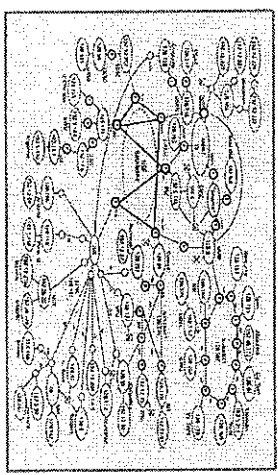
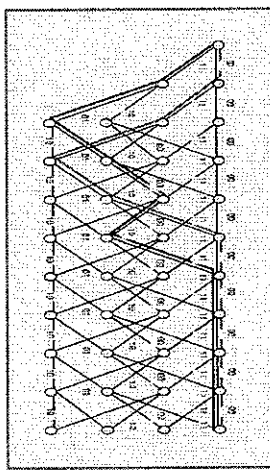
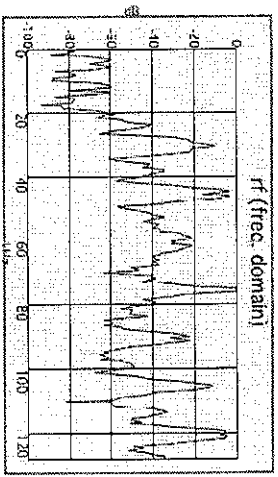
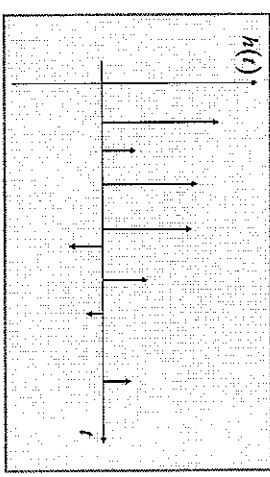


①

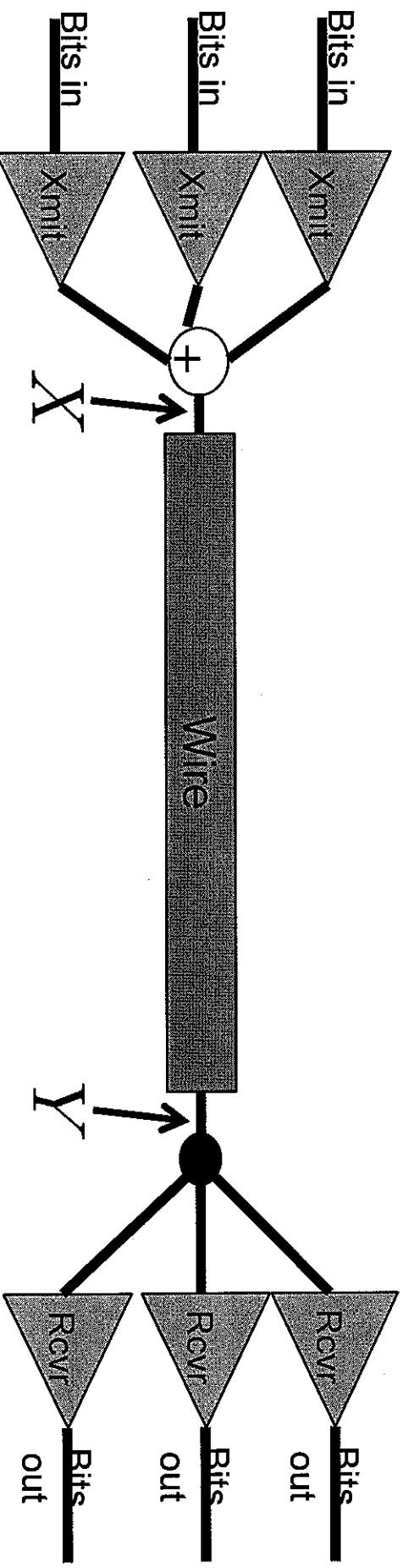


INTRODUCTION TO RECS II  
**DIGITAL  
COMMUNICATION  
SYSTEMS**

**6.02 Spring 2009  
Lecture #14**

- FDM and Data
- Spectrum and Discrete Fourier Transform
- DFT Examples
- Rise Time and Spectrum

# Simplified Big Picture

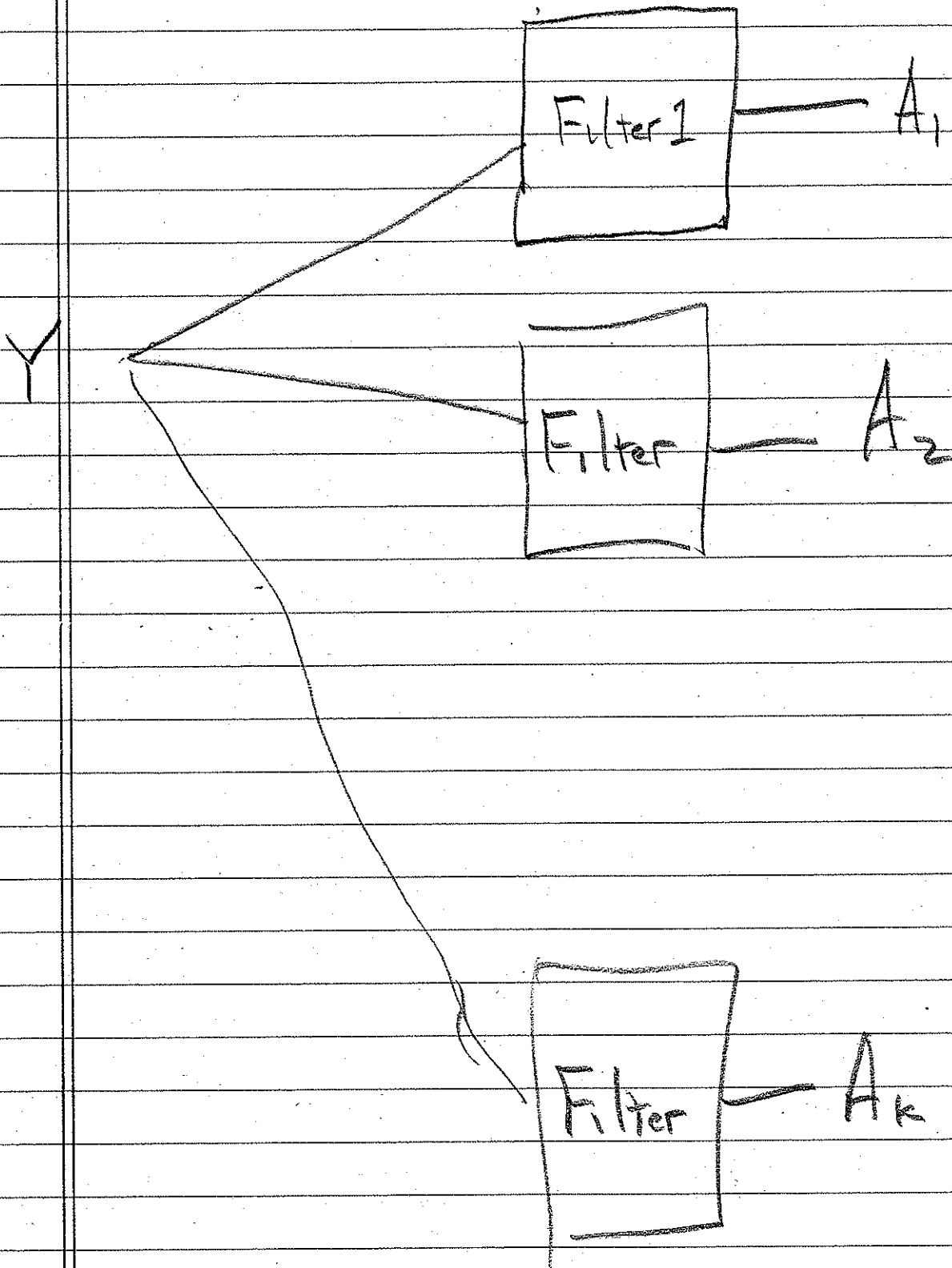


- Frequency Division Multiplexing Strategy
  - Represent each channel with a different frequency
    - For LTI systems, frequencies do not mix
- Use Filters to Separate Into different Channels

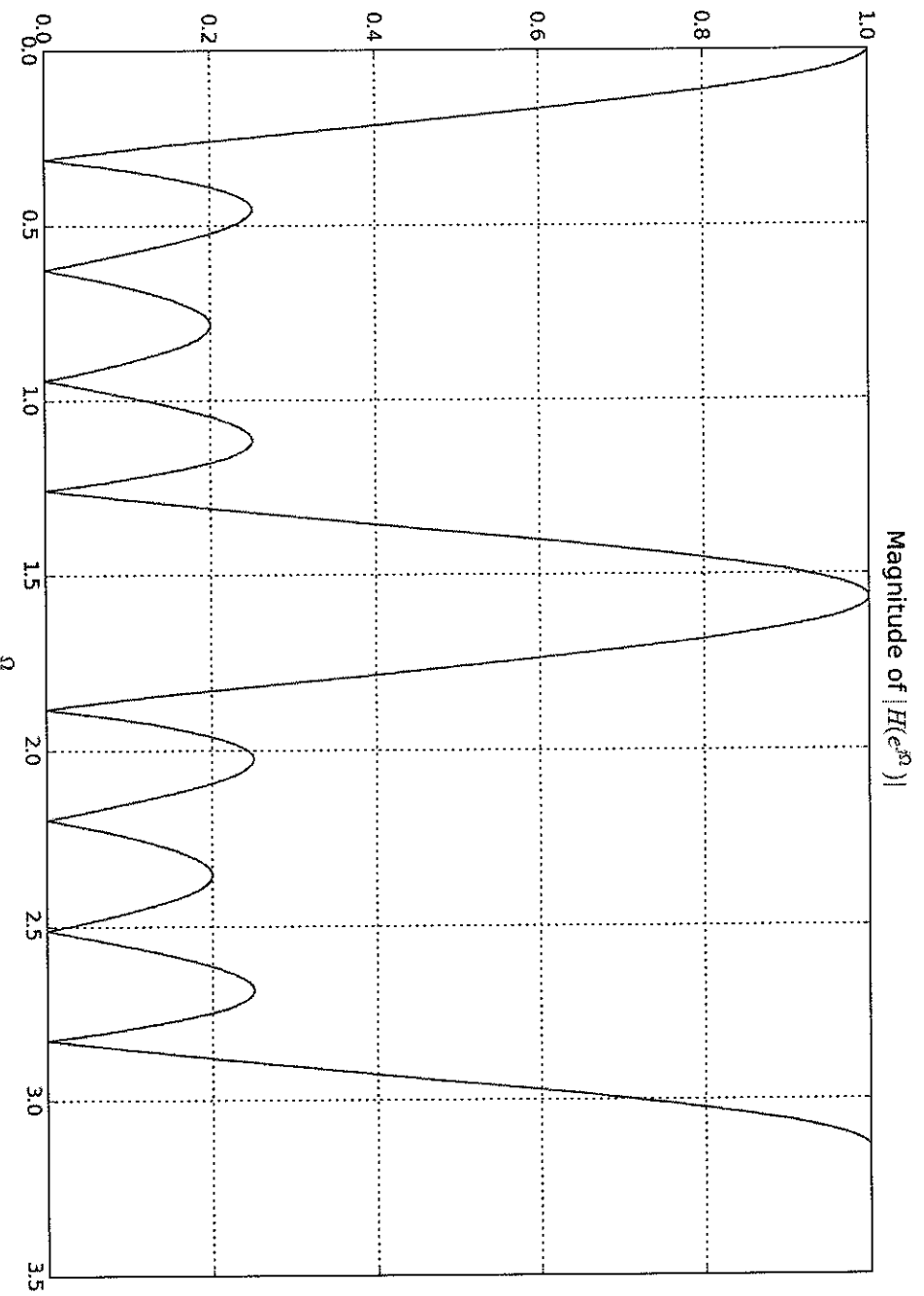
$$x[n] = A_1 e^{j\Omega_1 n} + \dots + A_K e^{j\Omega_K n}$$

$$y[n] = H(e^{j\Omega_1}) A_1 e^{j\Omega_1 n} + \dots + H(e^{j\Omega_K}) A_K e^{j\Omega_K n}$$

3



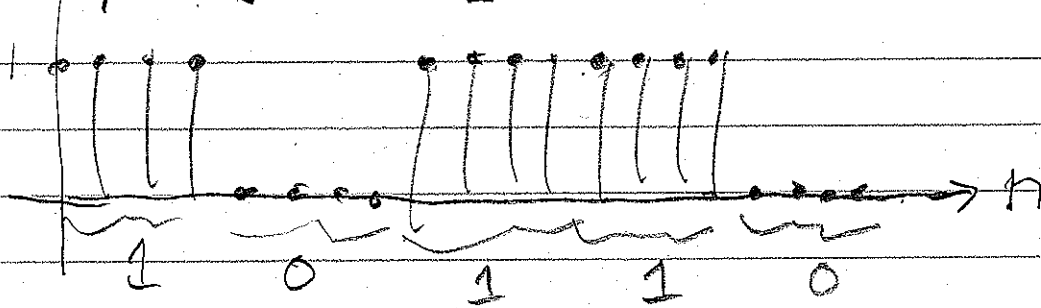
# Channel 5, 8 Zeros Filter



5

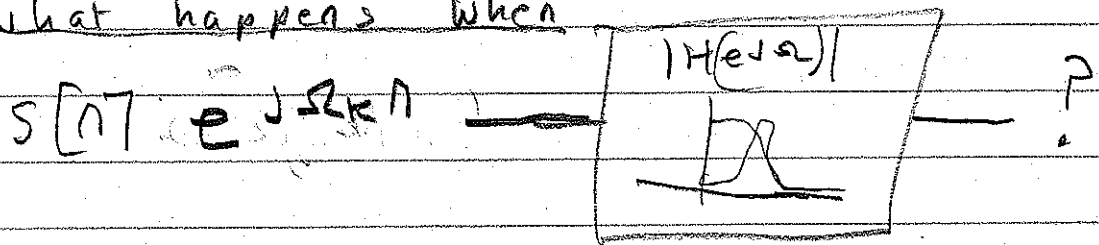
# Real Problem - Encoding Information

Samples  $[n] \equiv S[n]$



$$A_k e^{j\Omega_k n} \Rightarrow S[n] e^{j\Omega_k n}$$

what happens when



How to Analyze?

