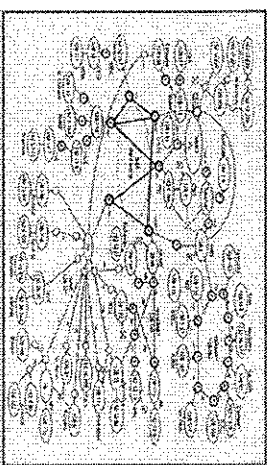
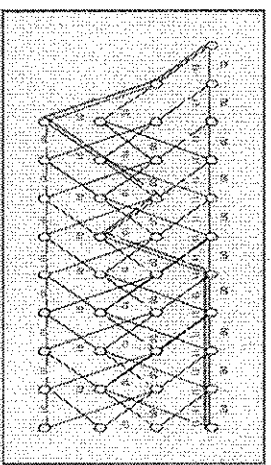
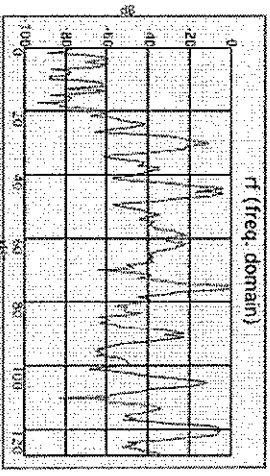
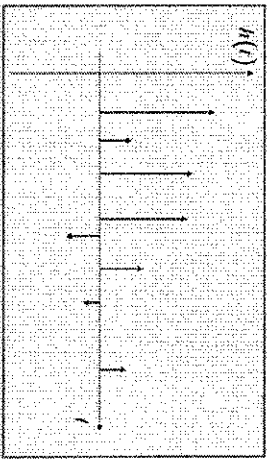


①



INTRODUCTION TO BECS II
**DIGITAL
COMMUNICATION
SYSTEMS**

**6.02 Spring 2009
Lecture #11**

- Summary of progress
- The problem of multiplexing
- Eternal Signals and LTI Systems
- Discrete-Time Sines and Cosines

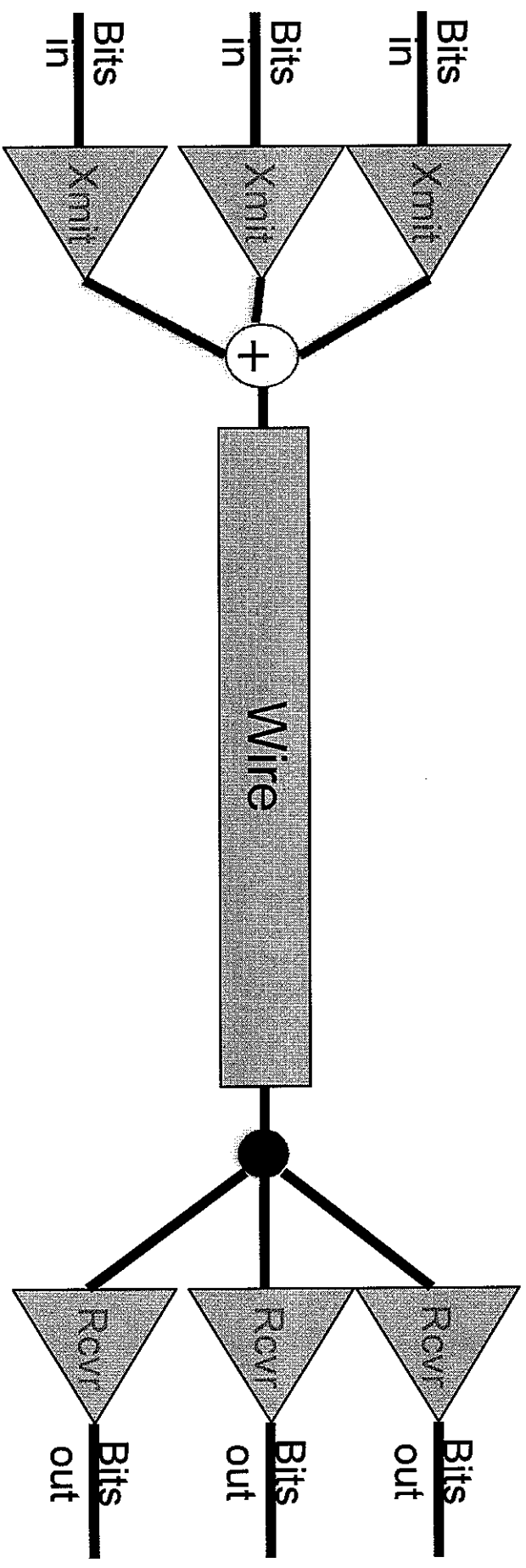
So Far - Low BER Transmission on Wires

2

- Problems and Analysis Techniques:
 - Intersymbol Interference (ISI)
 - LTI Systems and Unit Sample Responses
 - Eye Diagrams
 - Noise
 - Probability Density and Cumulative Distribution Functions
 - Normal (Gaussian) random variables
- Solution Approaches
 - Techniques based on LTI models
 - Deconvolution and Decision Feedback Equalization
 - Error Detection and Correction Codes
 - Parity bits, Reed-Solomon Codes, Viterbi Algorithm

