

INTRODUCTION TO EECS II
DIGITAL COMMUNICATION SYSTEMS

**6.02 Spring 2012
Lecture #7**

- Viterbi decoding of convolutional codes
Path and branch metrics
Hard-decision & soft-decision decoding
- Performance issues: decoder complexity, post-decoding BER, “free distance” concept

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Encoding & Decoding Convolutional Codes

- Transmitter (aka Encoder)
 - Beginning at starting state, processes message bit-by-bit
 - For each message bit: makes a state transition, sends $p_0p_1\dots$
 - Pad message with $K-1$ zeros to ensure return to starting state
- Receiver (aka Decoder)
 - Doesn't have direct knowledge of transmitter's state transitions; only knows (possibly corrupted) received parity bits, p_i
 - Must find **most likely sequence of transmitter states** that could have generated the received parity bits, p_i
 - If $BER < 1/2$, $P(\text{more errors}) < P(\text{fewer errors})$
 - *When $BER < 1/2$, maximum-likelihood message sequence is the one that generated the codeword (here, sequence of parity bits) with the smallest Hamming distance from the received codeword (here, parity bits)*
 - I.e., find nearest valid codeword *closest* to the received codeword – Maximum-likelihood (ML) decoding

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Example

- For the code
 $p_0 = x[n] + x[n-1] + x[n-2]$
 $p_1 = x[n] + x[n-1]$
- Received:
111011000110
- Some errors have occurred...
- What's the 4-bit message?
- Look for message whose codeword is closest to rcvd bits

Msg	Codeword	Received	Hamming distance
0000	000000000000	111011000110	7
0001	000000111110		8
0010	000011111000		8
0011	000011010110		4
0100	001111100000		6
0101	001111011110		5
0110	001101001000		7
0111	001100100110		6
1000	111110000000		4
1001	111110111110		5
1010	111101111000		7
1011	111101000110		2
1100	110001100000		5
1101	110001011110		4
1110	110010011000		6
1111	110010100110		3

Most likely: 1011

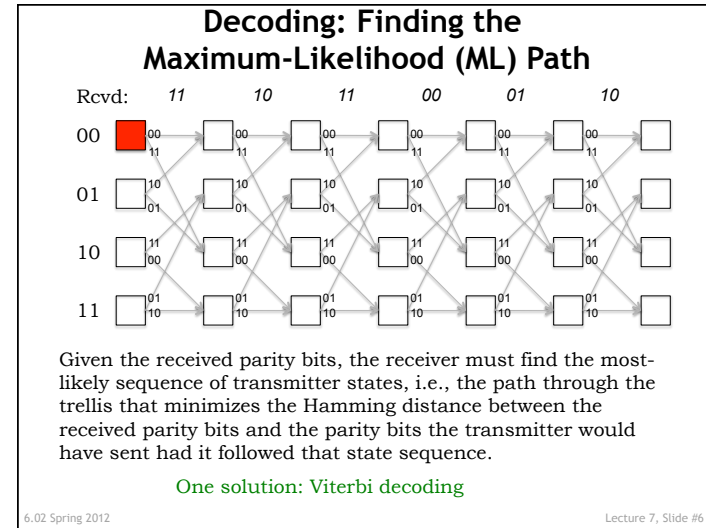
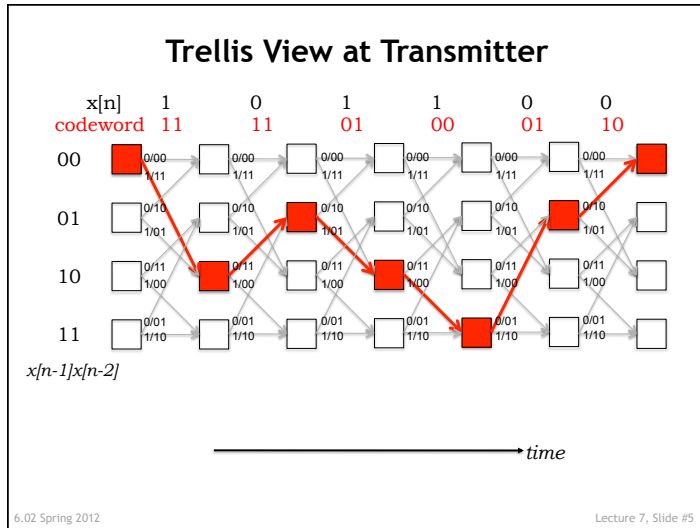
Initial state: 00

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Key Concept for Decoding: A Trellis

- Example: $K=3$, rate- $1/2$ convolutional code
 - $G_0 = 111$: $p_0 = 1*x[n] + 1*x[n-1] + 1*x[n-2]$
 - $G_1 = 110$: $p_1 = 1*x[n] + 1*x[n-1] + 0*x[n-2]$
- States labeled with $x[n-1] x[n-2]$
- Arcs labeled with $x[n] / p_0p_1$

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Given the received parity bits, the receiver must find the most-likely sequence of transmitter states, i.e., the path through the trellis that minimizes the Hamming distance between the received parity bits and the parity bits the transmitter would have sent had it followed that state sequence.