⑥

Make a periodic Approximation

For some large value of  $N$  (e.g.  $N=100,000$ )

$$\begin{aligned} \text{Assume } S[n+N] e^{j\Omega_k(n+N)} \\ = \\ S[n] e^{j\Omega_k n} \end{aligned}$$

Why: Makes analysis tractable.  
First Major ramification:

$$\text{If } e^{j\Omega_k(n+N)} = e^{j\Omega_k n}$$

Then only certain  $\Omega_k$ 's allowed

$\Omega_k(n+N)$  must equal  $\Omega_k n + 2\pi L$   
for some integer  $L$

(Recall  $-\pi \leq \Omega_k \leq \pi$ )

Therefore

$$\Omega_k = \pm \frac{2\pi}{N} \cdot 0, \pm \frac{2\pi}{N} \cdot 1, \pm \frac{2\pi}{N} \cdot 2, \dots, \pm \frac{2\pi}{N} B$$

$$B = \frac{N-1}{2} \quad \text{Assuming } N \text{ odd}$$

We will always assume  $N$  is odd

7

## Major Ramification 2

If  $S[n]$  is real, periodic with period  $N$ , and  $N$  is odd:

$$S[n] = \sum_{k=1}^K a_k \sin \frac{2\pi}{N} kn + \sum_{k=0}^K b_k \cos \frac{2\pi}{N} kn$$

No  $k=0$  term as  $\sin(0) = 0$        $K = \frac{N-1}{2} \Rightarrow \{a_k's, b_k's\}$  has  $N$  elements

Using Complex Exponentials (And "Negative" Frequencies)

$$S[n] = \frac{1}{N} \sum_{k=-K}^K \underbrace{S[k]}_{\text{Complex Coefficients}} e^{j \frac{2\pi}{N} kn}$$

$$N = 2K + 1$$

Note

$$\underbrace{S[n], n=0, \dots, N-1}_{N \text{ unique Numbers}} \Rightarrow \underbrace{S[-k], \dots, S[k]}_{N \text{ complex Numbers}}$$

But  $S[k]$ 's have certain properties

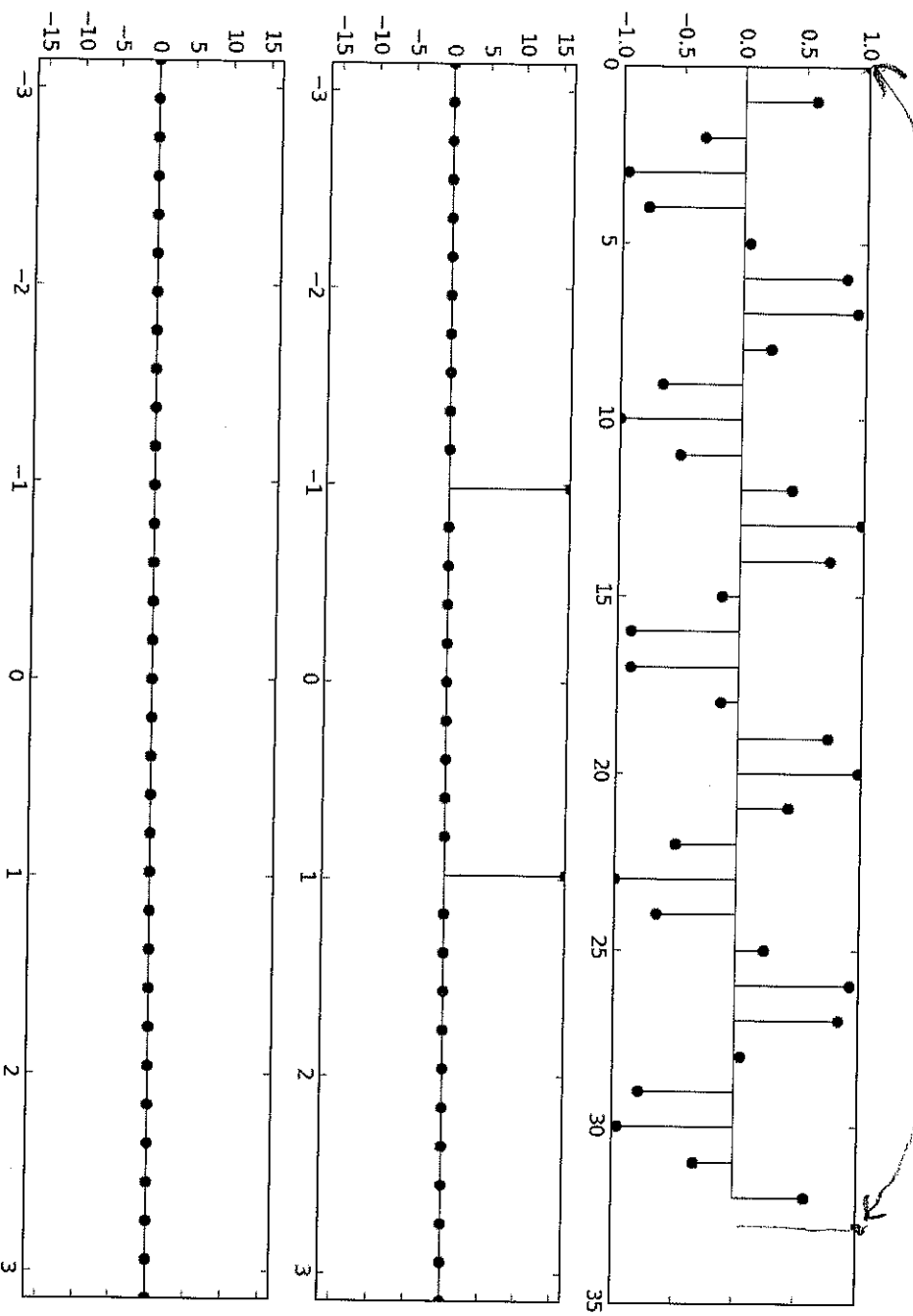
$$S[n] \text{ real} \Rightarrow \underbrace{\text{Re}(S[k]) = \text{Re}(S[-k])}_{\text{Real part even}}$$

$$\underbrace{\text{Im}(S[k]) = -\text{Im}(S[-k])}_{\text{Imag part odd}}$$

8

# Cosine in Time, 2 real nonzero DFT values

*Must be identical*

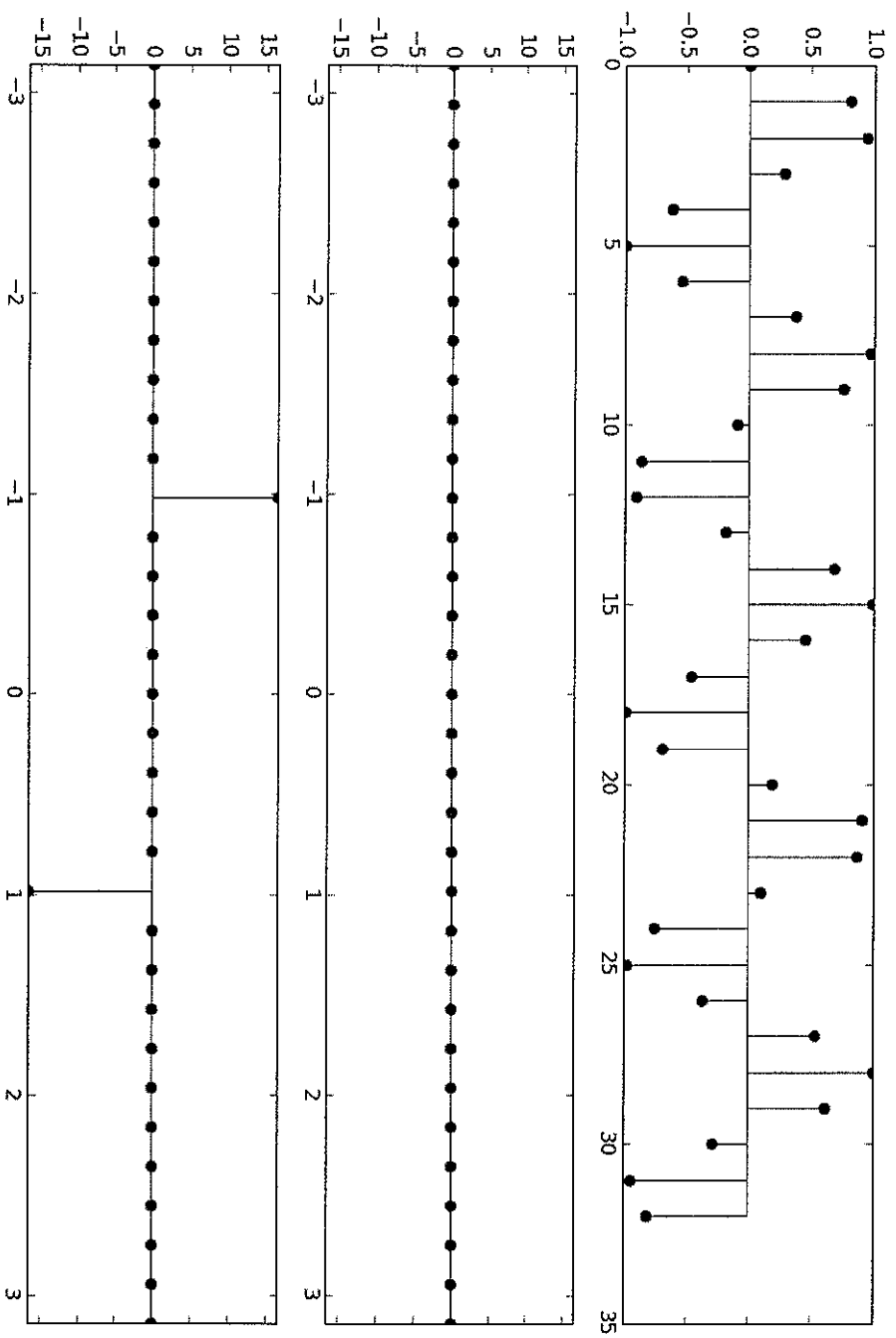


$$\cos\left(\frac{2\pi}{33} \cdot 5n\right)$$



91

# Sine in Time, 2 Imaginary DFT values



$$\sin\left(\frac{2\pi}{33} \cdot 5n\right)$$

# Computing Fourier Coefficients

$$e^{-j\frac{2\pi}{N}l n} s[n] = e^{-j\frac{2\pi}{N}l n} \left( \frac{1}{N} \sum_{k=-R}^R s[k] e^{+j\frac{2\pi}{N}k n} \right)$$

$$\sum_{n=0}^{N-1} \left( e^{-j\frac{2\pi}{N}l n} s[n] \right) = \frac{1}{N} \sum_{n=0}^{N-1} \left( \sum_{k=-R}^R s[k] e^{+j\frac{2\pi}{N}(k-l)n} \right)$$

$$= \frac{1}{N} \sum_{k=-R}^R s[k] \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}(k-l)n}$$

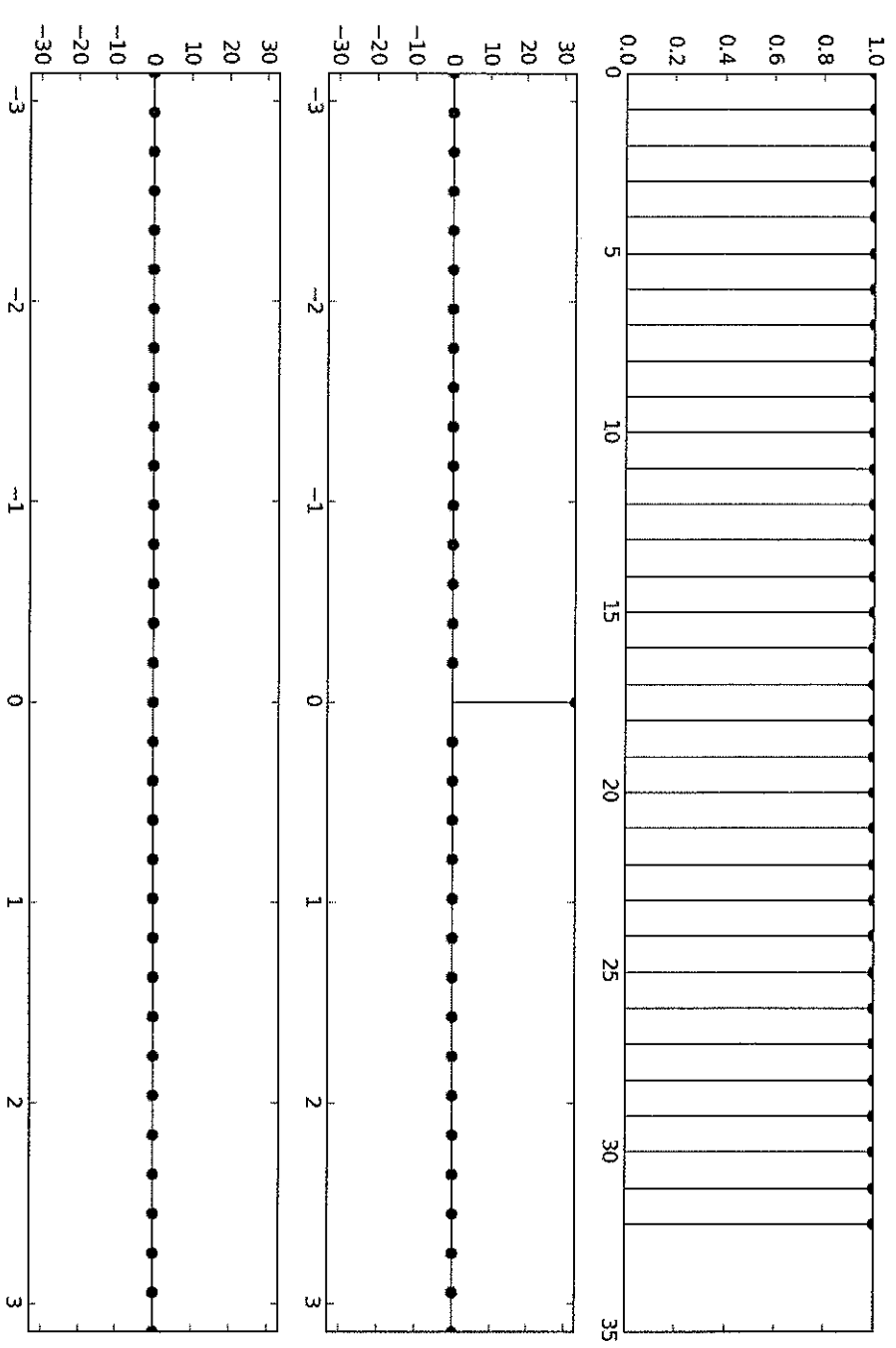
= 0 unless  $k=l$

$$= \frac{1}{N} \cdot s[l] \cdot N$$

$$\Rightarrow \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}k n} s[n] = s[k]$$