3

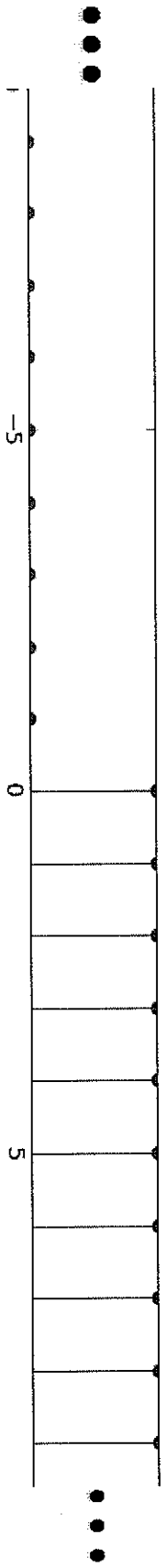
New Problem - Resource Sharing



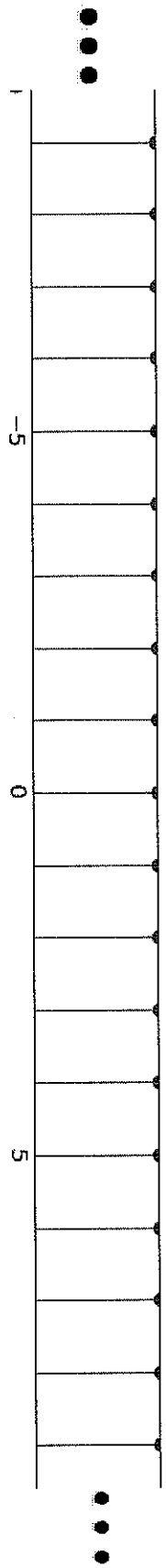
- Two Approaches
 - Time Division Multiplexing (TDM)
 - Each Xmit-Rcvr pair gets a time slot (how to decide?)
 - Used by wired internet
 - Frequency Division Multiplexing (FDM)
 - Each Xmit-Rcvr gets part of the spectrum (explained later)
 - Like Broadcast TV and Radio, many wireless devices
- Next Two Weeks on FDM

Consider Samples for $-\infty < n < \infty$

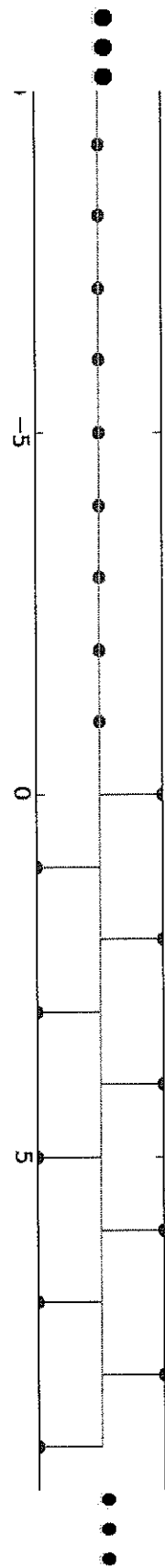
$$x[n] = 1^n u[n]$$



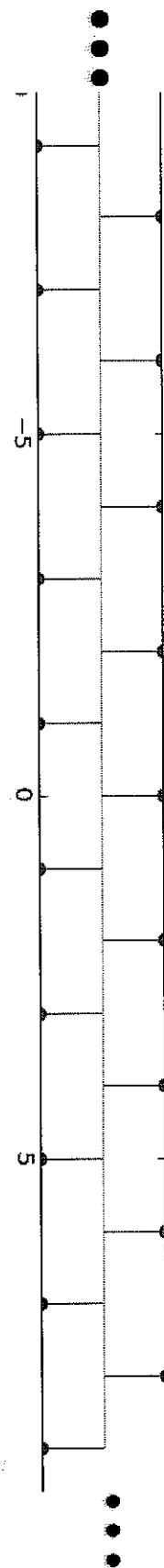
$$x[n] = 1^n$$



$$x[n] = (-1)^n u[n]$$



$$x[n] = (-1)^n$$



LTI Systems for Eternal Signals

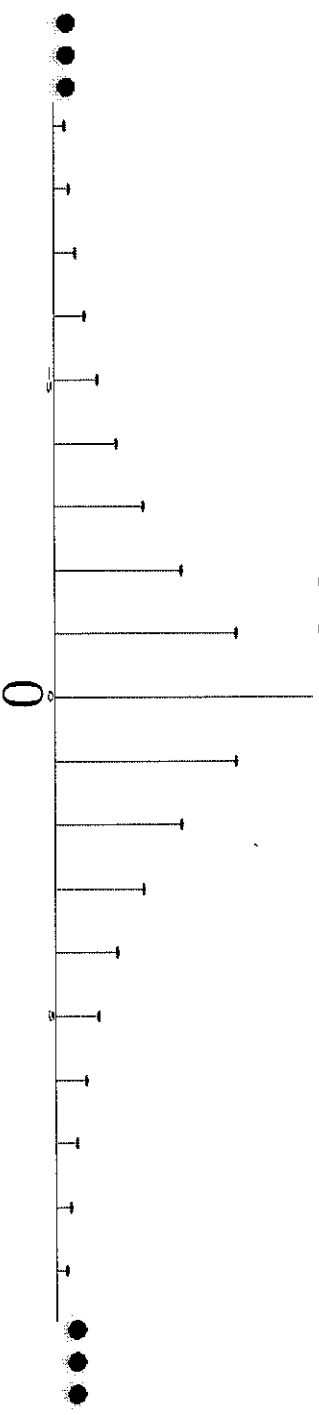
- Convolution Changes a little

$$y[n] = \sum_{m=-\infty}^{\infty} h[m]x[n - m]$$

$x[n]$ may be eternal

$h[m]$ may not be causal

$$h[n] = 0.7^{|n|}$$



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Suppose $x[n] \equiv z^n$, $-\infty < n < \infty$

