

INTRODUCTION TO EECS II
DIGITAL COMMUNICATION SYSTEMS

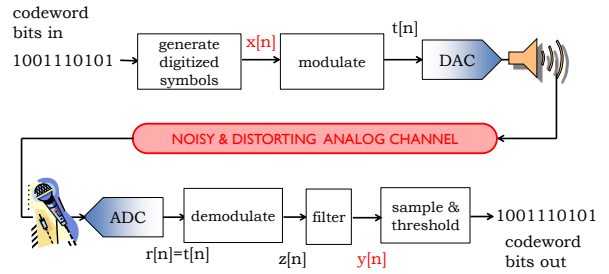
**6.02 Spring 2012
 Lecture #10**

- Input/output descriptions of systems
- Linear time-invariant (LTI) systems
- Constructing LTI system responses to input signals

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Lecture 10, Slide #1

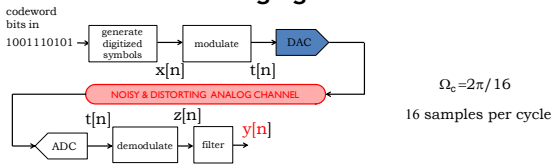
From Bits to Modulated Signal, and Back



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Lecture 10, Slide #2

Averaging filter



$y[n] = (z[n] + \dots + z[n-L]) / (L+1)$, $L+1$ length of the averaging filter

For $L+1=8$, $2\Omega_c$ component is at $2\pi/8$, which is 8 samples per cycle

So, the $\sum \cos(2\pi/8*(n-k)) = 0$ for $k=0, \dots, L$ and the $2\Omega_c$ gets integrated out*

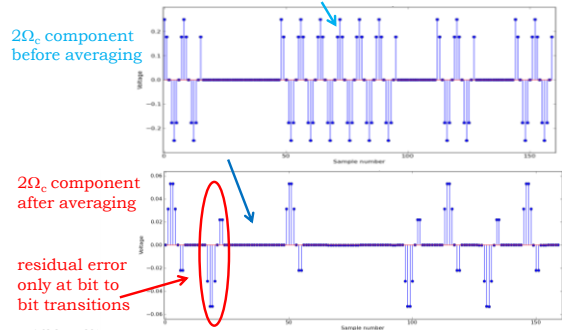
* $2\Omega_c$ component gets integrated out in regions where $x[n]$ is constant for $L+1$ samples. At transitions, there is a bit of degradation, but we make decisions on the middle samples

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Lecture 10, Slide #3

Averaging filter in action

$z[n] = 0.5x[n] + 0.5x[n]\cos(2\Omega_c n)$

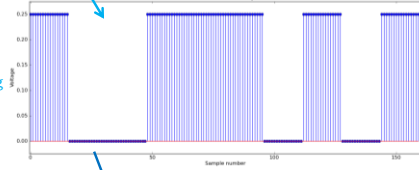


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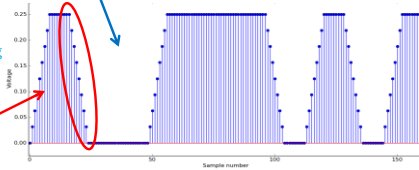
Averaging filter in action

$$z[n] = 0.5x[n] + 0.5x[n] \cos(2\Omega_c n)$$

desired component before averaging



desired component after averaging

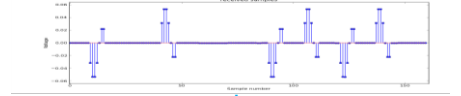


residual error only at bit to bit transitions

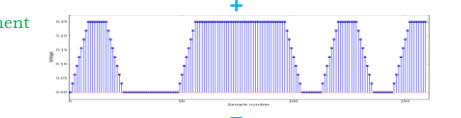
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Averaging filter in action

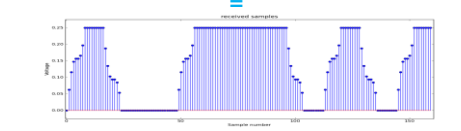
$2\Omega_c$ component after averaging



Desired component after averaging



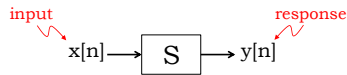
Output of the averaging filter



Need to learn a bit more about Linear Time Invariant (LTI) systems to design a better filter!

10, Slide #6

Filter design: System Input and Output



A discrete-time signal such as $x[n]$ or $y[n]$ is described by an infinite sequence of values, i.e., the time index n takes values in $-\infty$ to $+\infty$. The above picture is a snapshot at a particular time n .

In the diagram above, the sequence of output values $y[n]$ is the response of system S to the input sequence $x[n]$.

Time Invariant Systems

Let $y[.]$ be the response of S to input $x[.]$

If for all possible sequences $x[n]$ and integers D

$$x[n-D] \rightarrow S \rightarrow y[n-D]$$

then system S is said to be *time invariant* (TI). A time shift in the input sequence to S results in an identical time shift of the output sequence.