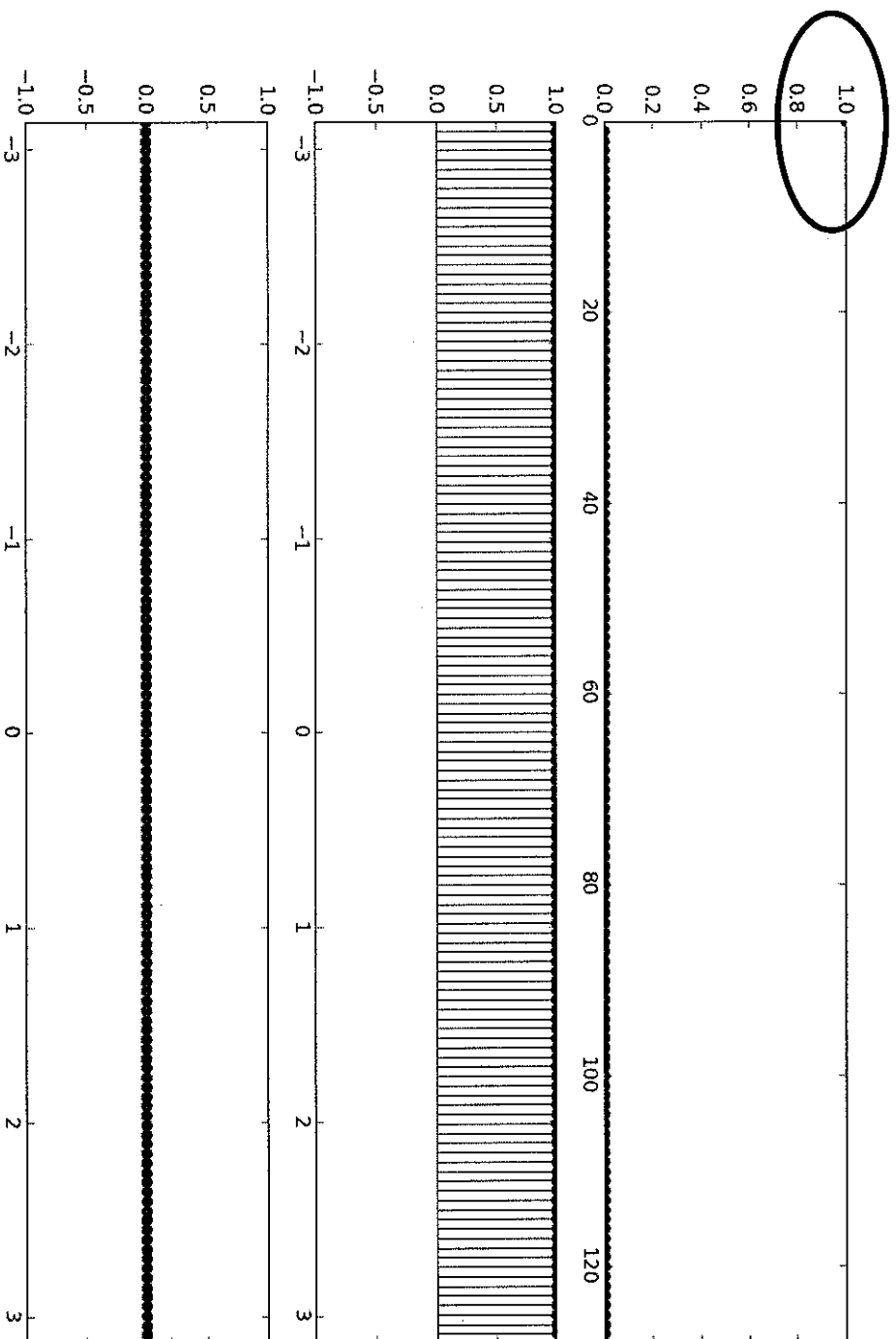


# Constant in Time, one real nonzero DFT value



# One Sample in Time, Constant DFT

12



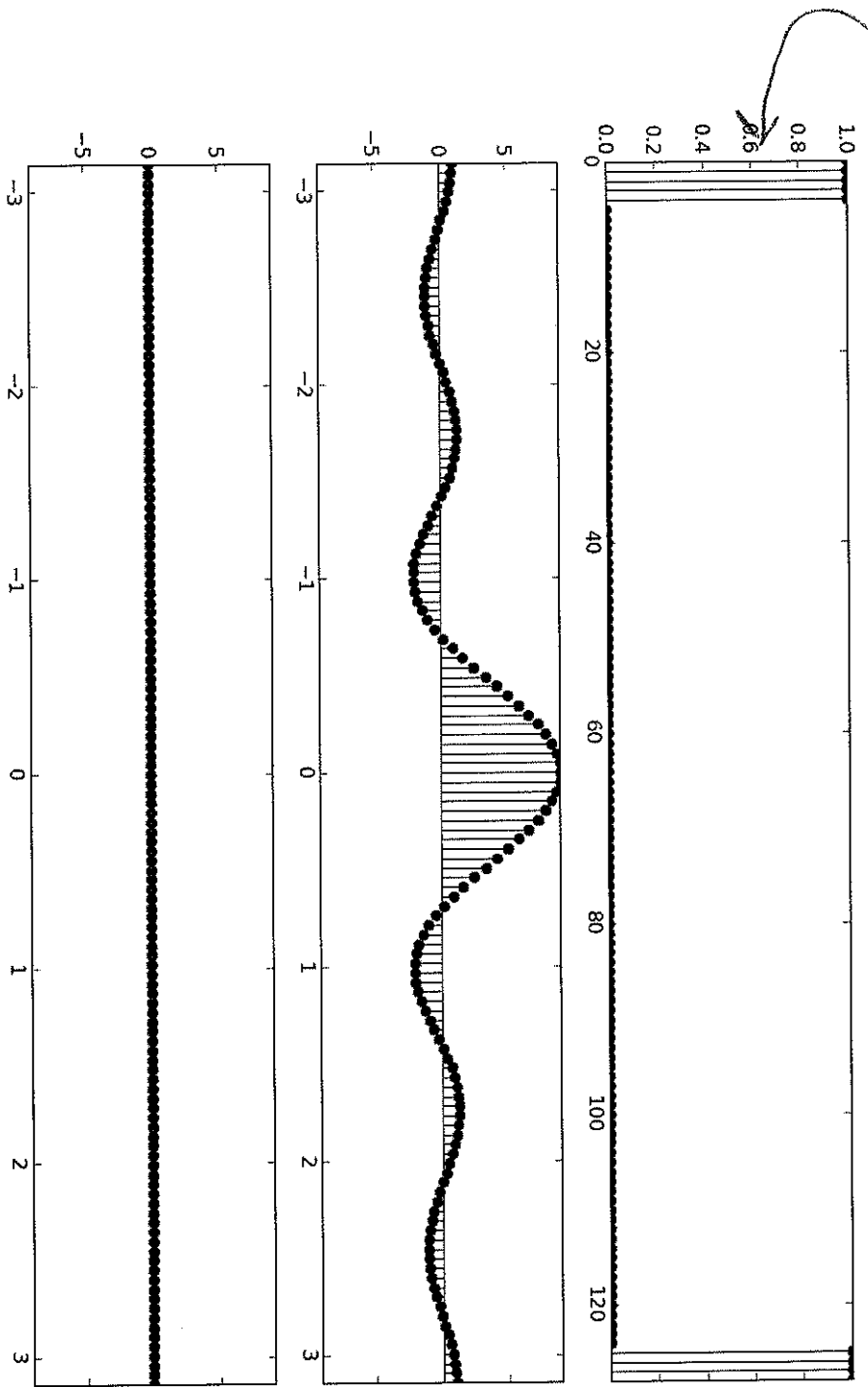
$\text{Re}(S[k])$  vs  $\Omega$

$\text{Im}(S[k])$  vs  $\Omega$

$$\sum_{n=0}^{N-1} x[n] e^{-j2\pi k n} = S[k]$$

# Medium Even Pulse, Real decaying DFT

(13)