- Find y using convolution

$$y[n] \equiv \sum_{m=-\infty}^{\infty} h[m] z^{n-m}$$

Reorganizing

$$y[n] \equiv \left(\sum_{m=-\infty}^{\infty} h[m] z^{-m} \right) z^n$$

Just a number if sum converges = $H(z)$

$$y[n] \equiv H(z) z^n$$

$y[n]$ has the same form as $x[n]$!

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So What

- Suppose $x[n] = A_1 z_1^n + A_2 z_2^n$

$$y[n] = \left(\sum_{m=-\infty}^{\infty} h[m] z_1^{-m} \right) A_1 z_1^n + \left(\sum_{m=-\infty}^{\infty} h[m] z_2^{-m} \right) A_2 z_2^n$$



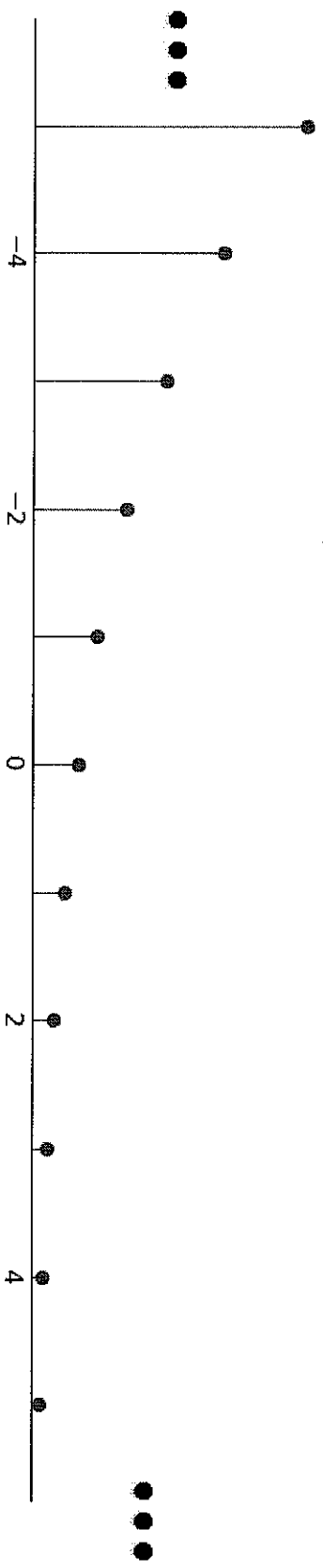
$$y[n] = H(z_1) A_1 z_1^n + H(z_2) A_2 z_2^n$$

- If we can separate z_1^n part from z_2^n part in $y[n]$
 - Can recover two messages
- Could then use lots of z_i 's
 - Multiple channels!

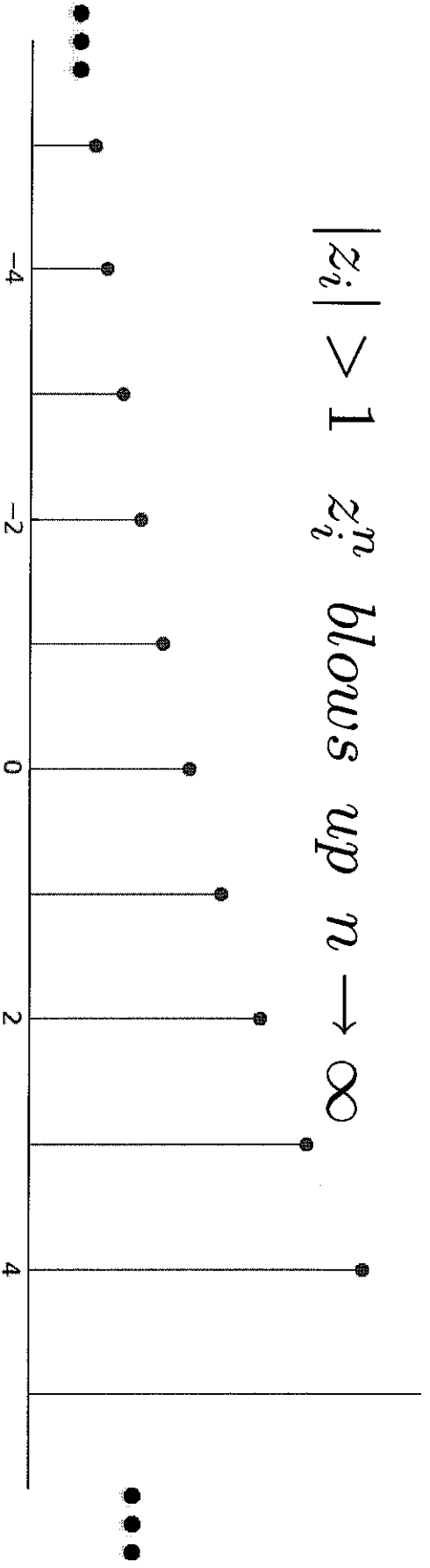
28

A problem with eternal signals

$|z_i| < 1$ z_i^n blows up $n \rightarrow -\infty$



$|z_i| > 1$ z_i^n blows up $n \rightarrow \infty$



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Use $|z_i| = 1$

- How many z 's are there with unit magnitude?
 - $z=1$, $z=-1$, any others?