One solution: Viterbi decoding

Viterbi Algorithm

- Want: Most likely message sequence
- Have: (possibly corrupted) received parity sequences
- Viterbi algorithm for a given K and r :
 - Works incrementally to compute most likely message sequence
 - Uses two metrics
- Branch metric: $BM(x_{mit}, r_{cvd})$ proportional to likelihood that transmitter sent x_{mit} given that we've received r_{cvd} .
 - "Hard decision": use digitized bits, compute Hamming distance between x_{mit} and r_{cvd} . Smaller distance is more likely if $BER < 1/2$
 - "Soft decision": use function of received voltages directly
- Path metric: $PM[s, i]$ for each state s of the 2^{K-1} transmitter states and bit time i where $0 \leq i < \text{len}(\text{message})$.
 - $PM[s, i]$ = most likely sum of $BM(x_{mit_m}, \text{received parity})$ over all message sequences m that place transmitter in state s at time i
 - $PM[s, i+1]$ computed from $PM[s, i]$ and $p_0[i], \dots, p_{r-1}[i]$

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Hard-decision Branch Metric

- $BM = \text{Hamming distance}$ between expected parity bits and received parity bits
- Compute BM for each transition arc in trellis
 - Example: received parity = 00
 - $BM(00, 00) = 0$
 - $BM(01, 00) = 1$
 - $BM(10, 00) = 1$
 - $BM(11, 00) = 2$
- Will be used in computing $PM[s, i+1]$ from $PM[s, i]$.
- We want to use the most likely BM , which, means minimum BM .

Time: i 00 $i+1$

State

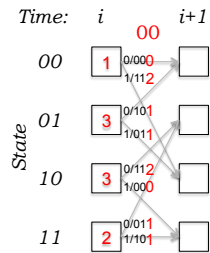
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Computing PM[s,i+1]

Starting point: we've computed PM[s,i], shown graphically as label in trellis box for each state at time i.

Example: $PM[00,i] = 1$ means there was 1 bit error detected when comparing received parity bits to what would have been transmitted when sending the most likely message, considering all messages that place the transmitter in state 00 at time i.

Q: What's the most likely state s for the transmitter at time i?
 A: state 00 (smallest PM[s,i])



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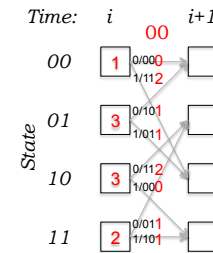
Computing PM[s,i+1] cont'd.

Q: If the transmitter is in state s at time i+1, what state(s) could it have been in at time i?

A: For each state s, there are two predecessor states α and β in the trellis diagram

Example: for state 01, $\alpha=10$ and $\beta=11$.

Any message sequence that leaves the transmitter in state s at time i+1 must have left the transmitter in state α or state β at time i.



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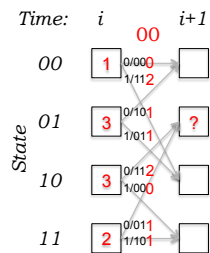
Computing PM[s,i+1] cont'd.

Example cont'd: to arrive in state 01 at time i+1, either

1) The transmitter was in state 10 at time i and the i^{th} message bit was a 0. If that's the case, the transmitter sent 11 as the parity bits and there were 2 bit errors since we received 00. Total bit errors = $PM[10,i] + 2 = 5$ OR

2) The transmitter was in state 11 at time i and the i^{th} message bit was a 0. If that's the case, the transmitter sent 01 as the parity bits and there was 1 bit error since we received 00. Total bit errors = $PM[11,i] + 1 = 3$

Which is more likely?



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Computing PM[s,i+1] cont'd.