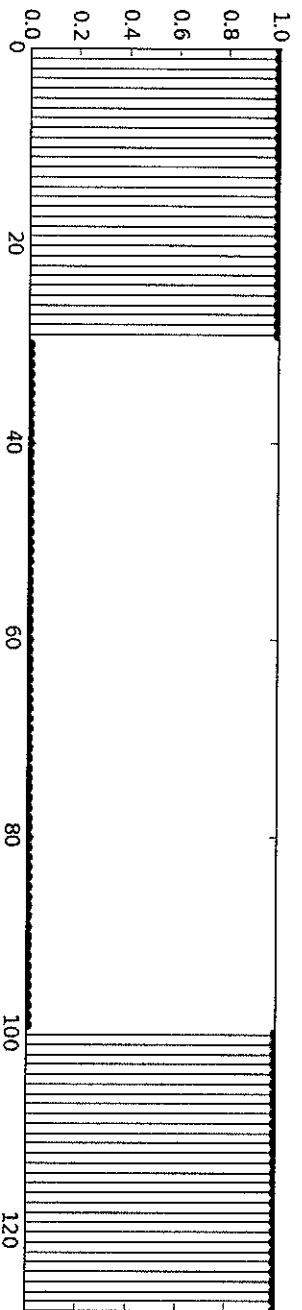


Note  
periodic

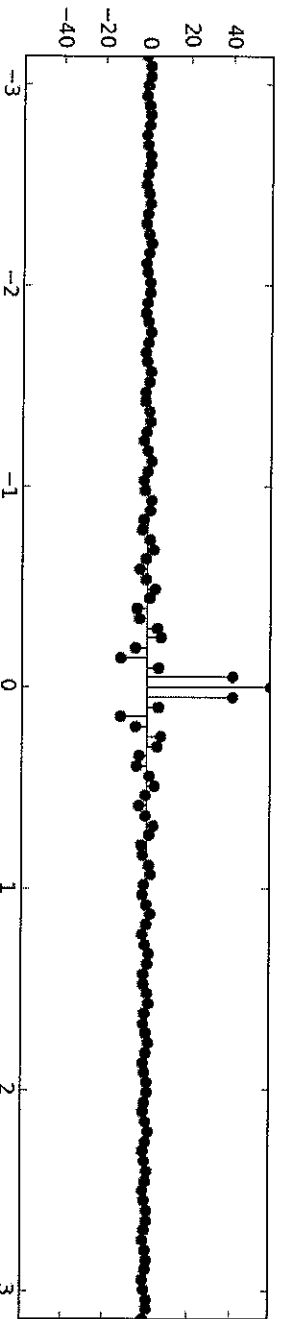
$Re(S[k])$  vs  $\Omega$

$Im(S[k])$  vs  $\Omega$

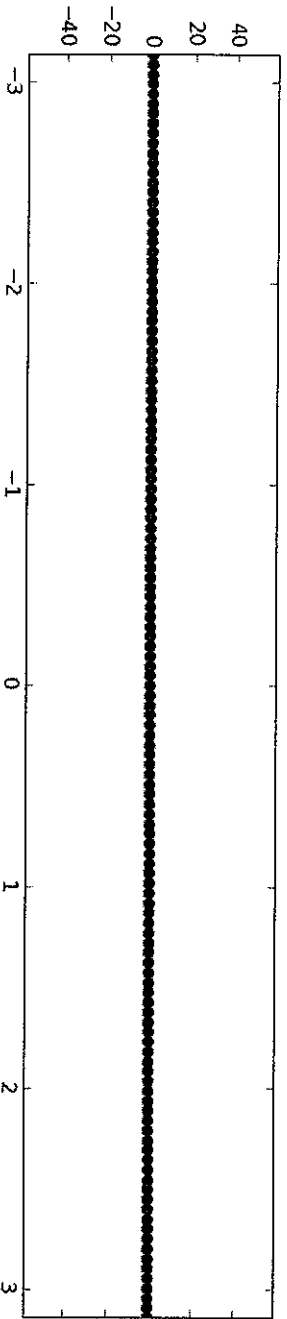
# Wider Even Pulse, Faster Decay



$Re(f(\omega))$  vs  $\omega$

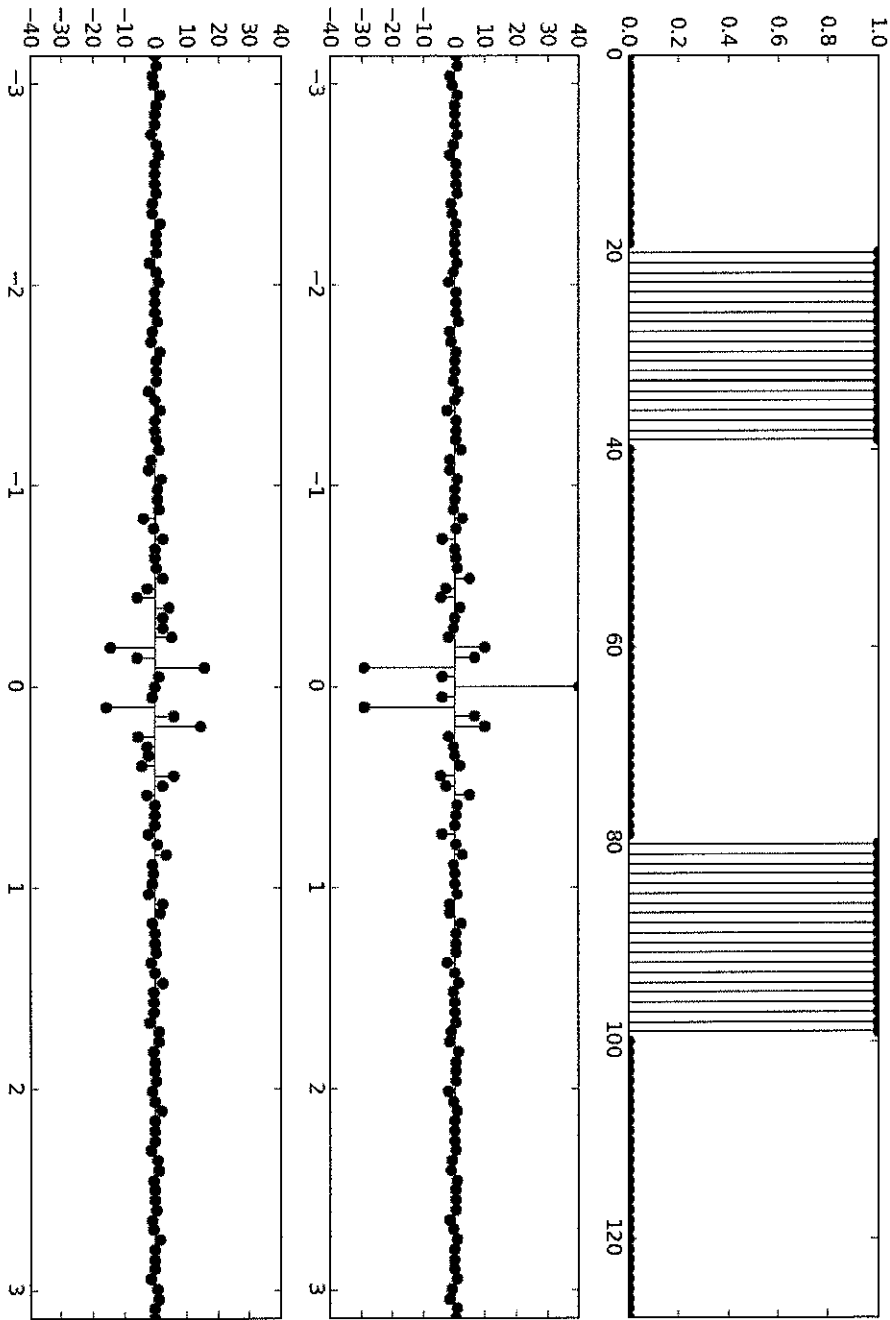


$Im(f(\omega))$  vs  $\omega$



15

# 010010 Bit sequence with Fast Rise

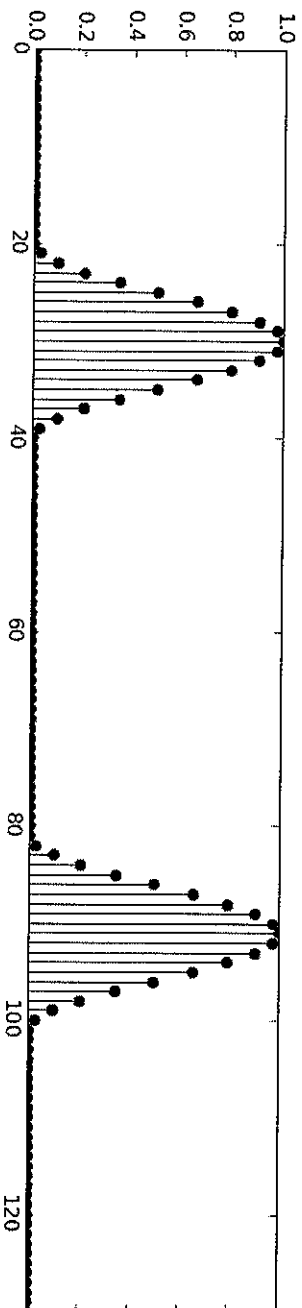


$Re(SkT)$  vs  $Q$

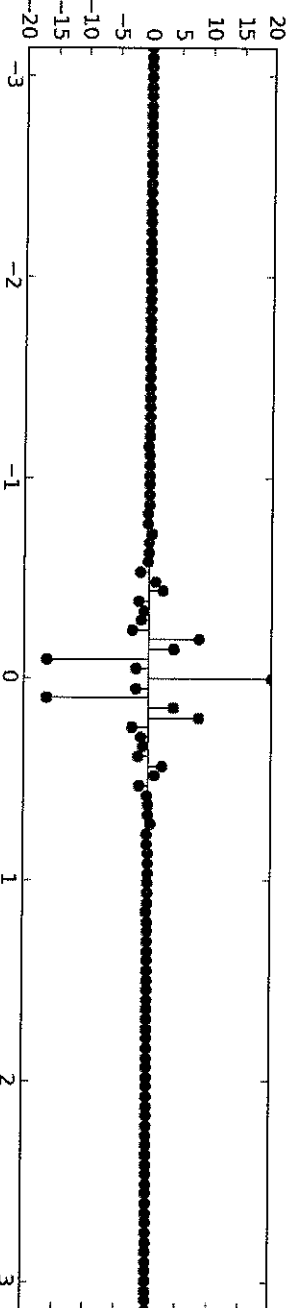
$Im(SkT)$  vs  $Q$

16

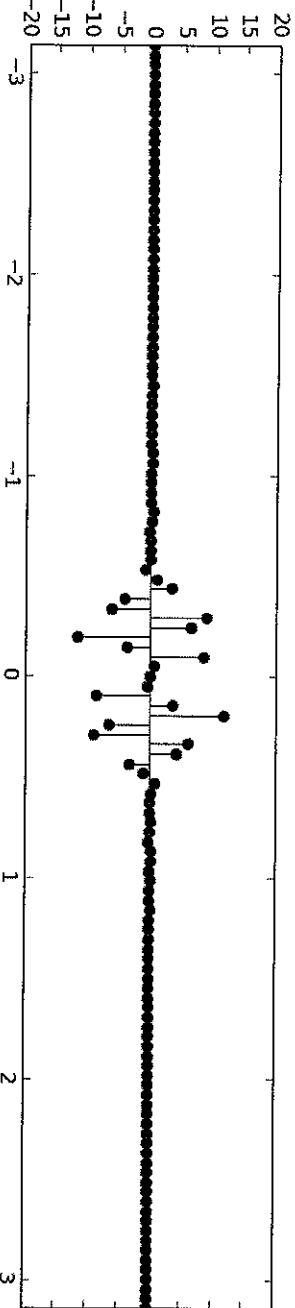
# 010010 Bit sequence with Slow Rise



$Re(SKJ)$  vs  $\Omega$



$Im(SKJ)$  vs  $\Omega$





# Lec. 14 Recitation Notes ①

Recall the Basic Formulas

$$S[n] = \frac{1}{N} \sum_{k=-R}^R \tilde{S}[k] e^{j \frac{2\pi}{N} k n} \quad \begin{array}{l} 2R+1=N \\ (N \text{ odd}) \end{array}$$

$$\tilde{S}[k] = \sum_{n=0}^{N-1} S[n] e^{-j \frac{2\pi}{N} k n}$$

Suppose

$$S[n] = \cos\left(\frac{2\pi}{7} n + \phi\right)$$

$$\Rightarrow S[n] = \frac{1}{2} \left( e^{j\left(\frac{2\pi}{7} n + \phi\right)} + e^{-j\left(\frac{2\pi}{7} n + \phi\right)} \right)$$

$$= \frac{e^{j\phi}}{2} e^{j\frac{2\pi}{7} n} + \frac{e^{-j\phi}}{2} e^{j\frac{2\pi}{7} n}$$

Let  $N = 21$ , what is  $\tilde{S}[k]$ ,  $-10 \leq k \leq 10$

$$\frac{2\pi}{21} k = \frac{2\pi}{7} \quad k = 3$$

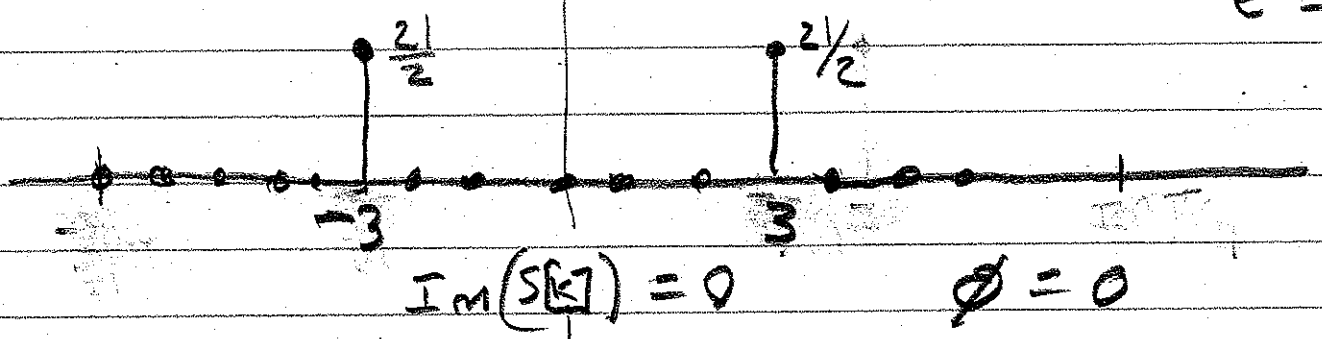
$$\Rightarrow \tilde{S}[k] = 0 \quad k \neq 3, -3$$

$$\tilde{S}[3] = \frac{e^{j\phi}}{2} \cdot N \quad \tilde{S}[-3] = \frac{e^{-j\phi}}{2} \cdot N$$

②

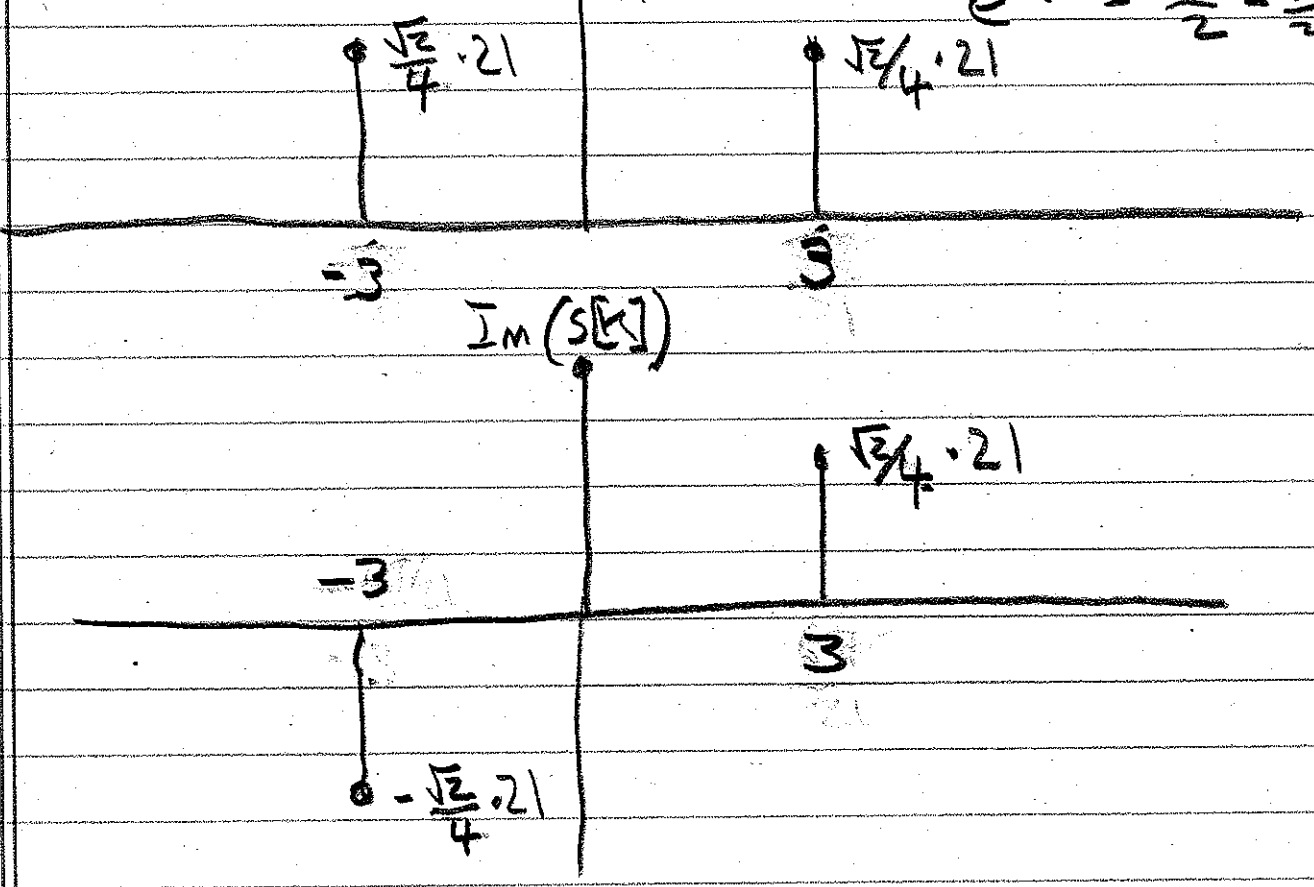
Re( $\Sigma[k]$ )

