
introduction to ebcs it
DIGITAL COMMUNICATION systems

### 6.02 Fall 2011 Lecture \#8

- State machines \& trellises (recap)

Path and branch metrics

- Viterbi decoding of convolutional codes
- Hard decision vs. soft decision decoding
- Puncturing, free distance, and performance


## Using Convolutional Codes

- Transmitter
- Beginning at starting state, processes message bit-by-bit
- For each message bit: makes a state transition, sends parity bits
- Receiver
- Doesn't have direct knowledge of transmitter's state transitions; only knows (possibly corrupted) received parity bits
- Must find most likely sequence of transmitter states that could have generated the received parity bits, $p_{i}$
- If BER is $<1 / 2$, then
- Most likely message sequence is the one that generated the sequence of parity bits with the smallest Hamming distance from the actual received $p_{i}$


## State Machine View



The state machine is the same for all $K=3$ codes. Only the $p_{i}$ labels change depending on number and values for the generator polynomials

- Example: $\mathrm{K}=3$, rate $1 / 2$ convolutional code
- States labeled with $x[n-1]$ x[n-2]
- Arcs labeled with $x[n] / p_{0} p_{1}$
- $\mathrm{msg}=101100 ;$ xmit = 111101000110
.02 Fall 2011 Lecture 8 , Slide \#2


## Example

- Using $K=3$, rate $1 / 2$ code from earlier slides
- Received

11101100011000

- Some errors have occurred...
What's the 4-bit message?
- Look for message whose xmit bits are closest to rcvd bits Most likely: 1011

| Msg | Xmit* | Rcvd | ${ }^{\text {d }}$ |
| :---: | :---: | :---: | :---: |
| 0000 | 000000000000 | 111011000110 | 7 |
| 0001 | 000000111110 |  | 8 |
| 0010 | 000011111000 |  | 8 |
| 0011 | 000011010110 |  | 4 |
| 0100 | 001111100000 |  | 6 |
| 0101 | 001111011110 |  | 5 |
| 0110 | 001101001000 |  | 7 |
| 0111 | 001100100110 |  | 6 |
| 1000 | 111110000000 |  | 4 |
| 1001 | 111110111110 |  | 5 |
| 1010 | 111101111000 |  | 7 |
| 1011 | 111101000110 |  | 2 |
| 1100 | 110001100000 |  | 5 |
| 1101 | 110001011110 |  | 4 |
| 1110 | 110010011000 |  | 6 |
| 1111 | 110010100110 |  | 3 |
| Msg pa | ded with 2 zeros | ore transmission |  |

## Trellis View @ Transmitter

$\mathrm{x}[\mathrm{n}]$

00
$\qquad$ $\square$ $10-$ ${ }^{0} 1 / 10$

11
$x[n-1 \mid x[n-2]$


## Hard-decision Branch Metric

- $\mathrm{BM}=$ Hamming distance
between expected parity bits and received parity bits
- Compute BM for each transition arc in trellis
- Example: received parity $=00$
- $\operatorname{BM}(00,00)=0$ $\operatorname{BM}(01,00)=1$ $\mathrm{BM}(10,00)=1$
$\mathrm{BM}(11,00)=2$
- Will be used in computing PM $[\mathrm{s}, \mathrm{i}+1]$ from $\mathrm{PM}[\mathrm{s}, \mathrm{i}]$.
- We want to use the most likely BM, which, means minimum BM.



## Viterbi Algorithm

- Want: Most likely message sequence
- Have: (possibly corrupted) received parity sequences
- Viterbi algorithm for a given $K$ and $r$ :
- Works incrementally to compute most likely message sequence
- Uses two metrics
- Branch metric: $\mathrm{BM}(\mathrm{xmit}, \mathrm{rcvd})$ proportional to likelihood that transmitter sent $x$ mit given that we' ve received rcvd.
- "Hard decision": use digitized bits, compute Hamming distance between xmit and rcvd. Smaller distance is more likely if BER < 1/2
- "Soft decision": use function of received voltages directly
- Path metric: $\mathrm{PM}[\mathrm{s}, \mathrm{i}]$ for each state $s$ of the $2^{\mathrm{K}-1}$ transmitter states and bit time $i$ where $0 \leq \mathrm{i}<l$ len(message).
- PM[s,i] = most likely sum of $\mathrm{BM}\left(\mathrm{xmit}_{\mathrm{m}}\right.$, received parity) over all message sequences $m$ that place transmitter in state $s$ at time $i$ - PM[s,i+1] computed from PM[s,i] and $\mathrm{p}_{0}[\mathrm{i}], \ldots, \mathrm{p}_{\mathrm{r}-1}[\mathrm{i}]$


## Computing PM[s,i+1]

Starting point: we've computed
$\mathrm{PM}[\mathrm{s}, \mathrm{i}]$, shown graphically as label in trellis box for each state at time $i$.


Q: What's the most likely state $s$ for the transmitter at time $i$ ?
A: state 00 (smallest PM[s,i])

## Computing PM[s,i+1] cont' d.

Q: If the transmitter is in state $s$ at time i+1, what state(s) could it have been in at time i?

the transmitter in state $s$ at time $i+1$ must have left the transmitter in state $\alpha$ or state $\beta$ at time $i$.
A: For each state s, there are two predecessor states $\alpha$ and $\beta$ in the trellis diagram

Example: for state $01, \alpha=10$ and $\beta=11$.
Any message sequence that leaves

## Computing PM[s,i+1] cont' d.

Formalizing the computation:
$\mathrm{PM}[\mathrm{s}, \mathrm{i}+1]=\min (\mathrm{PM}[\alpha, \mathrm{i}]+\mathrm{BM}[\alpha \rightarrow \mathrm{s}]$, $\operatorname{PM}[\beta, \mathrm{i}]+\mathrm{BM}[\beta \rightarrow \mathrm{s}])$