Formalizing the computation:

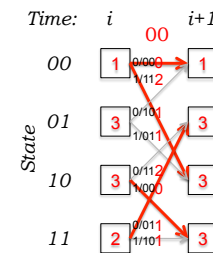
$$PM[s,i+1] = \min(PM[\alpha,i] + BM[\alpha \rightarrow s], PM[\beta,i] + BM[\beta \rightarrow s])$$

Example:

$$PM[01,i+1] = \min(PM[10,i] + 2, PM[11,i] + 1) = \min(3 + 2, 2 + 1) = 3$$

Notes:

- Remember which arc was min; saved arcs will form a path through trellis
- If both arcs have same sum, break tie arbitrarily (e.g., when computing $PM[11,i+1]$)



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Finding the Maximum-Likelihood Path

Rcvd: 11 10 11 00 01 10

- Path metric: number of errors on maximum-likelihood path to given state (min of all paths leading to state)
- Branch metric: for each arrow, the Hamming distance between received parity and expected parity

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Viterbi Algorithm

Rcvd: 11 10 11 00 01 10

- Compute branch metrics for next set of parity bits
- Compute path metric for next column
 - add branch metric to path metric for old state
 - compare sums for paths arriving at new state
 - select path with smallest value (fewest errors, most likely)

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Example (cont' d.)

Rcvd: 11 10 11 00 01 10

- After receiving 3 pairs of parity bits we can see that all ending states are equally likely
- Power of convolutional code: use future information to constrain choices about most likely events in the past

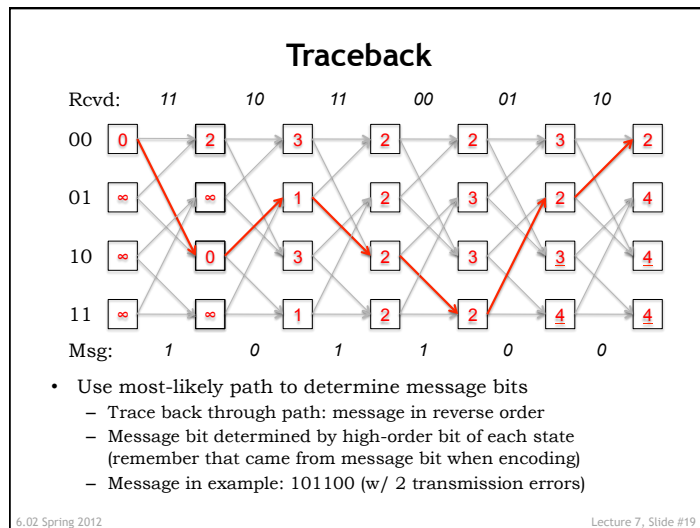
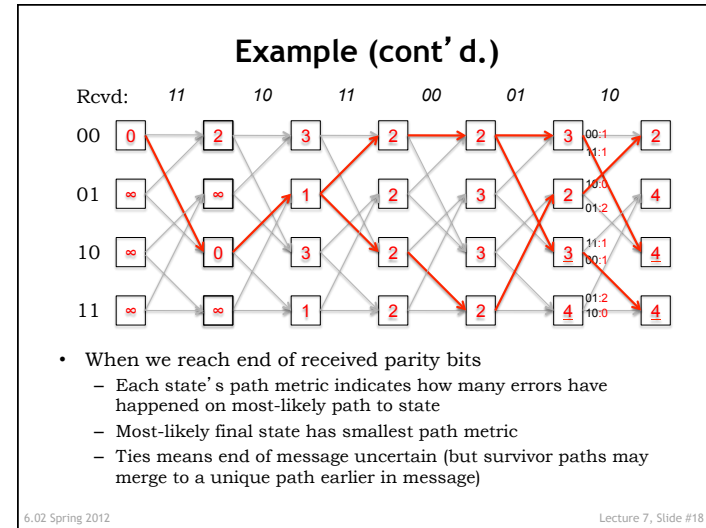
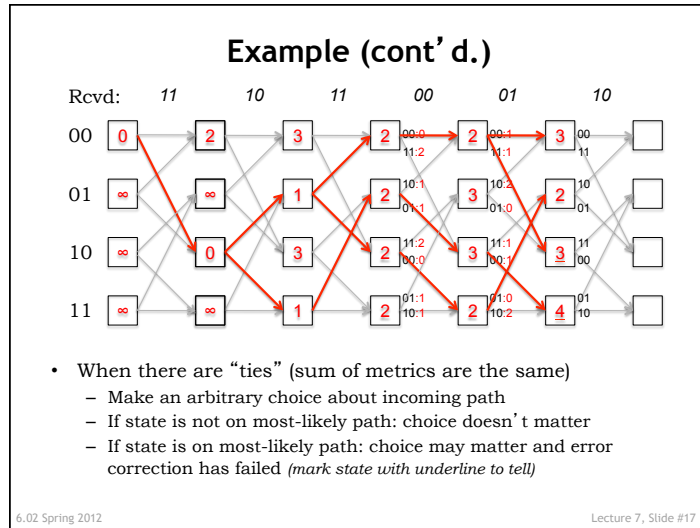
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Survivor Paths

Rcvd: 11 10 11 00 01 10

- Notice that some paths don't continue past a certain state
 - Will not participate in finding most-likely path: eliminate
 - Remaining paths are called *survivor paths*
 - When there's only one path: we've got a message bit!

