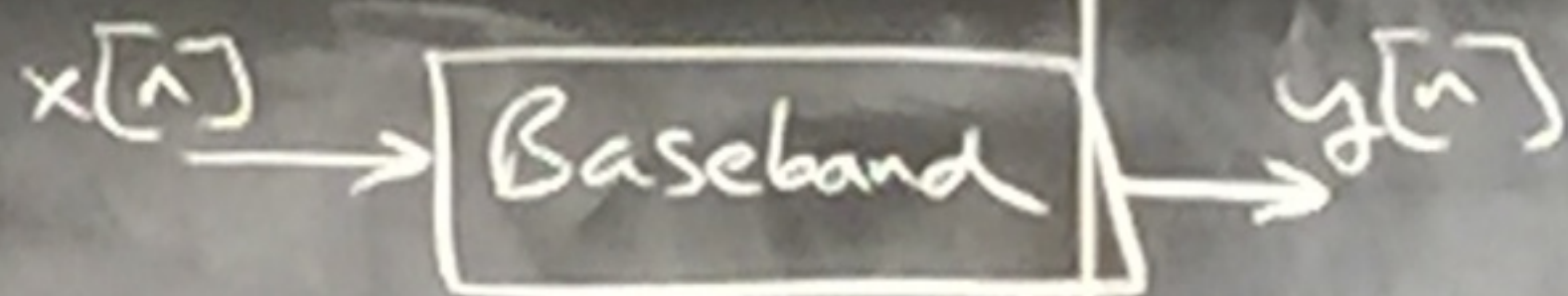
Prob of error :  $P(B=0)P('1'|0) + P(B=1)P('0'|1)$

If 0 and 1 are equally likely

$$P_{err} = \frac{Q\left(\frac{V}{\sigma}\right)}{2}$$

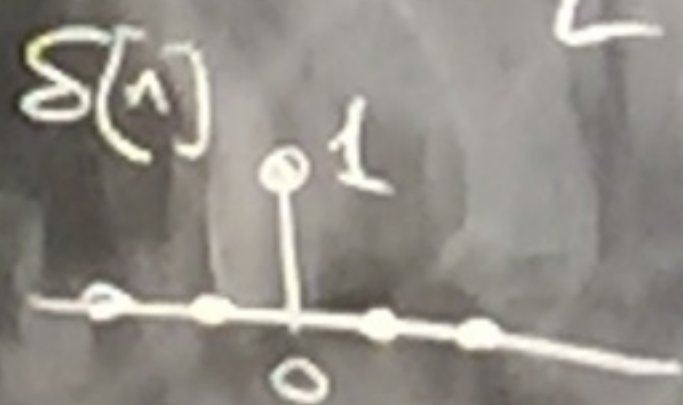
bipolar on-off





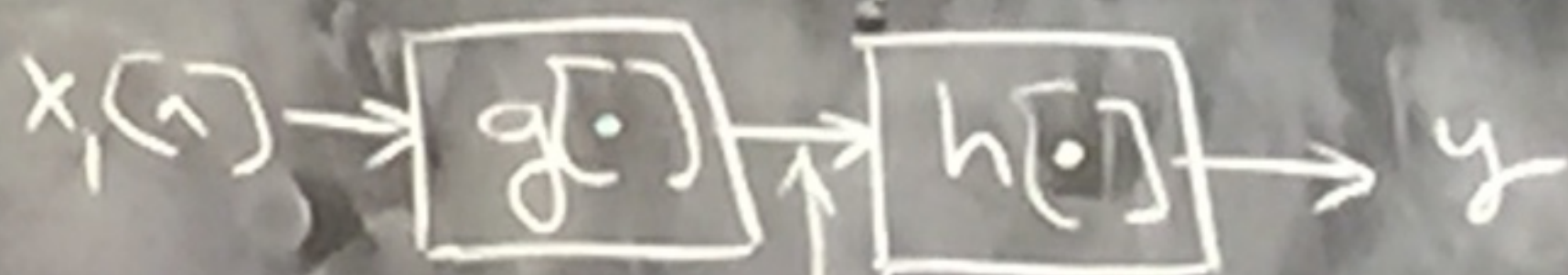
LTI model: 
$$y[n] = \sum_k x[k] h[n-k]$$