- Can generate more z 's using complex numbers

$$z = e^{j\Omega} = \cos\Omega + j\sin\Omega$$

- Magnitude is one for any frequency

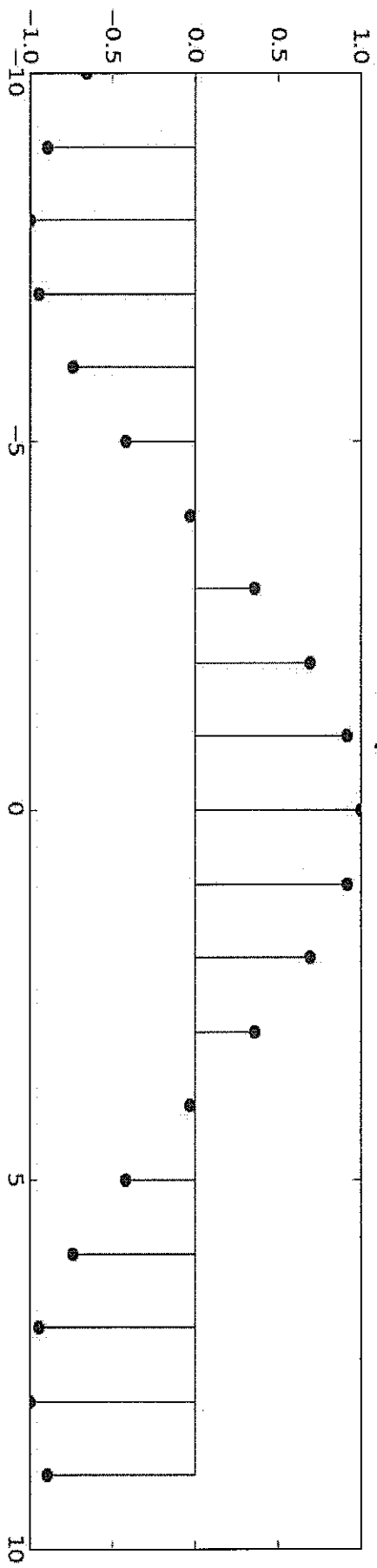
$$|z| = |e^{j\Omega}| = |\cos\Omega + j\sin\Omega| = \sqrt{\cos^2\Omega + \sin^2\Omega} = 1$$

- Know how to evaluate z^n

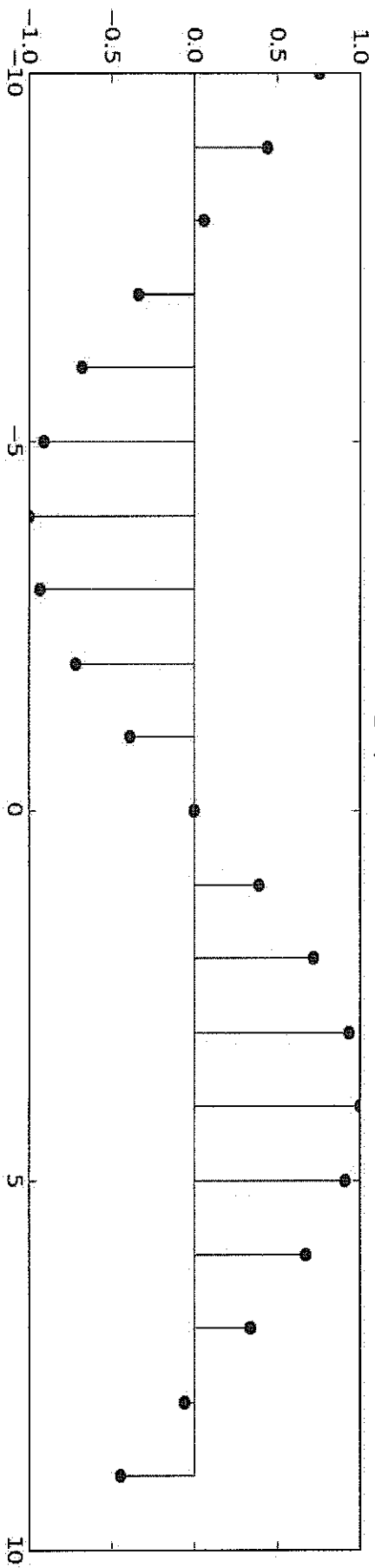
$$z^n = e^{j\Omega n} = \cos\Omega n + j\sin\Omega n$$

