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1. (a) From the definition of the frequency response,  $H(\Omega) = \sum_{m=-\infty}^{\infty} h[m]e^{-j\Omega m} = 1 + 2e^{-j\Omega} + e^{-2j\Omega}$ .
- (b) Plug in the given  $\Omega$  values into the expression for  $H(\Omega)$ , to get  $H(0) = 1, H(\pi/2) = 1 - 2j - 1 = -2j, H(\pi) = 1 - 2 + 1 = 0$ .
- (c) Let  $e^{-j\Omega} = z$ , so we want  $1 + 2z + z^2 = 0$ , which gives us  $(z + 1)^2 = 0$ , or  $z = -1$ . Hence,  $e^{-j\Omega} = -1$ , or  $\Omega = \pi$ .

2. (a) We'll use the fact that if the input to an LTI with frequency response  $H(\Omega)$  is a complex exponential  $x[n] = e^{j\Omega n}$ , then the output  $y[n] = H(\Omega)x[n] = H(\Omega)e^{j\Omega n}$ . Plugging that into the given equation or  $y[n]$ , we get  $H(z) = 1 + \alpha z + \beta z^2 + \gamma z^3$ , denoting  $e^{-j\Omega}$  by  $z$  for convenience (on both sides of the equation).

Let's now expand the given expression,  $H(\Omega) = 1 - 0.5e^{-j2\Omega} \cos \Omega$  in terms of complex exponentials. The idea is we can then match the coefficients to obtain the values of  $\alpha, \beta$ , and  $\gamma$ . Expanding the  $\cos(\Omega)$  term into complex exponential sums and multiplying, we get the expression  $H(z) = 1 - 0.25z - 0.25z^3$ . Matching coefficients, we conclude that  $\alpha = -0.25, \beta = 0, \gamma = -0.25$ .

- (b) For a series (cascade) of LTI systems, the frequency response is the *product* of the frequency responses of the individual LTI systems. So, the frequency response of this cascade is  $G(e^{j\Omega})H(e^{j\Omega})$ . For  $w[n]$  to be equal to  $x[n]$  for all  $n$ , the unit sample response of the cascade must be  $\delta[n]$ . The frequency response of the cascade is related the unit sample response by  $G(e^{j\Omega})H(e^{j\Omega}) = \sum_{m=0}^{\infty} \delta[m]e^{-j\Omega m} = 1$ . Since  $G(e^{j\Omega})H(e^{j\Omega}) = 1$ , and  $H(e^{j\Omega})$  is never zero for  $-\pi \leq \Omega \leq \pi$ , then  $G(e^{j\Omega}) = \frac{1}{H(e^{j\Omega})}$ . Hence,

$$G(e^{j\Omega}) = \frac{1}{1 - 0.5e^{-j2\Omega} \cos \Omega}.$$

- (c) First, observe that we can rewrite the given  $x[n]$  as

$$x[n] = e^{j0n} + 0.5e^{j\pi n}.$$

Then using the meaning of frequency response, and applying superposition, we can write

$$y[n] = 0e^{j0n} + Ae^{j\pi n} = H(e^{j0})e^{j0n} + H(e^{j\pi n})0.5e^{j\pi n}.$$

Matching terms, it then follows that  $H(e^{j0}) = 0$  and  $A = 0.5H(e^{j\pi n})$ . For this difference equation, the frequency response for an arbitrary  $\Omega$  is given by

$$H(e^{j\Omega}) = 1 + \alpha e^{-j\Omega} + \beta e^{-j2\Omega} + \gamma e^{-j3\Omega},$$

so for the special case of  $\Omega = 0$ ,

$$H(e^{j0}) = 1 + \alpha + \beta + \gamma.$$

Given that  $H(e^{j0}) = 0$ , and  $\alpha = \gamma = 1$ , it must be true that  $\beta = -3$ . Then to determine  $A$ , consider that

$$H(e^{j\pi n}) = 1 + \alpha e^{-j\pi} + \beta e^{-j2\pi} + \gamma e^{-j3\pi} = 1 - \alpha + \beta - \gamma = 1 - 1 - 3 - 1 = -4.$$

Therefore,  $A = 0.5H(e^{j\pi n}) = 0.5(-4) = -2$ .

3. (a) From the definition of the frequency response,  $H(\Omega) = \sum_{m=0}^{\infty} h[m]e^{-j\Omega m} = h[0]e^{-j\Omega 0} + h[1]e^{-j\Omega} + h[2]e^{-j2\Omega} + h[3]e^{-j3\Omega}$ . Plugging in values for  $h[i]$ , the frequency response is  $H(\Omega) = a + be^{-j\Omega} + be^{-j2\Omega} + ae^{-j3\Omega}$ .
- (b) Note that  $x[n] = (-1)^n$  can be written as  $x[n] = e^{-j\pi n}$ . This input corresponds to a complex exponential with angular frequency  $\Omega = \pi$ . We know that for everlasting complex exponential inputs,  $y[n] = H(\Omega)e^{j\Omega n}$ ; therefore we know that  $y[n] = H(\pi)e^{j\pi n}$ . Evaluating  $H(\Omega)$ ,  $y[n]$  is given by  $(a - b + b - a)$ , or  $y[n] = 0$  for all  $n$ .
- (c) The definition of convolution follows as  $y[n] = \sum_{m=0}^{\infty} h[m]x[n-m]$ . We find  $y[5]$  by solving  $\sum_{m=0}^{\infty} h[m]x[5-m]$ ; this gives  $-a + b - b + a = 0$ . Similarly,  $y[6] = \sum_{m=0}^{\infty} h[m]x[6-m]$ , giving  $a - b + b - a = 0$ . Moreover, since these calculations are representative of what one would get for odd and even  $n$  respectively,  $y[n] = 0$  for all  $n$ .
- (d) Decompose into cosines. Dividing both sides by  $e^{-j3\Omega/2}$  we know that  $G(\Omega) = \frac{1}{e^{-j3\Omega/2}}H(\Omega)$ . Dividing out  $e^{-j3\Omega/2}$  from each term in  $H(\Omega)$ , we get  $ae^{j3\Omega/2} + be^{j\Omega/2} + be^{-j\Omega/2} + ae^{j3\Omega/2}$ . Converting this expression to cosines, we find that