where  $h[\cdot]$  is  
the unit sample response



-f 0

Given:  $x_1(n) \rightarrow \boxed{h[\cdot]} \rightarrow 0$  for all  $n$



$\neq 0$   
 $\neq x_1$

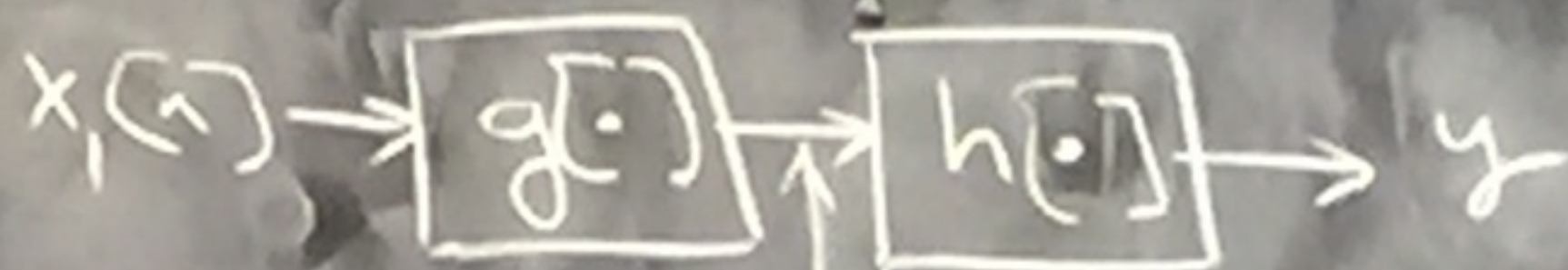
$$y = h * (g * x_1)$$

$$= (h * g) * x_1$$

$$= (a * b) * x_1 = a * (b * x_1)$$



Given:  $x_1[n] \rightarrow h[\cdot] \rightarrow 0$  for all  $n$



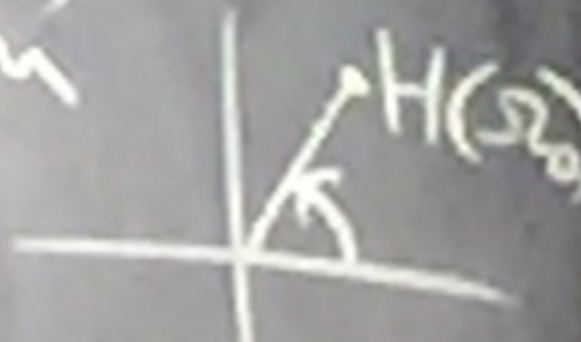
$\neq 0$   
 $\neq x_1$

$$\begin{aligned} y &= h * (g * x_1) \\ &= (h * g) * x_1 \\ &= (g * h) * x_1 = g * (h * x_1) \end{aligned}$$

Sinusoidal inputs  $A_0 \cos(\Omega_0 n + \theta_0) = x[n]$

Freq. resp.  $H(\Omega) = \sum_m h[m] e^{-j\Omega m}$

DTFT



$y[n] = |H(\Omega_0)| A_0 \cos(\Omega_0 n + \theta_0 + \angle H(\Omega_0))$

Filter design: moving average filters  
ideal low pass filters

IFT  $h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(\Omega) e^{j\Omega n} d\Omega$

$$y[n] = |H(\Omega_0)| A_{\text{pos}}(\Omega_0 n + \theta_0 + \angle H(\Omega_0))$$

Filter design

moving average filters

ideal low pass filters

INVERSE  
DTFT

$$h[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} H(\Omega) e^{j\Omega n} d\Omega$$

For any signal of interest in 6.02,  
eg  
abs. summ.