$\phi = 0$   
 $e^{j\phi} = 1$   
 $e^{-j\phi} = 1$

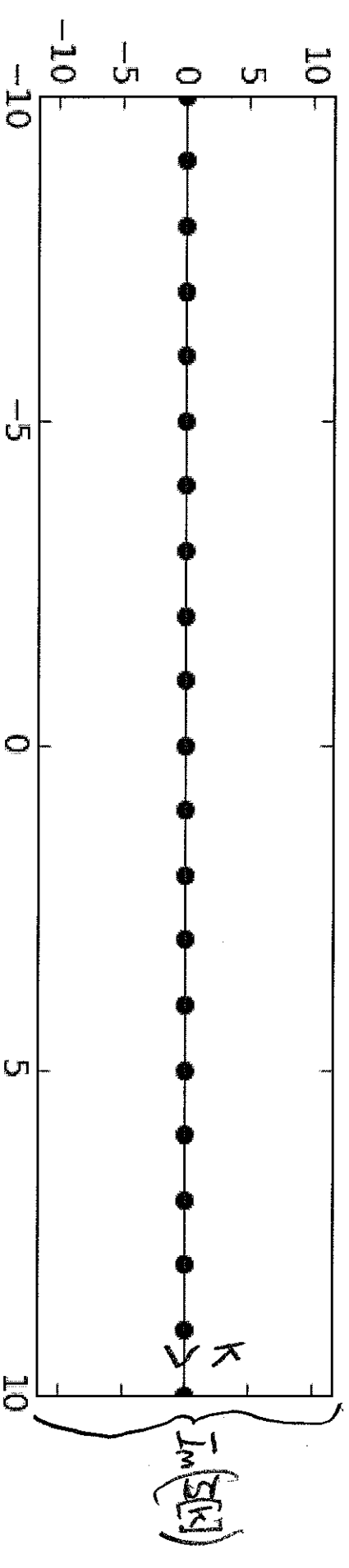
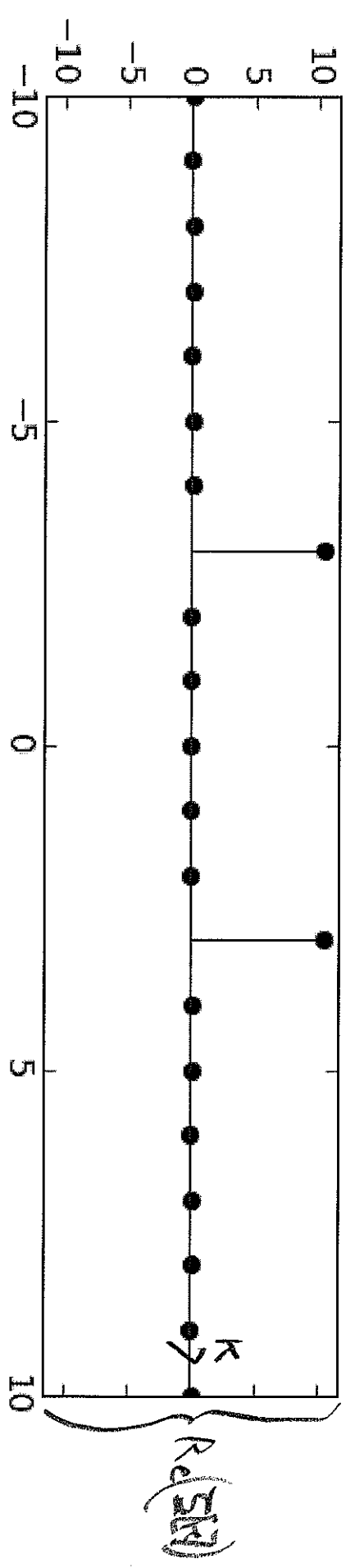
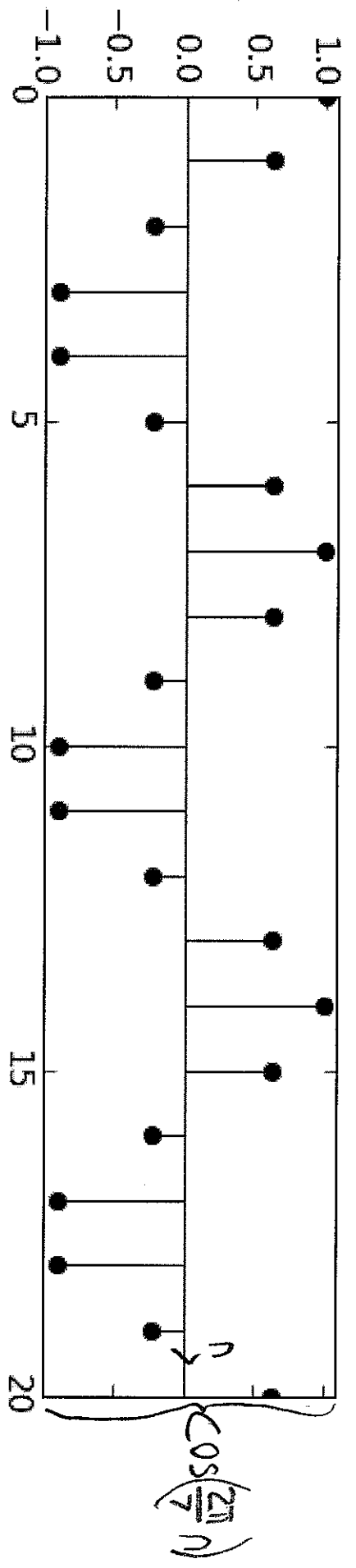


Re( $\Sigma[k]$ )

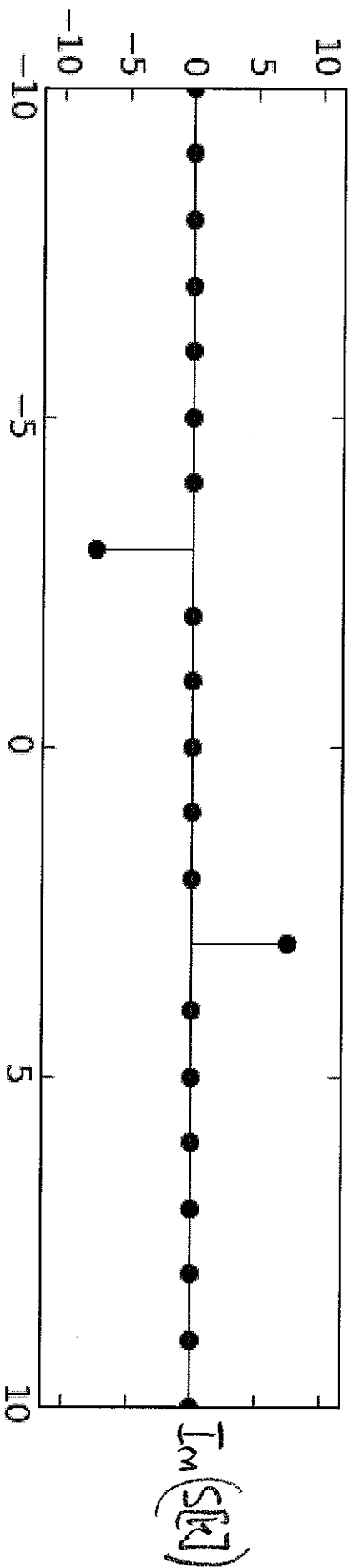
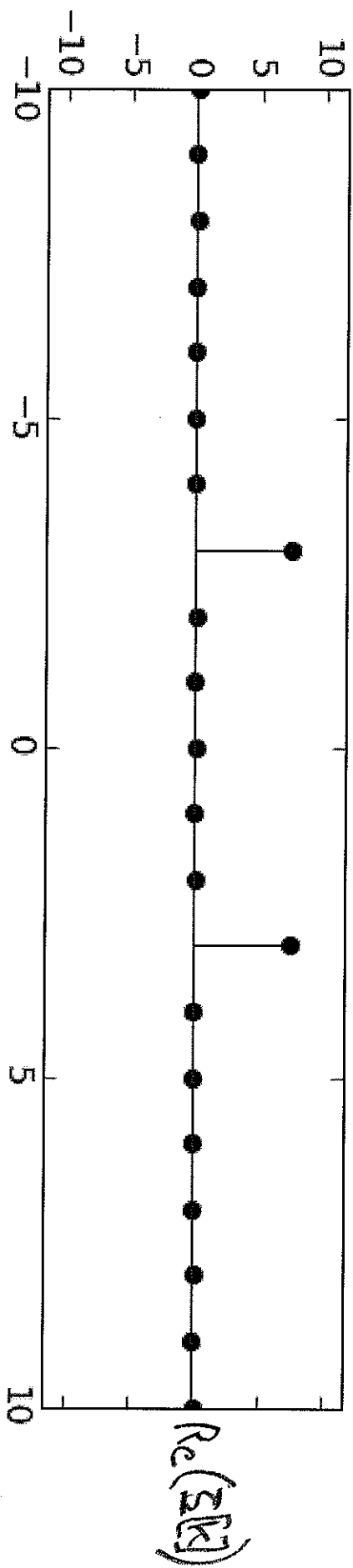
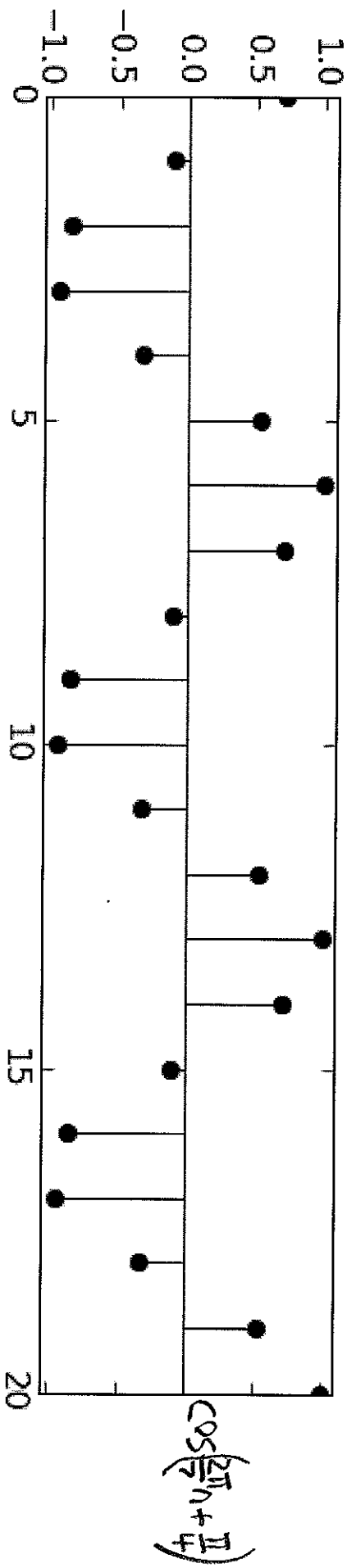
$\phi = \frac{\pi}{4}$   
 $e^{j\phi} = \frac{\sqrt{2}}{2} + \frac{j\sqrt{2}}{2}$   
 $e^{-j\phi} = \frac{\sqrt{2}}{2} - \frac{j\sqrt{2}}{2}$



(25)



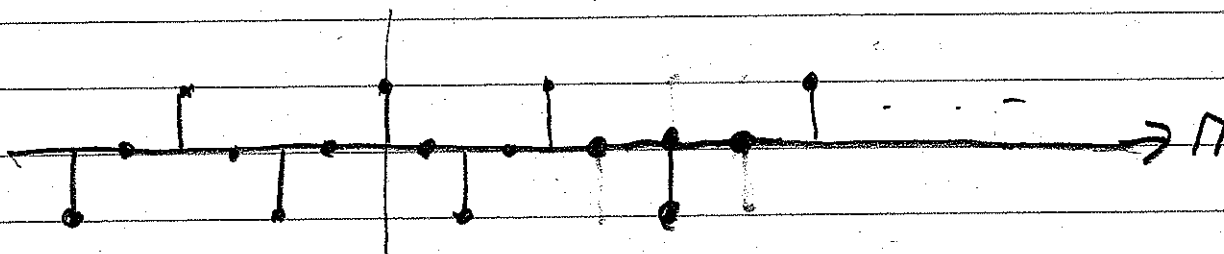
Ⓟ



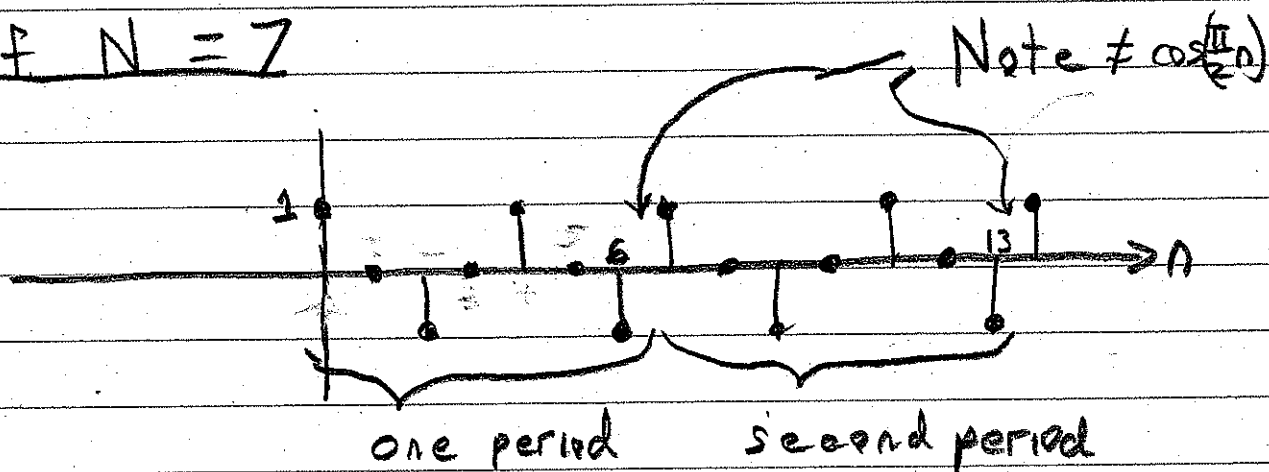
5

# Periodicity Issue

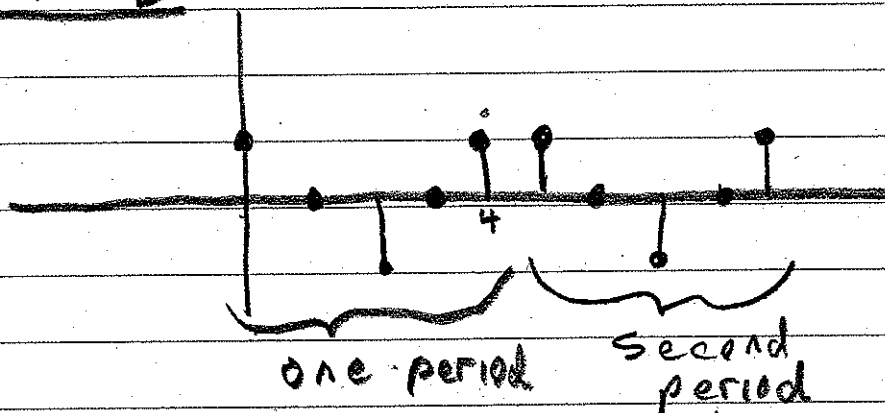
Suppose  $s[n] = \cos\left(\frac{\pi}{2}n\right)$