$$G(\Omega) = 2[a \cos(\frac{3\Omega}{2}) + b \cos(\frac{\Omega}{2})].$$

- (e) By superposition, the response is the sum of the responses to  $(-1)^n$  and to  $\cos(\frac{\pi}{2}n + \theta_0)$ . But we have already seen in parts (b) and (c) that the response to  $(-1)^n$  is 0 for all  $n$ . So what remains is to find the response to  $\cos(\pi n + \theta_0)$ . This response is simply

$$y[n] = |H(\pi/2)| \cos(\frac{\pi}{2}n + \theta_0 + \angle H(\pi/2)).$$

Note from the form of  $H(\Omega)$  in part (d) that

$$\begin{aligned} H(\pi/2) &= G(\pi/2)e^{-j3\pi/4} = 2[a \cos(\frac{3\pi}{4}) + b \cos(\frac{\pi}{4})] \\ &= \sqrt{2}(b - a)e^{-j3\pi/4} \end{aligned}$$

Since  $b > a$ , it follows that  $|H(\pi/2)| = \sqrt{2}(b - a)$  and  $\angle H(\pi/2) = -3\pi/4$ . Hence,

$$y[n] = \sqrt{2}(b - a) \cos\left(\frac{\pi}{2}n + \theta_0 - \frac{3\pi}{4}\right).$$

As a check, consider the case  $\theta_0 = 0$ , so the input of interest is  $\cos(\pi n)$ , which alternates between  $+1$  and  $-1$  at even values of  $n$ , and is 0 at all odd values of  $n$ . Convolution with the given unit sample response  $h[n]$  shows that  $y[0] = a - b$ . And this answer matches what we get on evaluating the above general expression for  $y[n]$  at  $n = 0$  and  $\theta_0 = 0$ .

4. (a) There are two key things to notice about this frequency response function. First off, it has no zeros. Second, looking at the denominator, when  $\Omega \approx \pm \frac{\pi}{2}$  the denominator becomes very small, causing  $H_A(e^{j\Omega})$  to become large. The only graph that satisfies these two criteria is  $H_I$ .
- (b) We can solve this problem by setting  $x[n] = e^{j\Omega n}$  and knowing that the frequency response is a complex exponential with the same frequency. We get

$$H(\Omega)e^{j\Omega n} + a_1H(\Omega)e^{j\Omega(n-1)} + a_2H(\Omega)e^{j\Omega(n-2)} = e^{j\Omega n}.$$

Dividing both sides by  $e^{j\Omega n}$  and solving for  $H(\Omega)$ , we get

$$H(\Omega) = \frac{1}{1 + a_1e^{-j\Omega} + a_2e^{-j2\Omega}}.$$

It is easier to expand  $H_A(\Omega)$  than to factor  $H(\Omega)$ , so we factor  $H_A(\Omega)$  to find the denominator is  $1 + 0.95e^{j(\Omega+\frac{\pi}{2})} + 0.95e^{j(\Omega-\frac{\pi}{2})} + 0.95e^{j(\Omega-\frac{\pi}{2})}0.95e^{j(\Omega+\frac{\pi}{2})}$ . The middle two terms cancel each other out (because of the  $\frac{\pi}{2}$  phase shifts) and the last term is equal to  $0.9025e^{-j2\Omega}$ . Matching coefficients with  $H(\Omega)$ , we find that  $a_1 = 0$  and  $a_2 = \left(\frac{19}{20}\right)^2 = 0.9025$ .

5. (a) We can make the following observations:

$$\begin{aligned} \max_n \left(\cos \frac{\pi}{6}n\right) &= 1, n = 0, 12, \dots \\ \max_n \left(\cos \frac{5\pi}{6}n\right) &= 1, n = 0, 12, \dots \\ \max_n \left(3(-1)^n\right) &= 1, n \text{ even.} \end{aligned}$$

Hence, the maximum value is 7, and the smallest *positive*  $n$  at which the maximum occurs is  $n = 12$ .

- (b) The frequency response at  $\Omega = \frac{\pi}{6}$  and at  $\Omega = \frac{5\pi}{6}$  must be zero, which means that the only possibility is  $H_{III}$ .

$$y[n] = H(0) \cdot 2 \cdot (1)^n + H(\pi) \cdot 3 \cdot (-1)^n = 4 \cdot 2 \cdot (1)^n + 4 \cdot 3 \cdot (-1)^n.$$

The numerical value of  $M$  is 4.

6. (a) The first equation tells us that the DC component of the frequency response, i.e.,  $H(0)$  is 5. The second and third equations tell us that  $H(\pi/2)$  and  $H(-\pi/2)$  are both 0.
- (b)  $H_{II}$  is the frequency response that best describes the above information, as it is the only curve that meets the constraints of part (a). The numerical value of  $M$  must be 5.
- (c)  $y[n]/x[n]$  is the frequency response  $H(\Omega)$ . Because the input is an everlasting exponential with frequency  $\Omega = \frac{\pi}{6}$ ,  $y[n]/x[n]$  is simply  $H(\frac{\pi}{6}) = h[0] + h[2]e^{-j\pi/3} = 3.75 - j(5\sqrt{3}/4)$ .