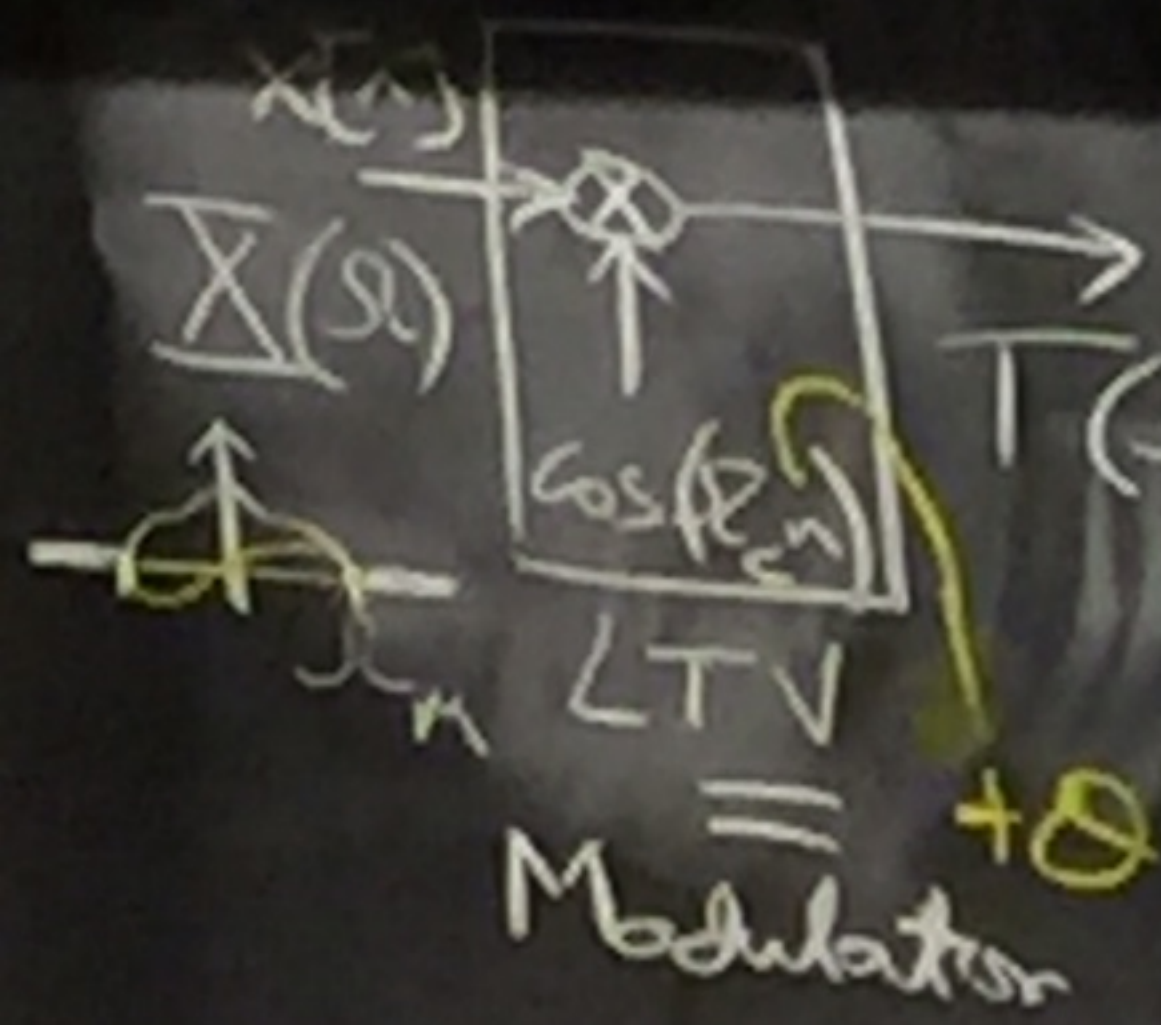
$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \bar{X}(\omega) e^{j\omega n} d\omega$$

where

$$\bar{X}(\omega) = \sum_n x[n] e^{-j\omega n}$$

$$X(\omega) \xrightarrow[\substack{\text{LTI} \\ H(\omega)}]{} Y(\omega) = H(\omega)X(\omega)$$

$$\begin{array}{c} x[n] \\ \hline \uparrow \\ X(\omega) \\ \hline \uparrow \\ \cos \omega n \end{array}$$