IF N=7



IF N=5



⑥

IF  $N=5$

$$\underline{S[k]} = \sum_{n=0}^4 s[n] e^{-j \frac{2\pi}{5} kn} =$$

$$1 \cdot e^{-j \frac{2\pi}{5} k \cdot 0} - 1 \cdot e^{-j \frac{2\pi}{5} k \cdot 2} + 1 e^{-j \frac{2\pi}{5} k \cdot 4}$$

$$\Rightarrow \underline{S[0]} = 1 - 1 + 1 = 1$$

$$\underline{S[1]} = 1 - 1 e^{-j \frac{4\pi}{5}} + 1 e^{-j \frac{8\pi}{5}}$$

$$\underline{S[-1]} = 1 - 1 e^{+j \frac{4\pi}{5}} + 1 e^{+j \frac{8\pi}{5}}$$

Note for  $N=5$   $\sum_{n=0}^4 s[n] \neq 0$

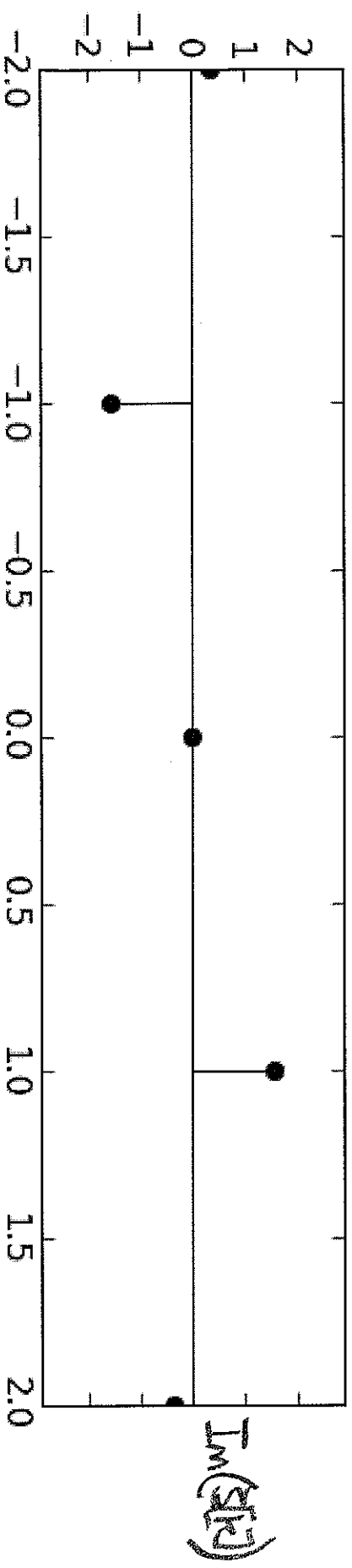
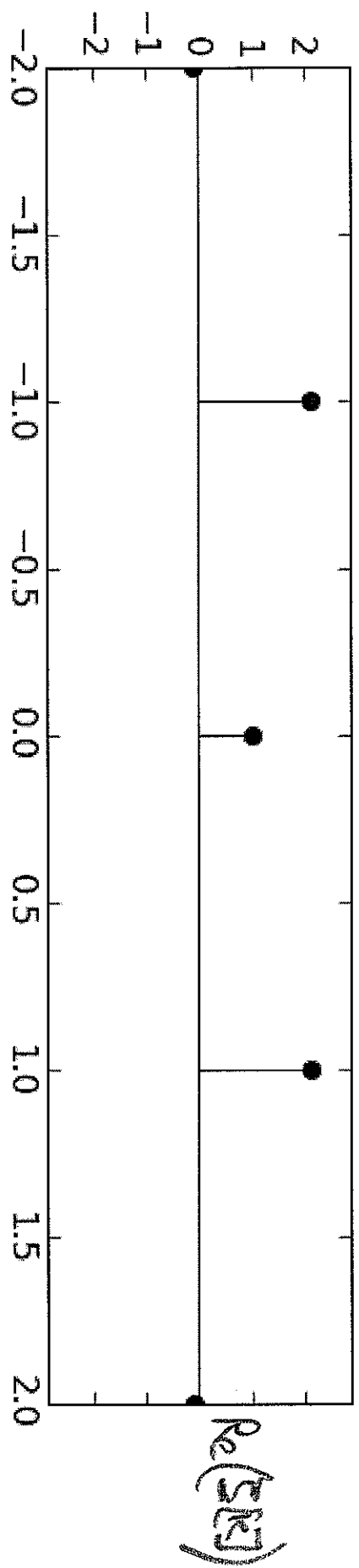
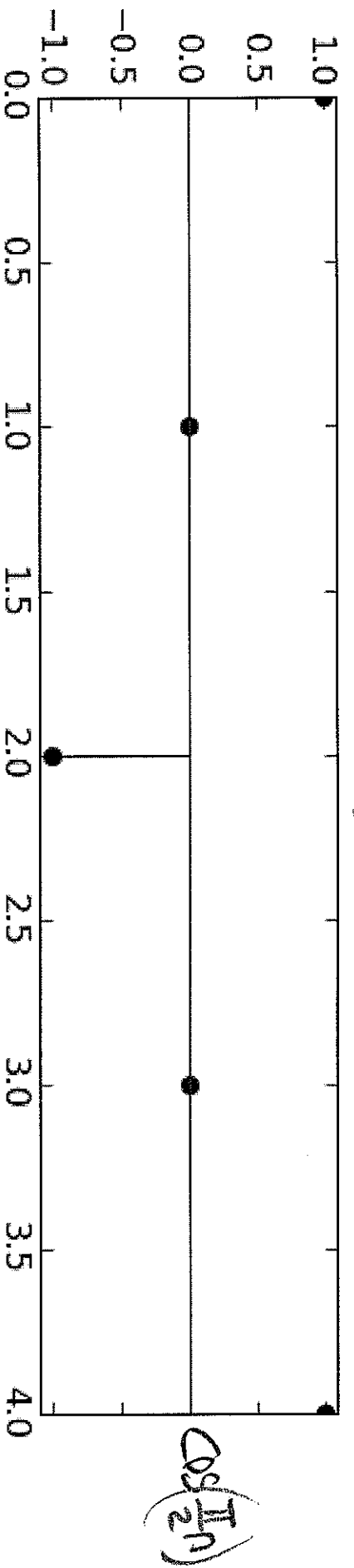
so  $\underline{S[0]} \neq 0$

IF  $N=7$

$$\underline{S[0]} = \sum_{n=0}^6 s[n] = 0$$

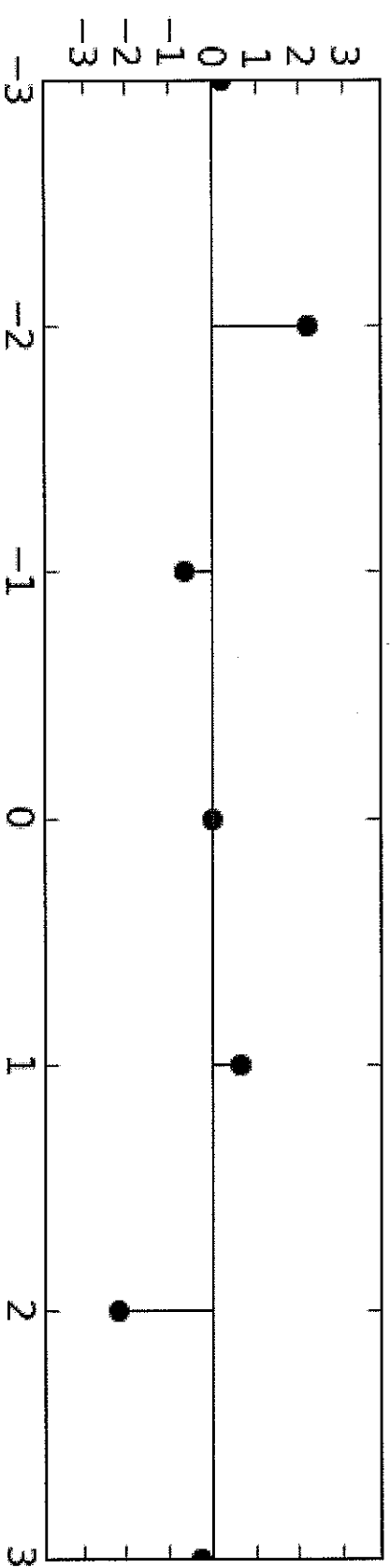
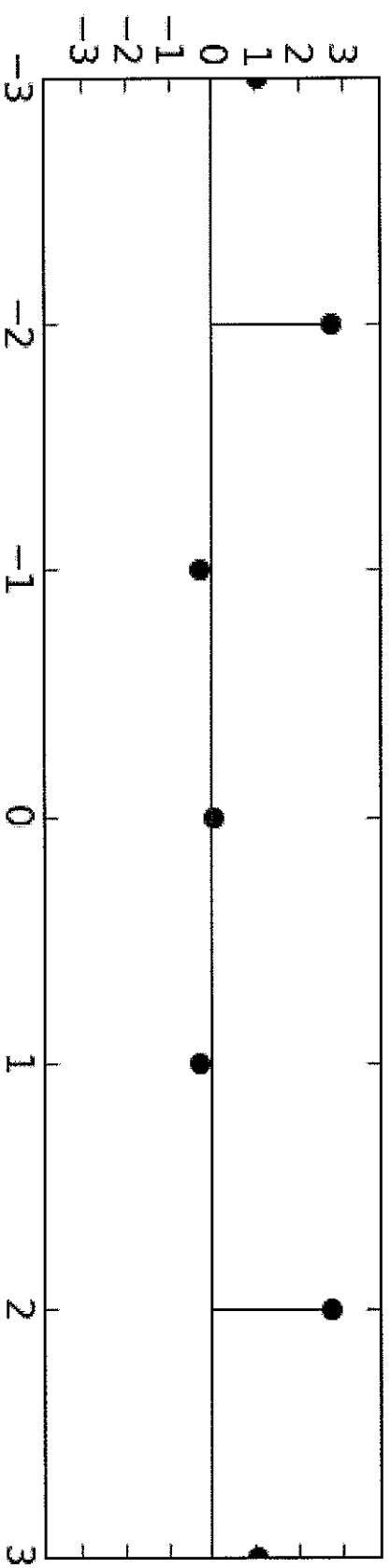
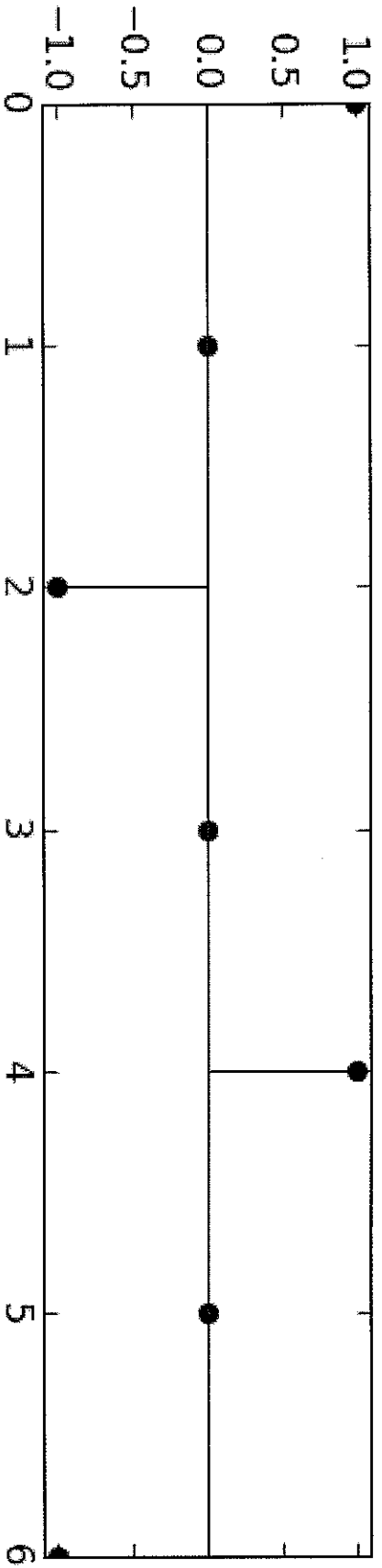
$N=5$

(7)



$N=7$

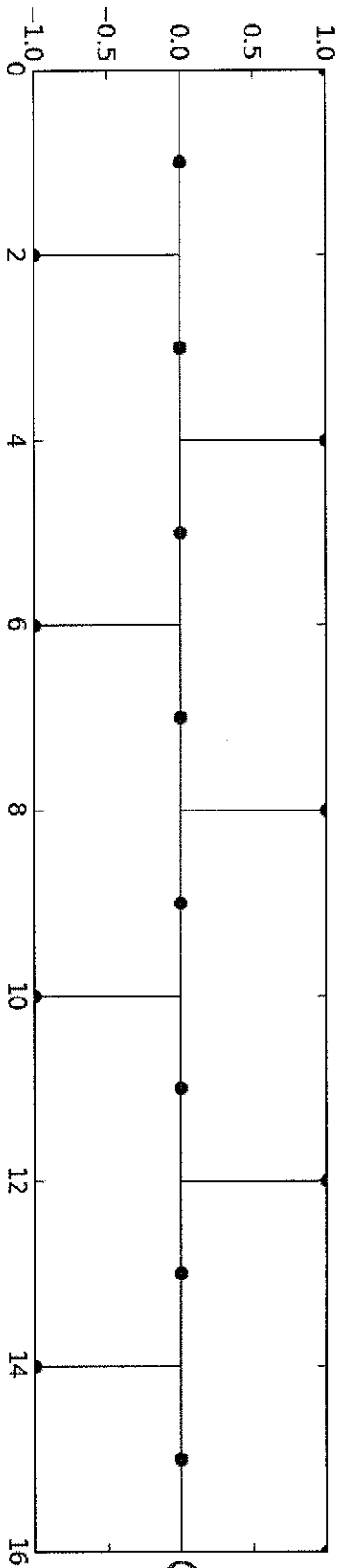
(6)



