Massachusetts Institute of Technology

Department of Electrical Engineering and Computer Science 6.245: MULTIVARIABLE CONTROL SYSTEMS

by A. Megretski

Problem Set 8 Solutions ¹

The problem set deals with H-Infinity optimization and Hankel Optimal model reduction.

Problem 8.1T

For all values of parameter $a \in \mathbb{R}$ for which the CT H-Infinity optimization setup

$$\dot{x}(t) = u(t) + w_1(t), \quad e(t) = ax(t) + u(t), \quad y(t) = x(t) + w_2(t)$$

IS WELL-POSED, FIND THE OPTIMAL CONTROLLER AND THE MINIMAL COST.

Answer: the setup is not singular if and only if $a \neq 0$. The minimal H-Infinity norm γ_0 of the closed loop transfer function is a for a > 0, and $\sqrt{r_0}$ for $a \leq 0$, where r_0 is the (unique) real root of the polynomial

$$\phi(r) = (r-1)^2(r-a^2) - 4a^2r$$

satisfying $r_0 \ge 1$. The optimal H-Infinity controller has the form $u(t) = -\gamma_0 y(t)$, even when a = 0.

Reasoning: for the square J of the closed loop H-Infinity norm to be smaller than a given number r > 0, there must exist real n-by-n matrices H = H' > 0 and L (where n > 1 is the order of the closed loop system) such that the quadratic form

$$\sigma(x,w,v,z,u,q) = rw^2 + rv^2 - (ax+u)^2 - 2 \begin{bmatrix} x \\ q \end{bmatrix}' H \begin{bmatrix} u+w \\ z \end{bmatrix}$$

¹Version of December 25, 2011.

is positive definite on the vector space

$$V_L = \left\{ (x, w, v, z, u, q) \in \mathbb{R} \times \mathbb{R} \times \mathbb{R} \times \mathbb{R} \times \mathbb{R}^{n-1} \times \mathbb{R} \times \mathbb{R}^{n-1} : \begin{bmatrix} q \\ u \end{bmatrix} = L \begin{bmatrix} z \\ x+w \end{bmatrix} \right\}.$$

According to the generalized Parrott's lemma, such L exists for a given $H=H^\prime>0$ if and only if

$$\sigma(x, w, -x, 0, 0, 0) > 0 \quad \forall \ (x, w) \neq 0 \tag{1}$$

(the "zero information" condition), and

$$\sup_{u,q} \sigma(x, w, v, z, u, q) > 0 \quad \forall \ (x, w, v, z) \neq 0.$$
 (2)

Let P and Q denote the upper left elements of H and H^{-1} respectively. Then (1) is equivalent to

$$r(r-a^2) > P^2, (3)$$

and (2) is equivalent to

$$2aQ > \frac{1-r}{r}. (4)$$

Taking into account that positive numbers P,Q can be upper left elements of H,H^{-1} for some H=H'>0 if and only if $PQ\geq 1$, we conclude that J can be made less than r if and only if either a>0 and $r>a^2$, or $a\leq 0$, r>1, and $\phi(r)>0$. This proves that γ_0 is not smaller than the value indicated in the answer.

Given positive numbers P, Q such that PQ > 1, one matrix H = H' > 0 such that P, Q are the upper left elements of H, H^{-1} is given by

$$H = \left[\begin{array}{cc} P & Q^{-1} - P \\ Q^{-1} - P & P - Q^{-1} \end{array} \right].$$

To show that optimality level $\gamma_0 = a$ can be achieved for a > 0, consider the matrix H which corresponds to the "extremal" values of r, P, Q^{-1} :

$$r = a^2$$
, $P = 0$, $Q = +\infty$, $Q^{-1} = 0$, $H = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$.

The corresponding

$$\sigma(x, w, v, z, u, q) = a^{2}w^{2} + a^{2}v^{2} - (ax + u)^{2}$$

is made positive semidefinite by

$$u = -ay = -a(x+w),$$

which proves the optimality of the feedback u = -ay for a > 0.

To show that optimality level $\gamma_0 = \sqrt{r_0}$ can be achieved for a < 0, consider the matrix H which corresponds to the "extremal" values of r, P, Q^{-1} :

$$r = r_0, P = \sqrt{r_0(r_0 - a^2)} = \frac{-2ar_0}{r_0 - 1}, Q = \frac{r_0 - 1}{-2ar_0}, H = \begin{bmatrix} -2ar_0/(r_0 - 1) & 0\\ 0 & 0 \end{bmatrix}.$$

The corresponding

$$\sigma(x, w, v, z, u, q) = r_0 w^2 + r_0 v^2 - (ax + u)^2 + \frac{4ar_0(u + w)}{r_0 - 1}$$

is made positive semidefinite by

$$u = -\sqrt{r_0}y = -\sqrt{r_0}(x+w),$$

which proves optimality of the feedback $u = -\sqrt{r_0}y$ for a < 0.