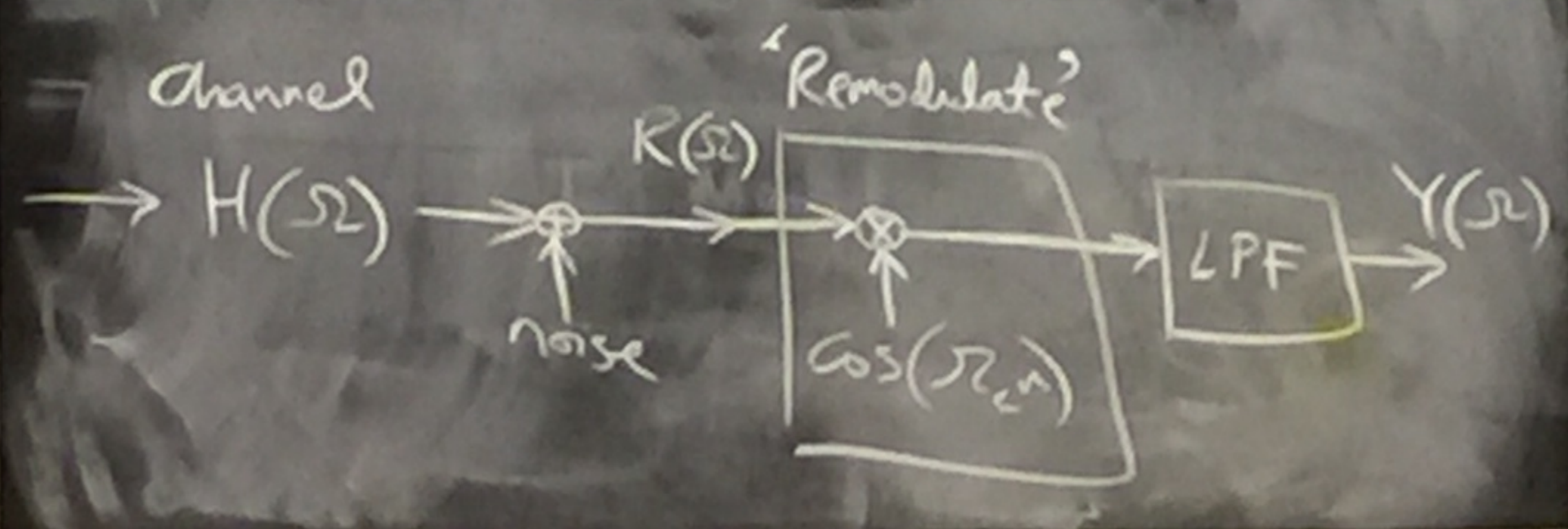


$$T(\Omega) = \frac{1}{2} \left[ X(\Omega - \Omega_c) + X(\Omega + \Omega_c) \right]$$

(Annotations:  $e^{+j\Omega_c t}$  points to  $X(\Omega - \Omega_c)$ ,  $e^{-j\Omega_c t}$  points to  $X(\Omega + \Omega_c)$ )