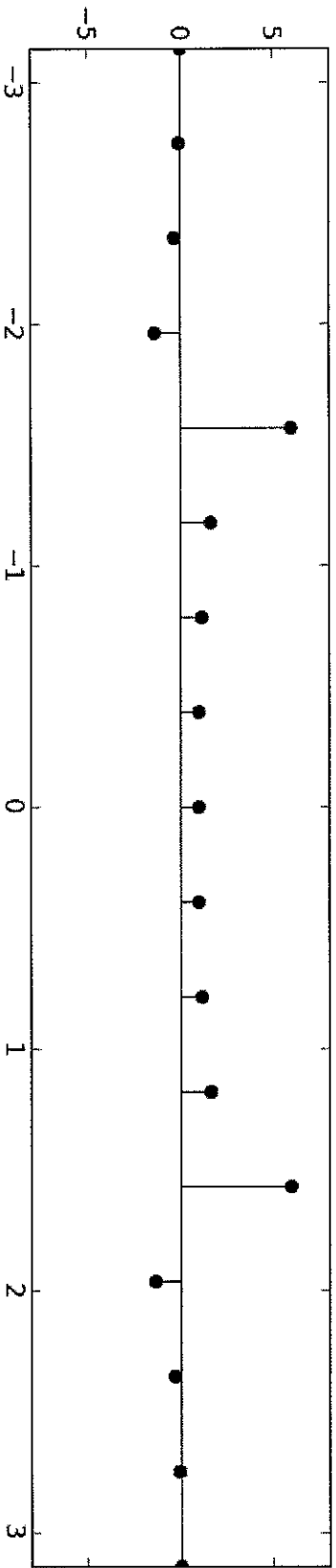


$N=17$

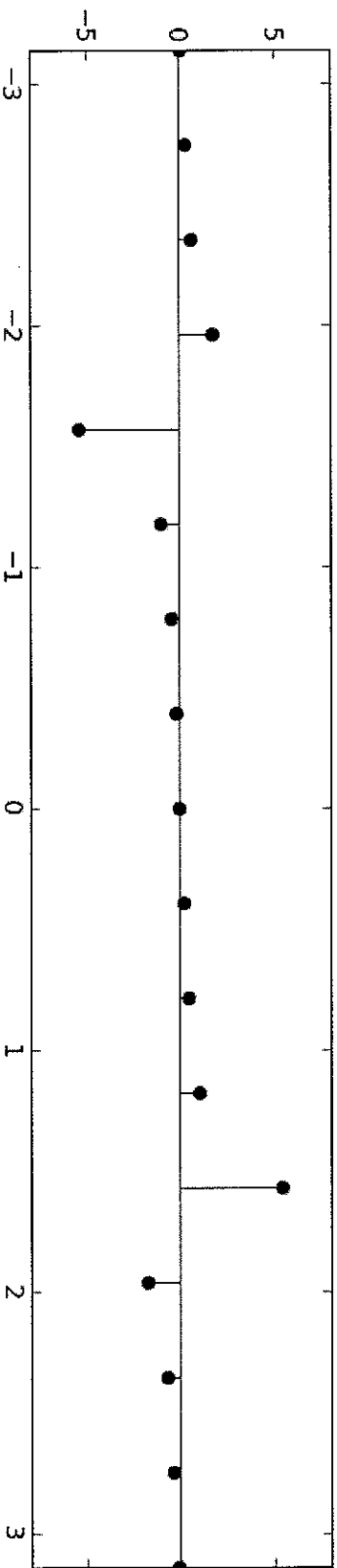
9



$\cos 2\pi kn$



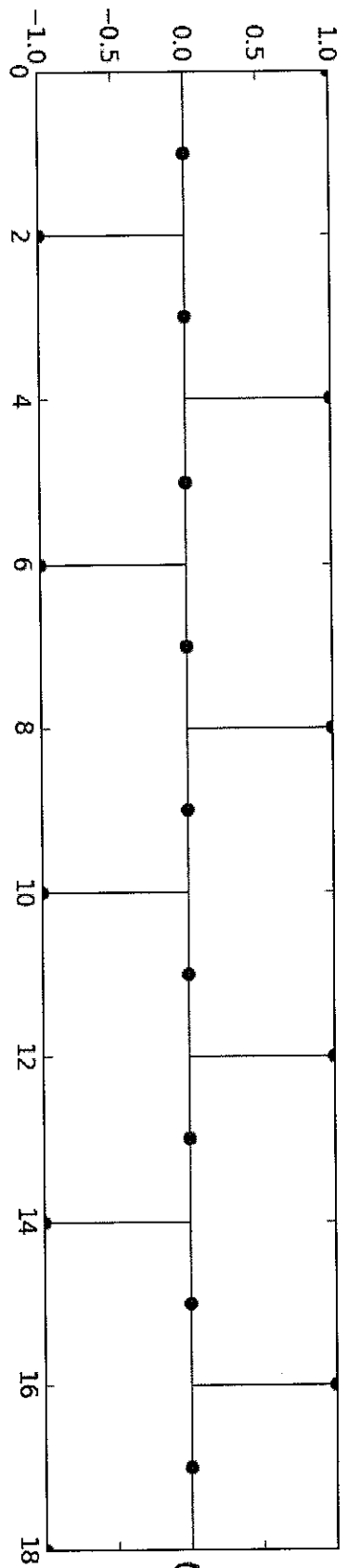
$\text{Re}\{S[k]\}$



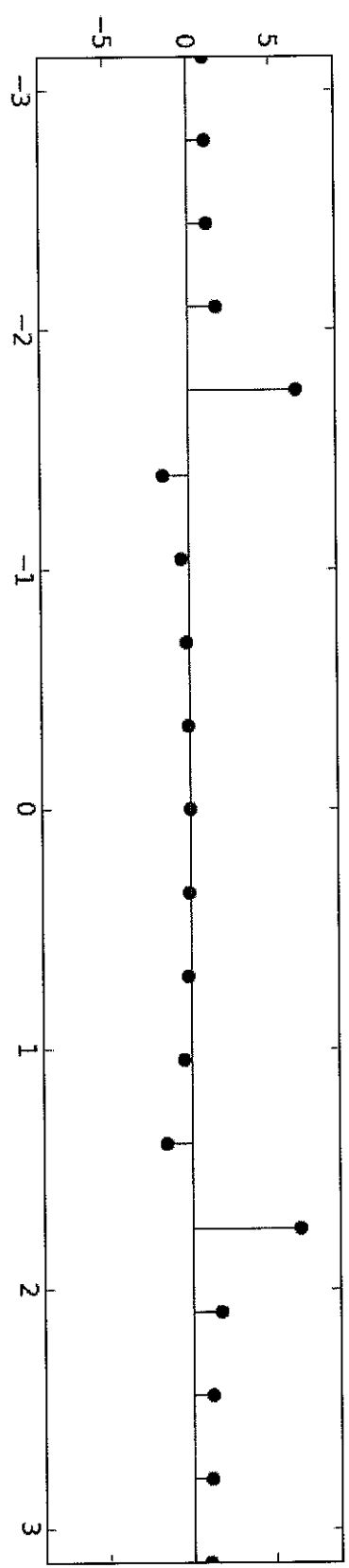
$\text{Im}\{S[k]\}$

$N = 19$

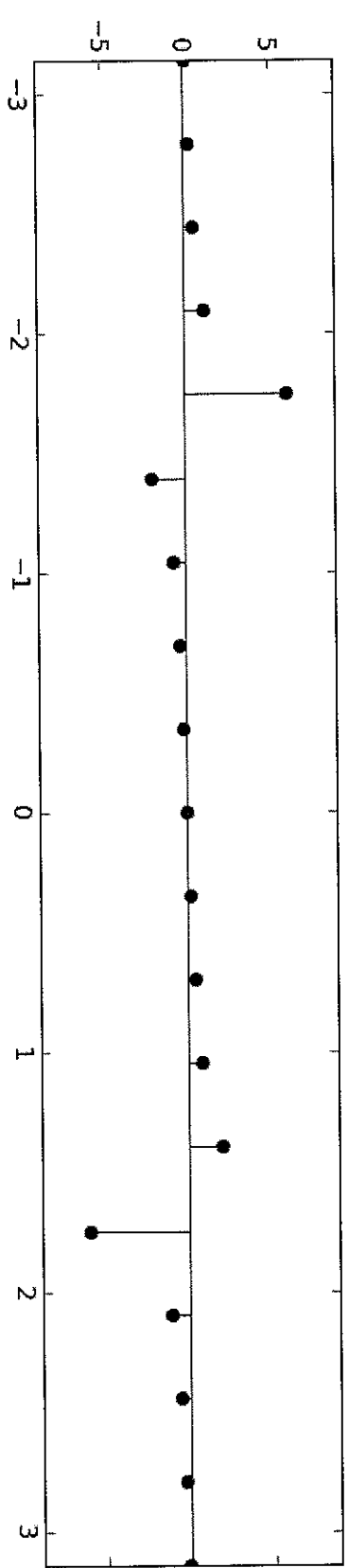
(10)



$\cos \frac{2k\pi}{N}$



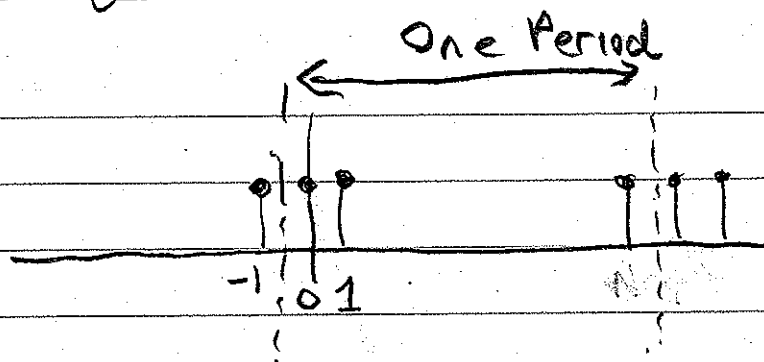
$\text{Re}(z^k)$



$\text{Im}(z^k)$

# Square Pulse Case

(11)



Suppose  $N = 11$

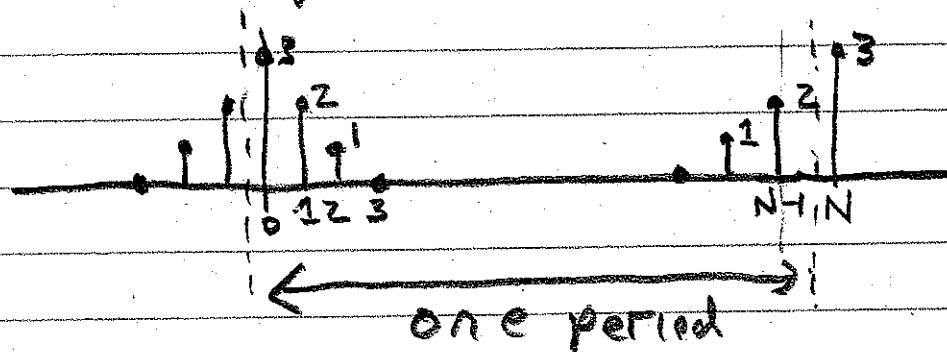
$$\begin{aligned} \bar{S}[k] &= s[0] \cdot 1 + s[1] \cdot e^{-j\frac{2\pi}{11}k \cdot 1} \\ &\quad + s[10] \cdot e^{-j\frac{2\pi}{11}k(10)} \\ &= 1 + e^{-j\frac{2\pi}{11}k} + \underbrace{e^{j\frac{2\pi}{11}k}}_{= e^{j\frac{2\pi}{11}k}} \\ &= 1 + 2\cos\left(\frac{2\pi}{11}k\right) \end{aligned}$$

For general  $N$

$$\bar{S}[k] = 1 + 2\cos\left(\frac{2\pi}{N}k\right)$$

(12)

## Triangular Pulse Case



$$\underline{N = 11}$$

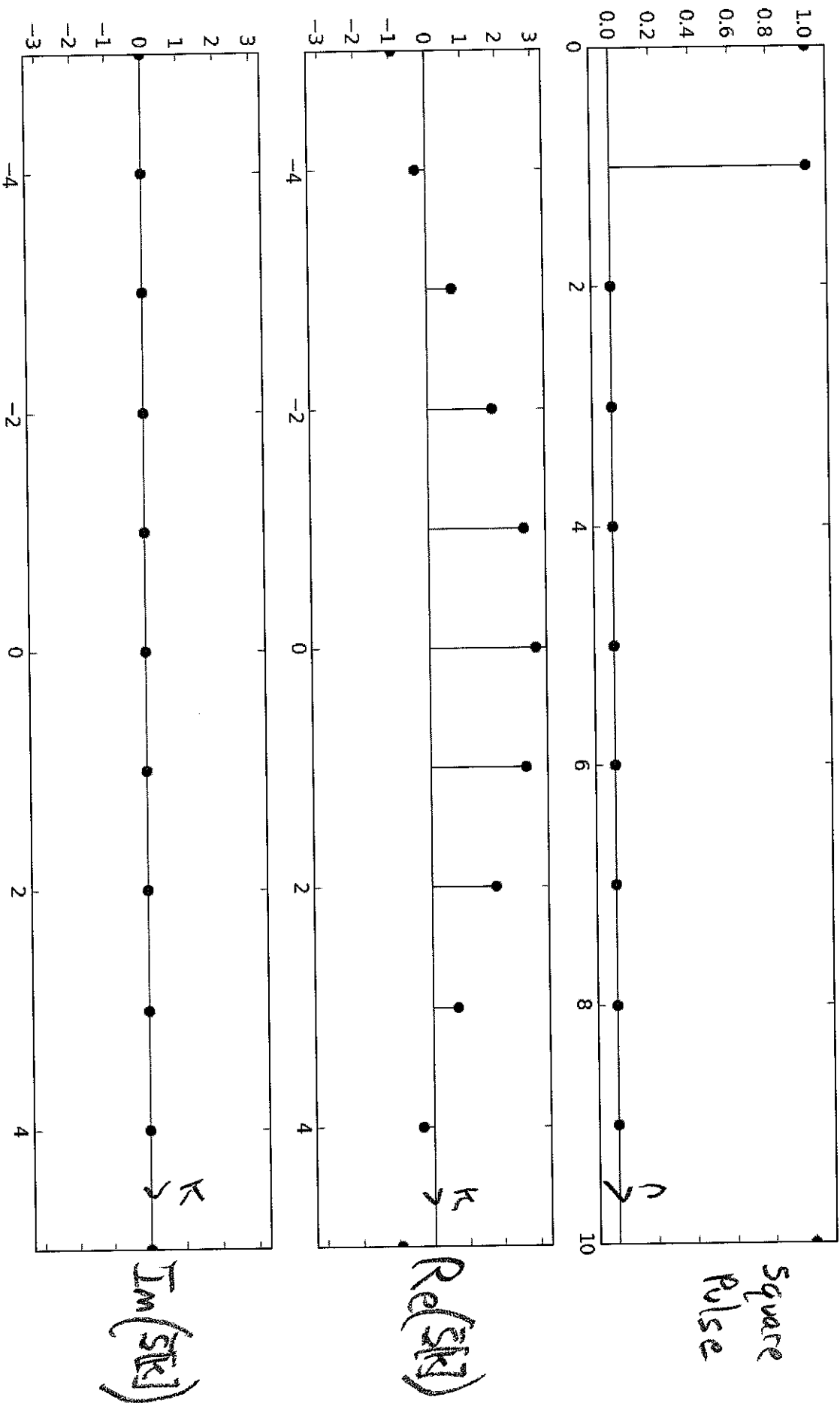
$$\begin{aligned} S[k] &= 3 \cdot 1 + 2 e^{-j \frac{2\pi}{11} k} + 1 e^{-j \frac{4\pi}{11} k} \\ &\quad + 1 e^{-j \frac{2\pi}{11} k} + 2 e^{-j \frac{2\pi}{11} k} \\ &= 3 + 4 \cos\left(\frac{2\pi}{11} k\right) + 2 \cos\left(\frac{4\pi}{11} k\right) \end{aligned}$$

Note: Compare to Product of Square Pulse  $S[k]$ 's

$$\begin{aligned} &\left(1 + e^{-j \frac{2\pi}{11} k} + e^{+j \frac{2\pi}{11} k}\right) \left(1 + e^{-j \frac{2\pi}{11} k} + e^{+j \frac{2\pi}{11} k}\right) \\ &= 3 + 2 e^{-j \frac{2\pi}{11} k} + e^{-j \frac{4\pi}{11} k} \\ &\quad + e^{+j \frac{4\pi}{11} k} + 2 e^{+j \frac{2\pi}{11} k} \end{aligned}$$