The analytical solution can be verified by the following code:

```
function ps81(a)
s=tf('s');
p=ss(0,[1 0 1],[a;1],[0 0 1;0 1 0]);
[K1, \tilde{GAM1}] = hinfsyn(p, 1, 1);
                              % numerical solution
if a>0,
    K=-a;
    GAM=a;
else
    r=roots([1, -2-a^2, 1-2*a^2, -a^2]);
    r=max(r(r==real(r)));
    P=sqrt(r*(r-a^2));
    K=-r/(P+a);
    GAM=sqrt(r);;
end
CL=lft(p,K);
GAM2=norm(CL, Inf);
K0=squeeze(freqresp(K1,[0]));
                 analytical %f, numerical %f\n',-K,-KO)
fprintf('-K:
fprintf('GAMMA: analytical %f, numerical %f, achieved %f\n', ...
    GAM, GAM1, GAM2)
```

Problem 8.2T

For all values of parameter $a \in \mathbb{R}$ find a Hankel optimal reduced model of order 1 for

$$G(z) = \frac{1 - a^2}{z} + \frac{a}{z^2}.$$

Answer: the reduced model $\hat{G}(z)$ is given by

$$\hat{G}(z) = \begin{cases} \frac{1-a^2}{z-a}, & |a| < 1, \\ \frac{1-a^2}{z+1/a}, & |a| > 1, \\ 0, & |a| = 1. \end{cases}$$

Reasoning: in general, in order to find an Hankel norm-optimal reduced model of order d for a stable controllable and observable state space model

$$x(t+1) = Ax(t) + Bw(t), \quad y(t) = Cx(t)$$
 (5)

with $x(t) \in \mathbb{R}^n$, $w(t) \in \mathbb{R}^m$, $y(t) \in \mathbb{R}^k$, we seek (N+k)-by-(N+m) real matrix L and an (n+N)-by-(n+N) real symmetric matrix H=H' with not more than n+d positive eigenvalues, such that the quadratic form

$$\sigma(x, w, v, u, q) = r|w|^2 - |Cx - u|^2 - \left[\begin{array}{c} Ax + Bw \\ q \end{array}\right]' H \left[\begin{array}{c} Ax + Bw \\ q \end{array}\right] - \left[\begin{array}{c} x \\ v \end{array}\right]' H \left[\begin{array}{c} x \\ v \end{array}\right]$$

is positive semidefinite on the subspace defined by

$$\left[\begin{array}{c} q \\ u \end{array}\right] = L \left[\begin{array}{c} v \\ w \end{array}\right]. \tag{6}$$

Once such H, L are found, with r being as small as possible, the optimal reduced model is the *stable* part of the system with input w and output u defined by the state space equations

$$\left[\begin{array}{c} v(t+1) \\ u(t) \end{array}\right] = L \left[\begin{array}{c} v(t) \\ w(t) \end{array}\right].$$

The theory suggests using

$$H = \begin{bmatrix} W_o & -\theta \\ -\theta & \theta \end{bmatrix}, \quad \theta = W_o - rW_c^{-1}, \tag{7}$$

where W_o, W_c are the observability and controllability Gramians of (5), i.e.

$$W_o = A'W_oA + C'C, \quad W_c = AW_cA' + BB',$$

and r is chosen to control the number of positive eigenvalues of H, which in this case equals n+d, where d is the number of positive eigenvalues of θ . For this particular H, we have

$$\sigma(x, w, v, u, q) = 2x'(A'\theta q + C'u - \theta v - A'W_o Bw) + r|w|^2 - |u|^2 + 2w'B'\theta q - q'\theta q + v'\theta v, (8)$$

which implies that the best pair $(\tilde{q}, u) = (\theta q, u)$ is uniquely defined as a linear function of $(\tilde{v}, w) = (\theta v, w)$ by

$$(\tilde{q}, u) = \tilde{L}(\tilde{v}, u) = \arg \min_{\tilde{q}, u: A'\tilde{q} + C'u = \theta v + A'W_oBw} \{r|w|^2 - |u|^2 + 2w'B'\tilde{q} - \tilde{q}'\theta^+\tilde{q} + \tilde{v}'\theta^+\tilde{v}\},$$
(9)

where θ^+ is a pseudo-inverse of θ , i.e. a matrix such that $\theta = \theta \theta^+ \theta$. This in turn determines the reduced system \hat{G} as the stable part of the state space model

$$(\tilde{v}(t+1), u(t)) = \tilde{L}(\tilde{v}(t), w(t)),$$

where $\tilde{v}(t)$ belongs to the range of θ for all t.