Example: $z^n = e^{0.4jn} = \cos 0.4n + j \sin 0.4n$

Real part of $e^{0.4jn}$



Imag part of $e^{0.4jn}$



$|e^{j\omega n}| = 1!$



Frequency Response

- From convolution

$$y[n] = \sum_{m=-\infty}^{\infty} h[m] e^{j\Omega(n-m)}$$



Reorganizing

$$y[n] = \left(\sum_{m=-\infty}^{\infty} h[m] e^{-j\Omega m} \right) e^{j\Omega n}$$

A complex number if the sum converges



$$y[n] = H(e^{j\Omega}) e^{j\Omega n}$$

$H(e^{j\Omega})$, $-\pi < \Omega < \pi$, is the frequency response

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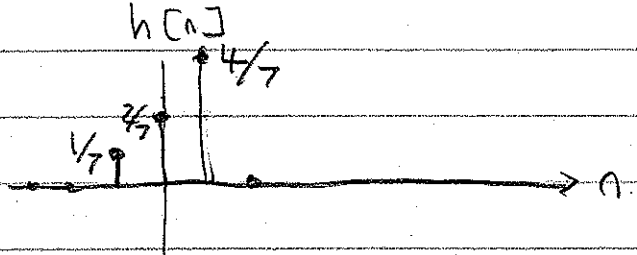
Summary and Next Time

- Frequency Division Multiplexing
 - Eternal z^n 's do not mix when input to LTI systems
 - Need many eternal z^n 's that do not blow up, $|z| = 1$
 - Use $z_i = e^{j\Omega_i}$, $-\pi < \Omega_i \leq \pi$
- Next Week
 - How do we separate the different frequencies in $y[n]$
 - Filters and Fourier Analysis
- After Spring Break
 - Encoding Information using different frequencies
 - What happens when we use
- Do the modulated complex exponentials still stay separated?

$$x[n] = A_1[n]e^{j\Omega_1 n} + A_2[n]e^{j\Omega_2 n}$$

Example

$$X[n] = \left(\frac{1}{2}\right)^n$$

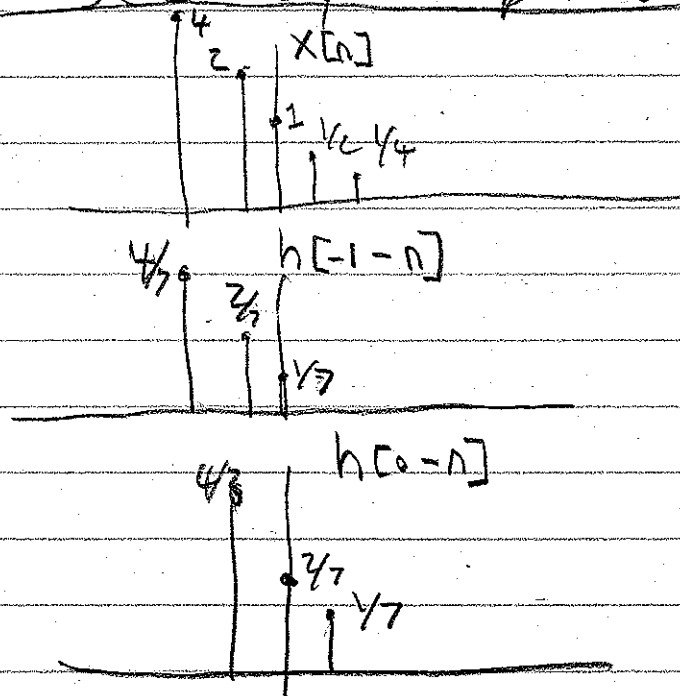
$$Y[n] = \sum h[m] \frac{1}{2}^{(n-m)}$$


$$Y[n] = \left(\frac{1}{7}\right) \frac{1}{2}^{(n+1)} + \left(\frac{2}{7}\right) \frac{1}{2}^{(n-0)} + \left(\frac{4}{7}\right) \left(\frac{1}{2}\right)^{n-1}$$

$$= \left(\frac{1}{2} \cdot \frac{1}{7} + \frac{2}{7} \cdot 1 + \frac{4}{7} \left(\frac{1}{2}\right)^{-1}\right) \left(\frac{1}{2}\right)^n$$

$$= \left(\frac{1}{14} + \frac{2}{7} + \frac{8}{7}\right) \left(\frac{1}{2}\right)^n = \frac{3}{2} \cdot \left(\frac{1}{2}\right)^n$$

Also see by flip & slide

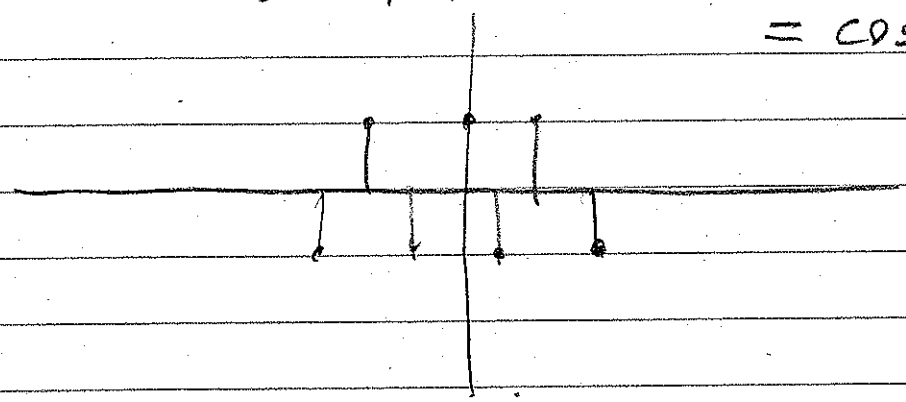


$$Y[-1] = \frac{1}{7} \cdot 1 + \frac{2}{7} \cdot 2 + \frac{4}{7} \cdot 4 = 3 = \frac{3}{2} \cdot 2$$

$$Y[0] = \frac{1}{2} \cdot \frac{1}{7} + 1 \cdot \frac{2}{7} + 2 \cdot \frac{4}{7} = \frac{3}{2}$$