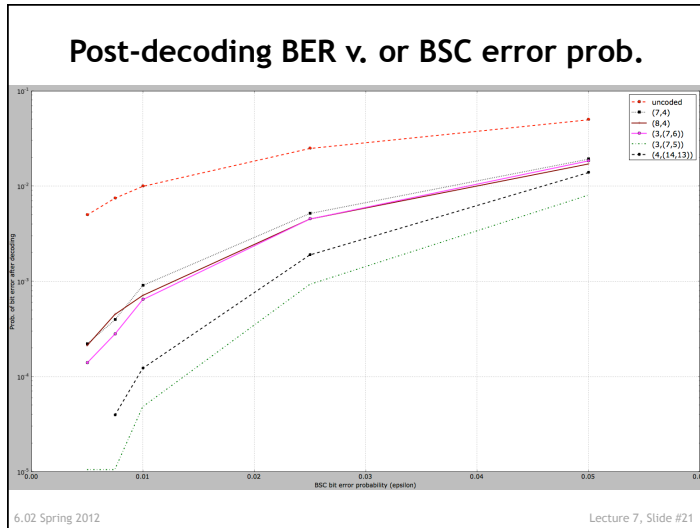
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Viterbi Algorithm with Hard Decisions

- Branch metrics measure the likelihood by comparing received parity bits to possible transmitted parity bits computed from possible messages.
- Path metric $PM[s,i]$ proportional to likelihood of transmitter being in state s at time i , assuming the mostly likely message of length i that leaves the transmitter in state s .
- Most likely message? The one that produces the most likely $PM[s,N]$.
- At any given time there are 2^{K-1} most-likely messages we're tracking \rightarrow time complexity of algorithm grows exponentially with constraint length K .

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Hard Decisions

- As we receive each bit it's immediately digitized to "0" or "1" by comparing it against a threshold voltage
 - We lose the information about how "good" the bit is: a "1" at .9999V is treated the same as a "1" at .5001V
- The branch metric used in the Viterbi decoder is the Hamming distance between the digitized received voltages and the expected parity bits
 - This is called *hard-decision decoding*
- Throwing away information is (almost) never a good idea when making decisions
 - Can we come up with a better branch metric that uses more information about the received voltages?

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Soft-Decision Decoding

- In practice, the receiver gets a voltage level, V , for each received parity bit
 - Sender sends V_0 or V_1 volts; V in $(-\infty, \infty)$ assuming additive Gaussian noise
- Idea: Pass received voltages to decoder **before** digitizing
- Define a "soft" branch metric as the square of the Euclidean distance between received voltages and expected voltages

- Soft-decision decoder chooses path that minimizes sum of the squares of the Euclidean distances between received and expected voltages

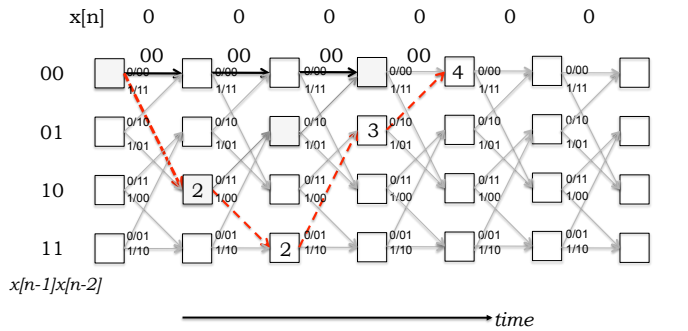
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What determines the "goodness" of a convolutional code?

- How much error correcting capability do we get from a convolutional code?
- In general, larger values of K and r (the number of parity streams or generators) provide higher error tolerance
- But what determines the error correction ability? (I.e., what's the equivalent of the Hamming distance?)
- Answer: With hard-decision decoding, it is the *free distance* of the code

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Free Distance of a Convolutional Code



The free distance is the difference in path metrics between the all-zeroes output and the path with the smallest non-zero path metric going from the initial 00 state to some future 00 state.