In this problem we have n = 2, m = k = d = 1, and a state space model for G is given by

$$A = \left[\begin{array}{cc} 0 & 0 \\ 1 & 0 \end{array} \right], \quad B = \left[\begin{array}{c} 1 \\ 0 \end{array} \right], \quad C = \left[\begin{array}{cc} 1 - a^2 & a \end{array} \right].$$

The corresponding Gramians are

$$W_o = \begin{bmatrix} 1 - a^2 + a^4 & a(1 - a^2) \\ a(1 - a^2) & a^2 \end{bmatrix}, W_c = I, A'W_oB = \begin{bmatrix} a(1 - a^2) \\ 0 \end{bmatrix}.$$

The Hankel singular numbers of system G are square roots of the eigenvalues of $W_o = W_o W_c$, i.e. 1 and a^2 . When |a| > 1, the largest possible value of r for which matrix H in (7) has a single positive eigenvalue is r = 1, which corresponds to

$$\theta = (a^2 - 1) \begin{bmatrix} -a \\ 1 \end{bmatrix} \begin{bmatrix} -a \\ 1 \end{bmatrix}', \quad A'\theta = (a^2 - 1) \begin{bmatrix} 1 \\ 0 \end{bmatrix} \begin{bmatrix} -a \\ 1 \end{bmatrix}'.$$

Representing the elements \tilde{v} in the range of θ by

$$\tilde{v} = \frac{1}{1+a^2} \begin{bmatrix} -a \\ 1 \end{bmatrix} \tilde{x}, \quad \tilde{x} \in \mathbb{R}$$

in

$$A'\tilde{v}(t+1) + C'u(t) = \theta\tilde{v}(t+1) + A'W_oBw(t)$$
(10)

yields

$$\tilde{x}(t+1) = -\frac{1}{a}\tilde{x}(t) - aw(t), \quad u(t) = \frac{a^2 - 1}{a}\tilde{x}(t),$$

which means that the optimal reduced model has transfer function

$$\hat{G}(z) = \frac{1 - a^2}{z + 1/a}.$$

When |a| < 1, the largest possible value of r for which matrix H in (7) has a single positive eigenvalue is $r = a^2$, which corresponds to

$$\theta = (1 - a^2) \begin{bmatrix} 1 \\ a \end{bmatrix} \begin{bmatrix} 1 \\ a \end{bmatrix}', \quad A'\theta = (1 - a^2) \begin{bmatrix} a \\ 0 \end{bmatrix} \begin{bmatrix} 1 \\ a \end{bmatrix}'.$$

Representing the elements \tilde{v} in the range of θ by

$$\tilde{v} = \frac{1}{1+a^2} \begin{bmatrix} 1 \\ a \end{bmatrix} \tilde{x}, \quad \tilde{x} \in \mathbb{R}$$

in (10) yields

$$\tilde{x}(t+1) = a\tilde{x}(t) + w(t), \quad u(t) = (1-a^2)\tilde{x}(t),$$

which means that the optimal reduced model has transfer function

$$\hat{G}(z) = \frac{1 - a^2}{z - a}.$$

The analytical solution can be checked by MATLAB code

```
function ps82(a)
z=tf('z');
G=(1-a^2+a/z)/z;
sig=hsvd(G);
if a>1,
        Gr=(1-a^2)/(z+1/a);
        GAM=1;
elseif a<1,
        Gr=(1-a^2)/(z-a);
        GAM=a^2;
else
        Gr=0;</pre>
```

GAM=1;

end

Gr1=tf(hankelmr(G,1))

er=max(hsvd(Gr-Gr1));

fprintf('GAMMA: analytical %f, numerical %f\n',GAM,sig(2))

fprintf('MISMATCH: %e\n',er)

fprintf('HinfErr: analytical %f, numerical %f\n',norm(G-Gr,Inf),norm(G-Gr1,Inf))

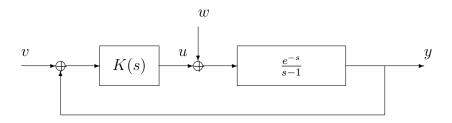


Figure 1: Problem 8.3 setup

Problem 8.3P

Consider the task of finding the minimal control effort required to stabilize the CT system with transfer function

$$P(s) = \frac{e^{-s}}{s-1}$$

USING A FIXED ORDER CONTROLLER K=K(s), AS SHOWN ON FIGURE 1. MORE SPECIFICALLY, YOU ARE ASKED TO USE HANKEL OPTIMAL MODEL REDUCTION AND H-INFINITY OPTIMIZATION TO DEVELOP AN ALGORITH FOR FINDING CONTROLLERS K=K(s) OF A GIVEN ORDER WHICH STABILIZE THE SYSTEM FROM FIGURE 1 (SO THAT THE CLOSED LOOP L2 GAIN FROM [v;w] TO [u,y] FINITE), AND MAKE THE CLOSED LOOP L2 GAIN FROM w TO u AS SMALL AS POSSIBLE.