Only consider $-\pi \leq \Omega \leq \pi$

Highest frequency = $e^{j\pi n} = \cos \pi n + j \sin \pi n$
 $= \cos \pi n = (-1)^n$



$$e^{j(-\pi)n} = \cos(\pi)n + j \sin(-\pi)n$$

$$= \cos \pi n = (-1)^n$$

Consider $\Omega = \pi + \Delta$ $\Delta > 0$ (outside $-\pi \rightarrow \pi$)

$$e^{j(\pi + \Delta)n} = \cos(\pi + \Delta)n + j \sin(\pi + \Delta)n$$

$$e^{j\pi n} e^{j\Delta n} = (-1)^n (\cos \Delta n + j \sin \Delta n)$$

$$e^{j(-\pi + \Delta)n} = e^{j(-\pi)n} e^{j\Delta n}$$

$$= (-1)^n (\cos \Delta n + j \sin \Delta n)$$

$$e^{j(\pi + \Delta)n} = e^{j(-\pi + \Delta)n}$$

$\Omega = \pi + \Delta$ same as $\Omega = \Delta - \pi$
 outside $-\pi \rightarrow \pi$ Inside $-\pi \rightarrow \pi$

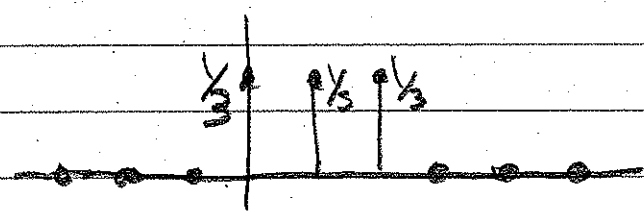
Example

$\infty \leftarrow$ eternal x

$$y[n] = \sum_{m=0}^{\infty} h[m] x[n-m]$$

\nwarrow causal h

$h[n]$



Case 1 $x[n] = \left(\frac{1}{2}\right)^n$

$$y[n] = \left(\sum_{m=0}^{\infty} h[m] \left(\frac{1}{2}\right)^{n-m} \right) \left(\frac{1}{2}\right)^n$$

$$= \left(\frac{1}{3} \cdot \left(\frac{1}{2}\right)^0 + \frac{1}{3} \left(\frac{1}{2}\right)^{-1} + \frac{1}{3} \left(\frac{1}{2}\right)^{-2} \right) \left(\frac{1}{2}\right)^n$$

$$= \frac{7}{3} \left(\frac{1}{2}\right)^n \text{ for all } n$$

Case 2 $x[n] = \left(\frac{1}{2}\right)^n u[n]$

$$y[0] = \frac{1}{3} \cdot \left(\frac{1}{2}\right)^0 = \frac{1}{3} \neq \frac{7}{3} \left(\frac{1}{2}\right)^0$$

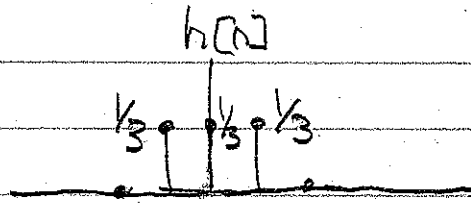
$$y[1] = \frac{1}{3} \cdot \left(\frac{1}{2}\right)^0 + \frac{1}{3} \left(\frac{1}{2}\right)^1 = \frac{1}{2} \neq \frac{7}{3} \left(\frac{1}{2}\right)^1$$

$$y[2] = \frac{1}{3} \left(\frac{1}{2}\right)^0 + \frac{1}{3} \left(\frac{1}{2}\right)^1 + \frac{1}{3} \left(\frac{1}{2}\right)^2 = \frac{7}{12} = \frac{7}{3} \left(\frac{1}{2}\right)^2$$

\vdots
 $y[n] = \frac{7}{3} \left(\frac{1}{2}\right)^n$ only for large n

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Example



$$H(e^{j\Omega}) = \frac{1}{3} e^{j\Omega} + \frac{1}{3} e^{-j\Omega} + \frac{1}{3}$$