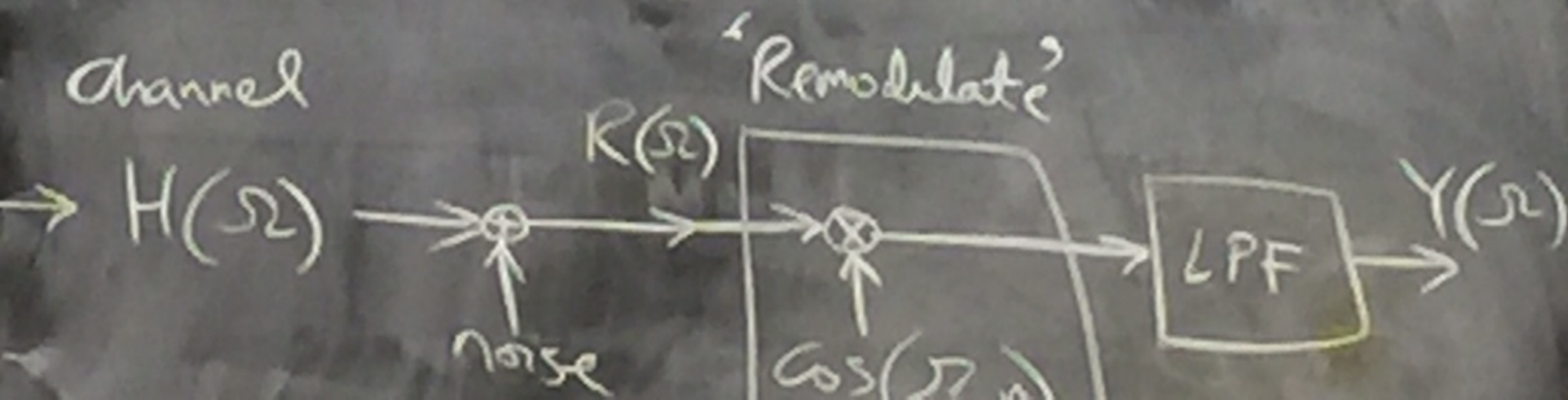


$$2 \left( \cos(\omega_c t) + \sin(\omega_c t) \right)$$



If channel freq. resp = 1 in  
carrier freq. range  
(no distortion)

$$W(\Omega) = \frac{1}{2} [R(\Omega - \Omega_c) + R(\Omega + \Omega_c)]$$

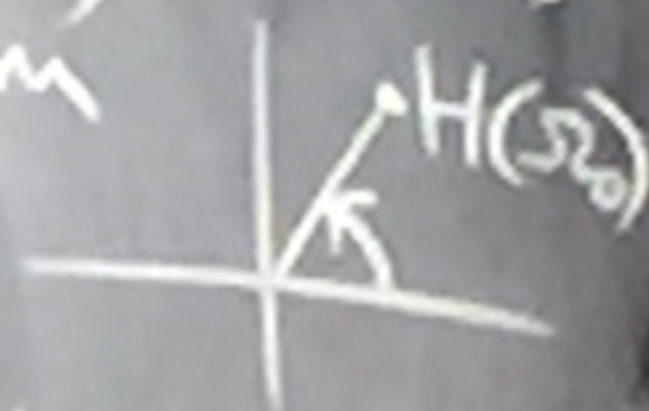




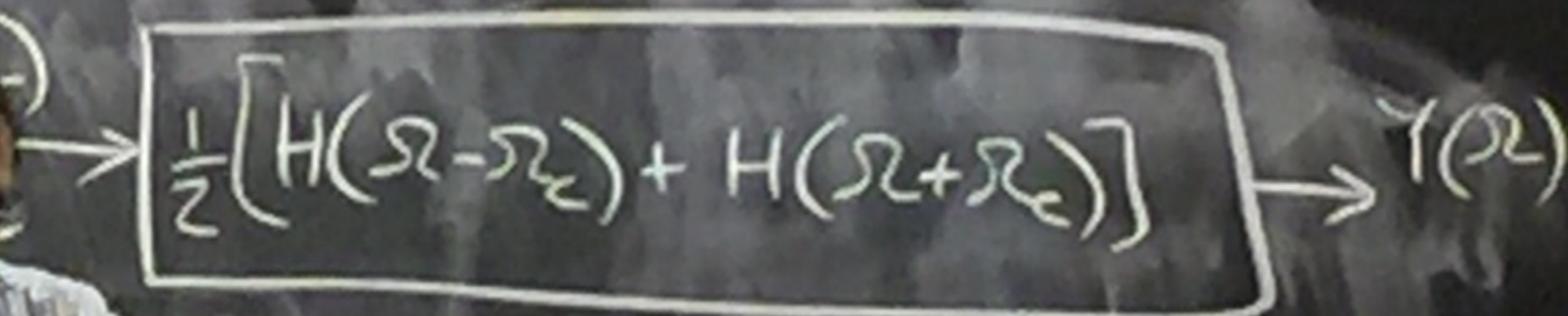
sinusoidal inputs  $A_0 \cos(\Omega_0 n + \theta_0) = x[n]$

Freq. resp  $H(\Omega) = \sum_m h[m] e^{-j\Omega m}$

DTFT



$$y[n] = |H(\Omega_0)| A_0 \cos(\Omega_0 n + \theta_0 + \angle H(\Omega_0))$$



$$\frac{1}{2} [H(\Omega - \Omega_c) + H(\Omega + \Omega_c)] \rightarrow Y(\Omega)$$

Baseband