You are not expected to look for the true optimal solution (there are no efficient methods for doing this), but rather to explore ways in which a good low order approximation of an H_{∞} class transfer function containing e^{-s} , such as

$$G(s) = \frac{e^{-s}}{s+1}$$
, or $G(s) = \frac{e^{-s} - e^{-1}}{s-1}$,

CAN BE USED TO SET UP SOME H-INFINITY OPTIMIZATION TO AID IN DESIGNING A GOOD CONTROLLER.

PRODUCE A TABLE OF VALUES OF BEST $w \to u$ L2 Gains achieved by your algorithm with controllers of order m for $m \in \{1, 2, 4, 8, 16\}$ (it's OK if some of the numbers in the table equal ∞).

Conclusion: for $m \in \{1, 2, 4, 8, 16\}$, robust stabilization is possible for $m \in \{4, 8, 16\}$, with L2 gain bounds of (respectively) 11.7, 5.0, 4.2. This does not mean that better L2 gains are not achievable.

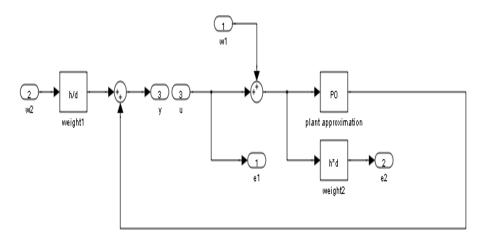


Figure 2: Open loop model for Problem 8.3

Code: uses design block diagram ps83des.mdl shown on Figure 2. The essential part of ps83.m is shown below:

```
function ps83(g,m,n,dd)
% function ps83(g,m,n,dd)
%
% g - desired closed loop L2 gain bound
% m - desired controller order
% n - order of the initial approximation for (exp(-s)-exp(-1))/(s-1)
% dd - vector of scaling factors to use
if nargin<1, g=100; end
if nargin<2, m=2; end
if nargin<4, dd=logspace(-1,1,10)'; end</pre>
```

```
dd=dd(:);
nd=length(dd);
s=tf('s');
[A,B,C,D]=ssdata(ss((1-s*(0.5/n))/(1+s*(0.5/n)))^n);
sys=pck(A,B,C/(A-eye(n)),0);
[sys,sig]=sysbal(sys);
[Ar,Br,Cr,Dr]=unpck(hankmr(sys,sig,m-1,'d'));
LO=ss(Ar,Br,Cr,Dr);
                       % approximation for (\exp(-s)-\exp(-1))/(s-1)
L0=hankelmr(ss(A,B,C/(A-eye(n)),0),m);
P0=exp(-1)/(s-1)+L0;
ww=logspace(-2,2,10000);
sss=1i*ww(:);
mm=abs(squeeze(freqresp(L0,sss))-(exp(-sss)-exp(-1))./(sss-1));
r=max(mm);
%close(gcf);semilogx(ww,mm,[ww(1) ww(end)],[r r]);grid;pause
fprintf('approximation error: %f\n',r)
h=sqrt(g*r);
assignin('base','P0',P0)
assignin('base','h',h)
gg=zeros(nd,1);
for i=1:length(dd),
    d=dd(i);
    assignin('base','d',d)
                                    % export variables
    p=linmod('ps83des');
                                    % extract open loop
    p=ss(p.a,p.b,p.c,p.d);
    [K, \tilde{gAM}] = hinfsyn(p, 1, 1);
                                % optimize controller
    gg(i)=GAM;
    fprintf('.')
end
Af=ssdata(K);
fprintf('order: %d\n',size(Af,1))
close(gcf); plot(dd,gg/g); grid
return
                                % export controller
assignin('base','K',K)
p=linmod('ps54test');
                                % extract closed loop
S=ss(p.a,p.b,p.c,p.d);
```

```
\label{eq:final_stability} $$frintf('[T=\%f] stability: \%f<0, small gain: \%f<\%f\n', \dots \\ T,max(real(eig(p.a))),norm(S,Inf),1/r) \\ close(gcf);plot(ww,mm,[ww(1),ww(end)],[r r]);grid
```