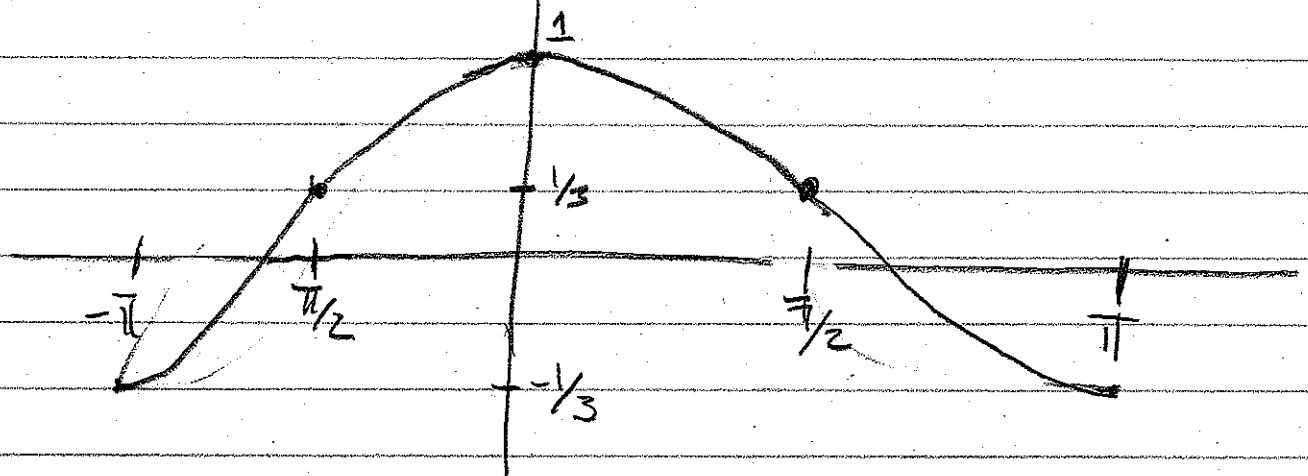
$$= \frac{1}{3} (\cos\Omega + j\sin\Omega)$$

$$+ \frac{1}{3} (\cos\Omega - j\sin\Omega)$$

$$+ \frac{1}{3}$$

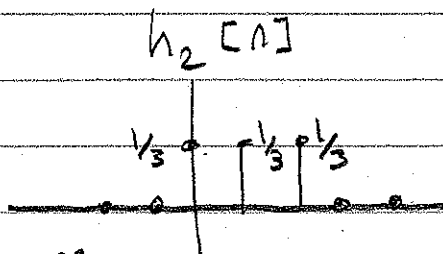
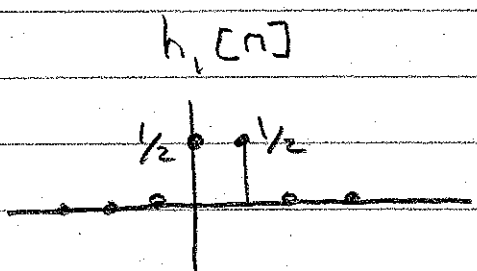
$$= \frac{2}{3} \cos\Omega + \frac{1}{3}$$

$\text{Re } H(e^{j\Omega})$



Example

Frequency Response in two cases

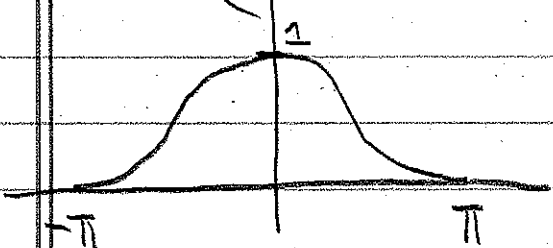


$$H(e^{j\Omega}) = \sum_{m=0}^{\infty} h[m] e^{-j\Omega m}$$

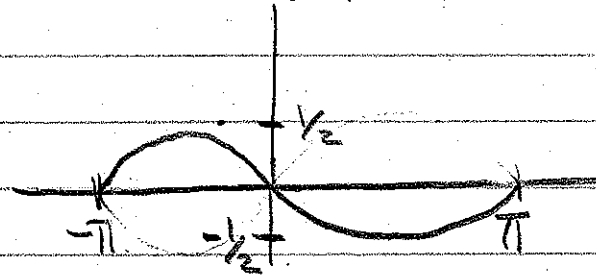
$$\begin{aligned} H_1(e^{j\Omega}) &= h_1[0] \cdot e^{-j\Omega \cdot 0} + h_1[1] e^{-j\Omega \cdot 1} \\ &= \frac{1}{2}(1 + \cos \Omega) - \frac{j}{2} \sin \Omega \end{aligned}$$

$$\begin{aligned} H_2(e^{j\Omega}) &= \frac{1}{3} e^{j\Omega \cdot 0} + \frac{1}{3} e^{j\Omega \cdot 1} + \frac{1}{3} e^{-j\Omega \cdot 2} \\ &= \frac{1}{3} (1 + \cos \Omega + \cos 2\Omega) \\ &\quad - \frac{j}{3} (\sin \Omega + \sin 2\Omega) \end{aligned}$$