Linear Systems

Let $y_1[\cdot]$ be the response of S to input $x_1[\cdot]$, and $y_2[\cdot]$ be the response to $x_2[\cdot]$

If the response to linear combinations of these two inputs equals the **same** linear combination of the individual responses, then system S is said to be *linear*.

$$a_1x_1[n] + a_2x_2[n] \rightarrow \boxed{S} \rightarrow a_1y_1[n] + a_2y_2[n]$$

If the input is the weighted sum of several signals, the response is the corresponding **superposition** (i.e., weighted sum with **same weights**) of the response to those signals.

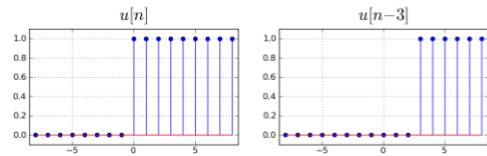
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Lecture 10, Slide #9

Unit Step

A simple but useful discrete-time signal is the *unit step* signal or function, $u[n]$, defined as

$$u[n] = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$



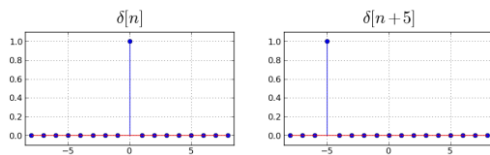
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Lecture 10, Slide #10

Unit Sample

Another simple but useful discrete-time signal is the *unit sample* signal or function, $\delta[n]$, defined as

$$\delta[n] = u[n] - u[n-1] = \begin{cases} 0, & n \neq 0 \\ 1, & n = 0 \end{cases}$$



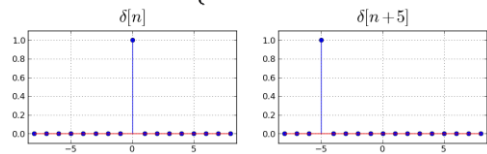
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Lecture 10, Slide #11

Unit Sample

Another simple but useful discrete-time signal is the *unit sample* signal or function, $\delta[n]$, defined as

$$\delta[n] = u[n] - u[n-1] = \begin{cases} 0, & n \neq 0 \\ 1, & n = 0 \end{cases}$$

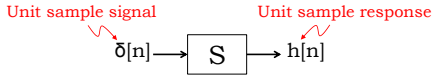


Note that standard algebraic operations on signals (e.g. subtraction, addition, scaling by a constant) are defined in the obvious way, instant by instant.

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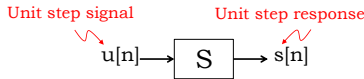
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Unit Sample Response & Unit Step Response



The *unit sample response* of a system S is the response of the system to the unit sample input. We will typically denote the unit sample response as $h[n]$.

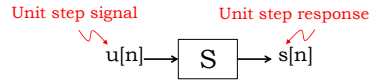
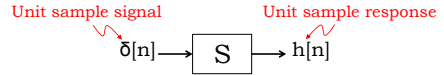
Similarly, the *unit step response* $s[n]$:



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Relating $h[n]$ and $s[n]$ of an LTI System

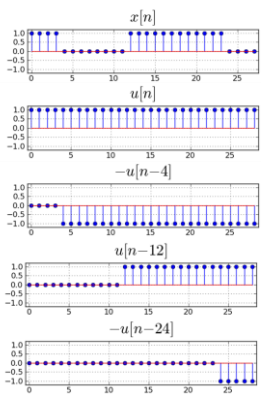


$$\delta[n] = u[n] - u[n-1] \implies h[n] = s[n] - s[n-1]$$

from which it follows that $s[n] = \sum_{k=0}^n h[k], n \geq 0$

(assuming $s[k] = 0, k < 0$, i.e., a **causal*** LTI system)

*causal system has $h[n]=0$ and $s[n]=0$ for $n < 0$.



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Unit Step Decomposition

“Rectangular-wave” digital signaling waveforms, of the sort we have been considering, are easily decomposed into **time-shifted, scaled unit steps** --- each transition corresponds to another shifted, scaled unit step.

e.g., if $x[n]$ is the transmission of 1001110 using 4 samples/bit:

$$\begin{aligned} x[n] &= u[n] \\ &\quad - u[n-4] \\ &\quad + u[n-12] \\ &\quad - u[n-24] \end{aligned}$$

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... so the corresponding response is

$$\begin{aligned} x[n] &= u[n] \\ &\quad - u[n-4] \\ &\quad + u[n-12] \\ &\quad - u[n-24] \end{aligned} \implies \begin{aligned} y[n] &= s[n] \\ &\quad - s[n-4] \\ &\quad + s[n-12] \\ &\quad - s[n-24] \end{aligned}$$

Note how we have invoked **linearity and time invariance!**

Let's apply to our averaging filter example ...

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Spot Quiz