Example:
$\mathrm{PM}[01, \mathrm{i}+1]=\min (\mathrm{PM}[10, \mathrm{i}]+2$,

$$
\operatorname{PM}[11, \mathrm{i}]+1)
$$

$=\min (3+2,2+1)=3$
Notes:

1) Remember which arc was min; saved arcs will form a path through trellis
2) If both arcs have same sum, break tie arbitrarily (e.g., when computing PM $[11, i+1])$


## Computing PM $[\mathrm{s}, \mathbf{i}+1]$ cont' d .

Example cont' d: to arrive in state 01 at time i+1, either
1)The transmitter was in state 10 at time $i$ and the $i^{\text {th }}$ message bit was a 0 . If that's the case, the transmitter sent 11 as the parity bits and there were 2 bit errors since we received 00. Total bit errors $=\mathrm{PM}[10, \mathrm{i}]+2=5$ OR
2)The transmitter was in state 11 at time $i$ and the $i^{\text {th }}$ message bit was a 0 . If that's the case, the transmitter sent 01 as the parity bits and there was 1 bit error since we received 00 .
Total bit errors = PM[11,i] + $1=3$
Which is more likely?


Finding the Maximum-Likelihood Path

| Rcvd: | 11 | 10 | 11 | 00 | 01 | 10 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |



- Path metric: number of errors on maximum-likelihood path to given state (min of all paths leading to state)
- Branch metric: for each arrow, the Hamming distance between received parity and expected parity


## Viterbi Algorithm



- Compute branch metrics for next set of parity bits
- Compute path metric for next column
- add branch metric to path metric for old state
- compare sums for paths arriving at new state
- select path with smallest value (fewest errors, most likely)


## Survivor Paths

$\begin{array}{lllllll}\text { Rcvd: } & 11 & 10 & 11 & 00 & 01 & 10\end{array}$
00
01


- Notice that some paths don' t continue past a certain state
- Will not participate in finding most-likely path: eliminate
- Remaining paths are called survivor paths
- When there's only one path: we' ve got a message bit!


## Example (cont'd.)



- After receiving 3 pairs of parity bits we can see that all ending states are equally likely
- Power of convolutional code: use future information to constrain choices about most likely events in the past


## Example (cont'd.)

$\begin{array}{lllllll}\text { Rcvd: } & 11 & 10 & 11 & 00 & 01 & 10\end{array}$


- When there are "ties" (sum of metrics are the same)
- Make an arbitrary choice about incoming path
- If state is not on most-likely path: choice doesn't matter
- If state is on most-likely path: choice may matter and error correction has failed (mark state with underline to tell)


## Example (cont' d.)



- When we reach end of received parity bits
- Each state's path metric indicates how many errors have happened on most-likely path to state
- Most-likely final state has smallest path metric
- Ties means end of message uncertain (but survivor paths may merge to a unique path earlier in message


## Viterbi Algorithm with Hard Decisions

- Branch metrics measure the likelihood by comparing receive parity bits to possible transmitted parity bits computed from possible messages.
- Path metric PM[s,i] proportional to likelihood of transmitter being in state s at time $i$, assuming the mostly likely message of length i that leaves the transmitter in state $s$
- Most likely message? The one that produces the most likely PM[s,N].
- At any given time there are $2^{\mathrm{K}-1}$ most-likely messages we' re tracking $\rightarrow$ time complexity of algorithm grows exponentially with constraint length K


## Traceback



- Use most-likely path to determine message bits
- Trace back through path: message in reverse order
- Message bit determined by high-order bit of each state (remember that came from message bit when encoding)
- Message in example: 101100 (w/ 2 transmission errors)


## Hard Decisions

- As we receive each bit it's immediately digitized to " 0 " or " 1 " by comparing it against a threshold voltage
- We lose the information about how "good" the bit is:
a " 1 " at .9999 V is treated the same as a " 1 " at .5001 V
- The branch metric used in the Viterbi decoder is the Hamming distance between the digitized received voltages and the expected parity bits
- This is called hard-decision decoding
- Throwing away information is (almost) never a good idea when making decisions
- Can we come up with a better branch metric that uses more information about the received voltages?


## Soft Decision Decoding

- In practice, the receiver gets a voltage level, V , for each received parity bit
- Sender sends V0 or V1 volts; V in $(-\infty, \infty)$ assuming additive Gaussian noise
- Idea: Pass received voltages to decoder before digitizing
- Define a "soft" branch metric as the square of the Euclidian distance between received voltages and expected voltages

- Soft-decision decoder chooses path that minimizes sum of the squares of the Euclidean distances between received and expected voltages
.02 Fall201- Different BM \& PM values, but otherwise the same algorithm 8, slide ${ }^{2}$


## Free Distance of a Convolutional Code

| $x[n]$ | 0 | 0 | 0 | 0 | 0 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

00


11

$$
\longrightarrow \text { time }
$$

The free distance is the difference in path metrics between the output when the input is all zeroes, and the output the first input bit along being a ' 1 '. In this example, it is because the first transition outputs ' 11 ', the second outputs ' 11 ', and the third ' 10 at which time it converges to the correct state.

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## Performance of Viterbi Decoding

- Complexity is linear in message length and exponential in $K$, the constraint length
- Code rate: $1 / r$
- How to get higher rates or other rates?
- Answer: Puncturing
- How much error correcting capability do we get from a convolutional code?
- In general, larger values of $K$ and $r$ (the number of parity streams or generators) provide higher error tolerance
- But what determines the error correction ability? (I.e., what's the equivalent of the Hamming distance?)
- Answer: Free distance