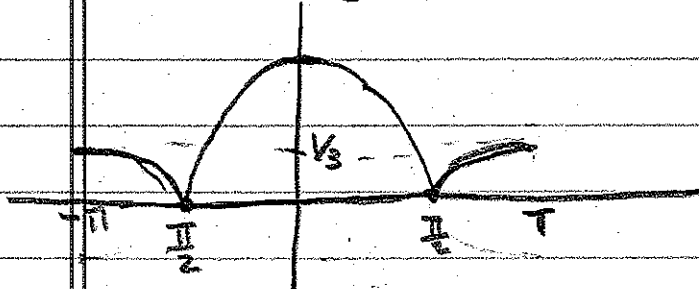
$\text{Re}(H_1(e^{j\Omega}))$



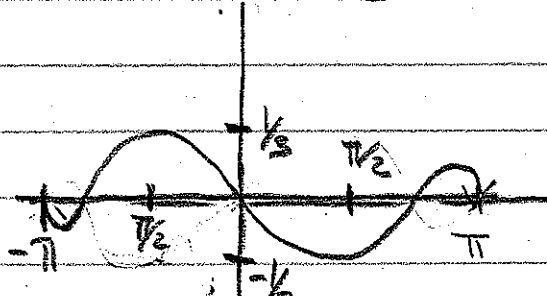
$\text{Im}(H_1(e^{j\Omega}))$



$\text{Re} H_2(e^{j\Omega})$



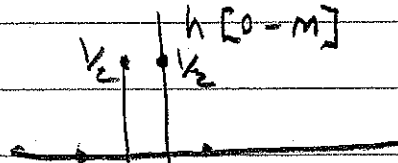
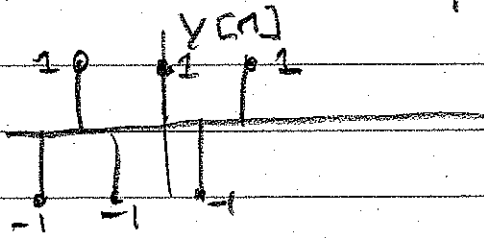
$\text{Im} H_2(e^{j\Omega})$



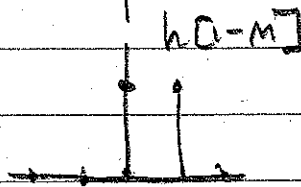
Note if $x[n] = (-1)^n = e^{j\pi n}$

$$y_1[n] = H_1(e^{j\pi}) e^{j\pi n} = 0 \cdot e^{j\pi n} = 0$$

Makes sense convolve h_1 with y



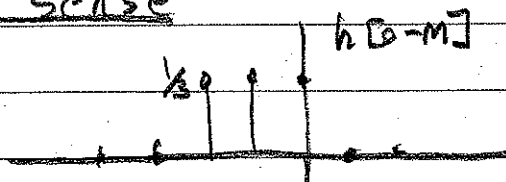
$$\Rightarrow y[0] = 1 \cdot \frac{1}{2} + (-1) \cdot \frac{1}{2} = 0$$



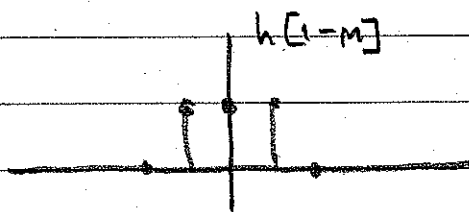
$$y[1] = (-1) \cdot \frac{1}{2} + (1) \cdot \frac{1}{2} = 0$$

$$y_2[n] = H_2(e^{j\pi}) e^{j\pi n} = \frac{1}{3} e^{j\pi n} = \frac{1}{3} (-1)^n$$

Makes sense



$$y[0] = \frac{1}{3}$$



$$y[1] = -\frac{1}{3}$$

$$y[2] = \frac{1}{3}$$

Suppose $X[n] = \cos \Omega n = \frac{1}{2}(e^{j\Omega n} + e^{-j\Omega n})$

$$y[n] = \sum h[m] \cos \Omega(n-m)$$

$$y[n] = H(e^{j\Omega}) \frac{1}{2} e^{j\Omega n} + H(e^{-j\Omega}) \frac{1}{2} e^{-j\Omega n}$$

Consider $\Omega = \frac{\pi}{2}$ and system 2

$$H(e^{j\frac{\pi}{2}}) = 0 + -\frac{1}{3}j$$

$$H(e^{-j\frac{\pi}{2}}) = 0 + \frac{1}{3}j$$

$$\begin{aligned}
 y[n] &= -\frac{1}{3}j \frac{1}{2} e^{j\frac{\pi}{2}n} - \frac{1}{3}j \frac{1}{2} e^{-j\frac{\pi}{2}n} \\
 &= -\frac{1}{6}j \left(\cos \frac{\pi}{2}n + j \sin \frac{\pi}{2}n - (\cos \frac{\pi}{2}n - j \sin \frac{\pi}{2}n) \right) \\
 &= -\frac{1}{6}j (2j \sin \frac{\pi}{2}n)
 \end{aligned}$$

$$\rightarrow = \frac{1}{3} \sin \frac{\pi}{2}n$$

Same Amplitude Different phase

Suppose $h[n] = \frac{1}{3} \delta[n-1] + \frac{1}{3} \delta[n] + \frac{1}{3} \delta[n+1]$

(Analyzed Before)

$$\begin{aligned}
 y[n] &= \underbrace{H(e^{j\frac{\pi}{2}})}_{\frac{1}{3}} \frac{1}{2} e^{j\frac{\pi}{2}n} + \underbrace{H(e^{-j\frac{\pi}{2}})}_{\frac{1}{3}} \frac{1}{2} e^{-j\frac{\pi}{2}n} \\
 &= \frac{1}{6} (e^{j\frac{\pi}{2}n} + e^{-j\frac{\pi}{2}n}) \\
 &= \frac{1}{3} \cos \frac{\pi}{2}n
 \end{aligned}$$