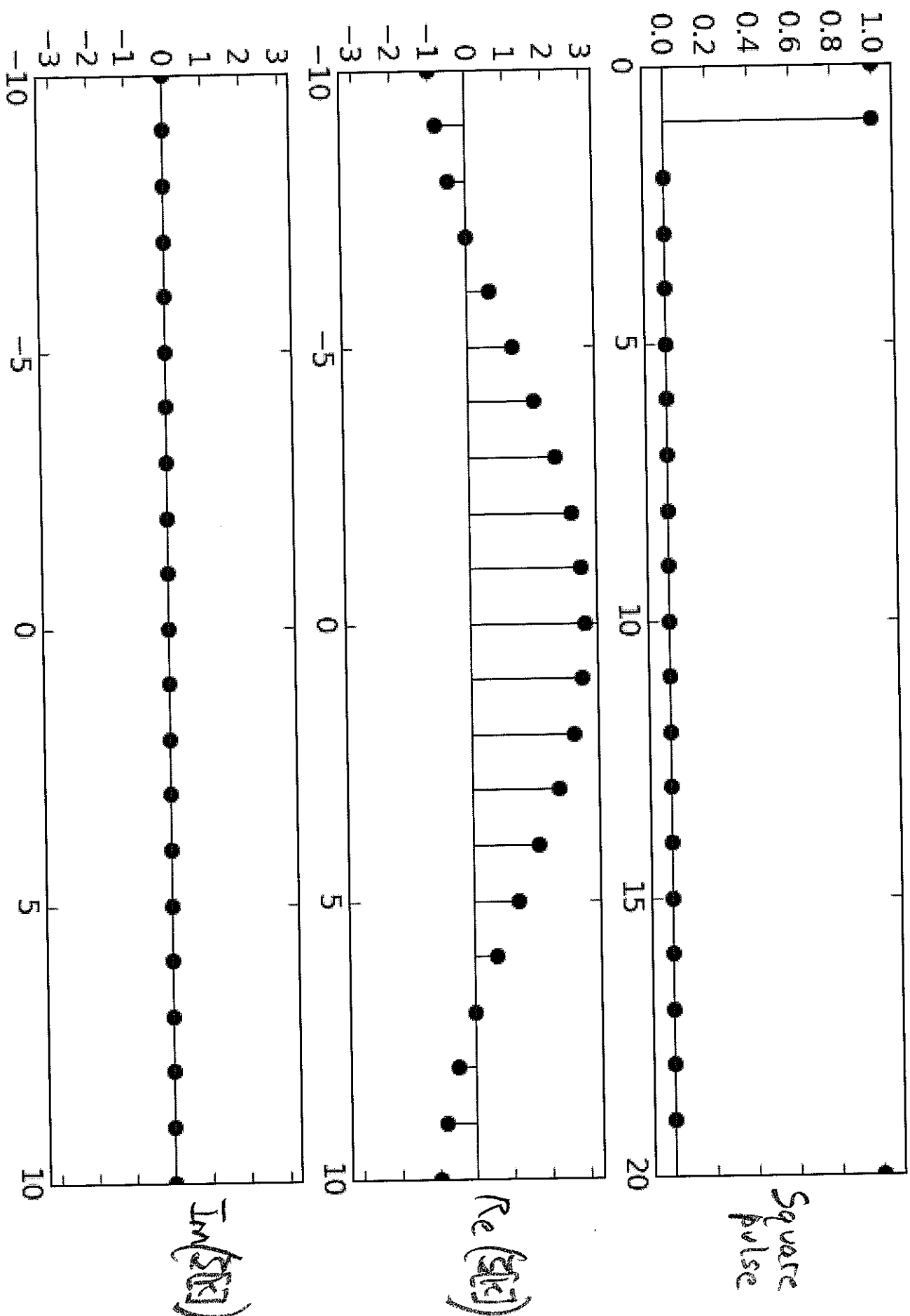
$N = 11$

12



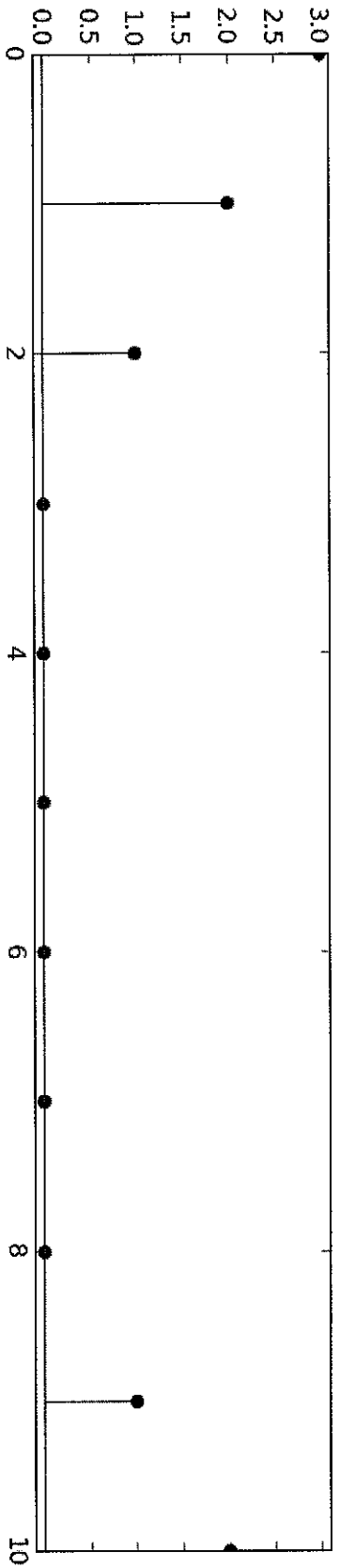
$N=21$

14

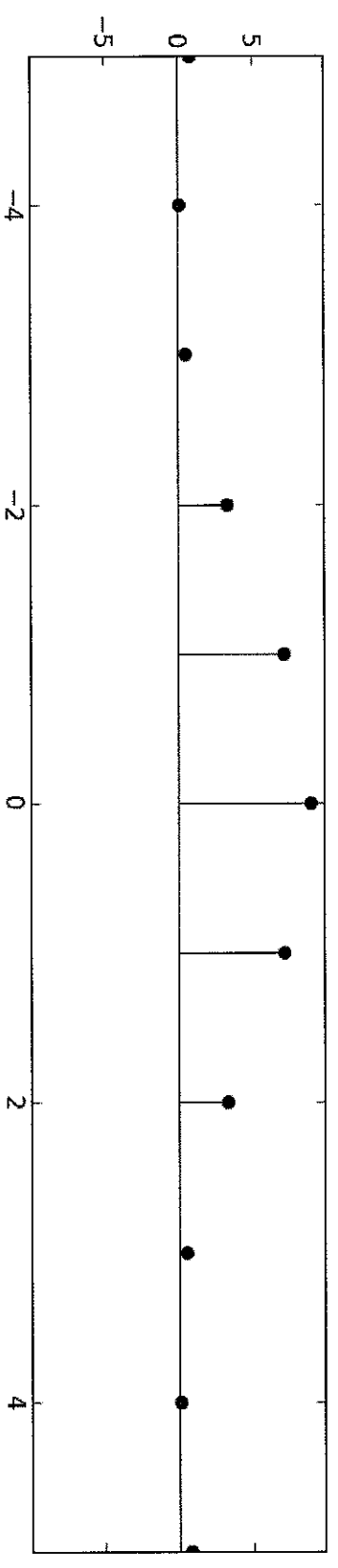


$N=11$

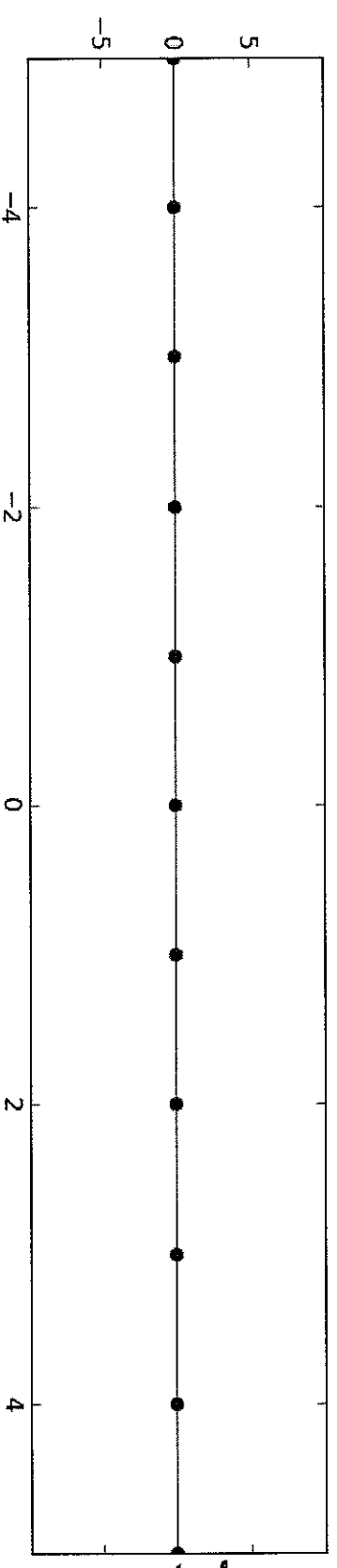
15



Triangular  
Pulse



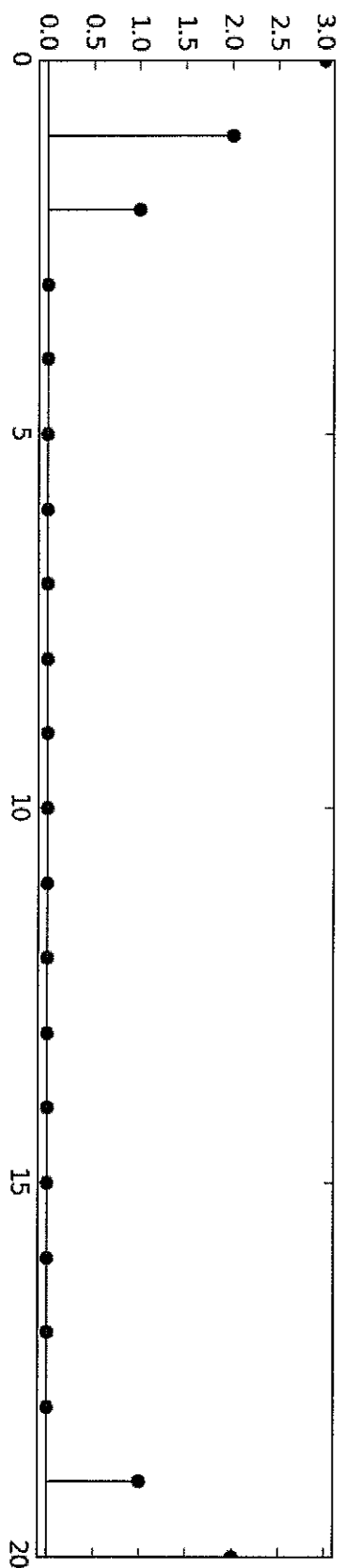
$Re(DFT)$



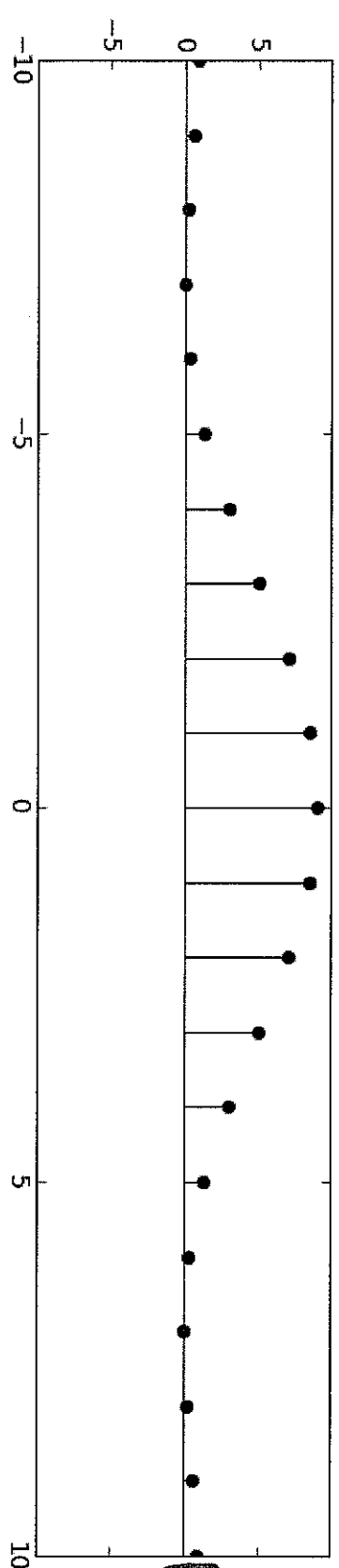
$Im(DFT)$

$N=21$

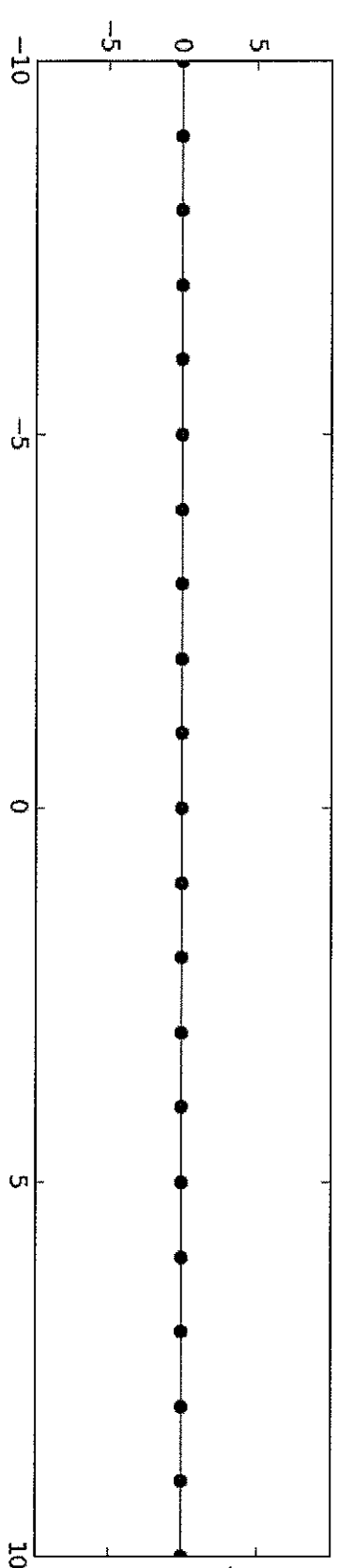
16



Transposed  
Pulse



$Re(S_k)$



$Im(S_k)$