MMSE estimation and lattice encoding/decoding for linear Gaussian channels

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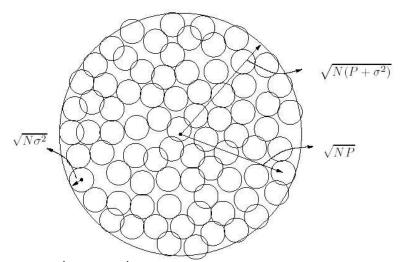
$$Y = X + N$$

where $N \sim \mathcal{N}\left(0, \sigma_N^2\right), \frac{1}{n} \sum_{i=1}^n X_i^2 \leq P_X.$
• Shannon: capacity is
1 P_Y

$$C = \frac{1}{2} \log_2 \left(1 + SNR\right), \ SNR = \frac{P_X}{\sigma_N^2}$$

• Random coding argument: generate 2^{nC} i.i.d. $\mathcal{N}(0, P_X)$ codewords. Averaging across all codebooks: under ML decoding $P(e) \rightarrow 0$ as $n \rightarrow \infty$.

Geometrically Achieving Capacity



- LLN: $\mathcal{N}(0, \sigma^2)$ i.i.d. *n*-vector lies in sphere of radius $\sqrt{n\sigma^2}$.
- $X \sim \mathcal{N}(0, P_X), N \sim \mathcal{N}(0, \sigma_N^2) \Rightarrow Y \sim \mathcal{N}(0, P_X + \sigma_N^2). \Rightarrow Y$ lies in sphere of radius $\sqrt{n(P_X + \sigma_N^2)}.$ • Codewords chosen as centers of
- Codewords chosen as centers of non-overlapping spheres w/ radius $\sqrt{n\sigma_N^2}$
- Volume of n *n*-sphere w/ radius *r* is $A_n r^{\hat{n}}$.
- ⇒ max no. of non-overlapping decoding spheres:

$$\frac{A_n \left[n(P_X + \sigma_N^2) \right]^{\frac{n}{2}}}{A_n \left[n\sigma_N^2 \right]^{\frac{n}{2}}} = 2^{\frac{n}{2}\log_2 \left(1 + \frac{P_X}{\sigma_N^2} \right)} = 2^{nC}$$

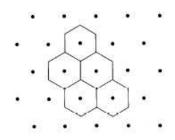
Structured Coding for AWGN channels

- Researchers for decades interested in structured codes, encoding mechanisms, and decoding mechanisms
- Desire: achieve capacity on the AWGN channel for *arbitrary* SNRs.
- Devote our attention to lattices: algebraic in nature.
- Basis of today's talk: How Uri Erez and Ram Zamir solved the decades-old problem of achieving the AWGN channel capacity at all SNRs, using lattice codes and lattice decoding.
- Surprisingly and non-so-intuitive at first glance:
 - using a *biased* MMSE estimator at the decoder is essential to achieve capacity.
 - related to deep connection between mutual information and MMSE estimation (Baris's talk in a couple weeks).

Lattices

 Lattice: a discrete group which is a subset of Rⁿ. Described in terms of a generator matrix:

 $\Lambda = \{\lambda = Gx : x \in \mathbb{Z}^n\}, \ G \in \mathbb{R}^{n \times n}$



• Fundamental Voronoi region of Λ:

 $\mathcal{V} = \{ x \in \mathbb{R}^n | \| x - \mathbf{0} \| \le \| x - \lambda \| \ \forall \ \lambda \in \Lambda \} \,.$

• Any $x \in \mathbb{R}^n$ uniquely expressed as

$$x = \lambda + r, \text{ where } \lambda \in \Lambda, r \in \mathcal{V}$$
$$= Q_{\mathcal{V}}(x) + x \mod_{\mathcal{V}} \Lambda.$$

 $\ensuremath{\mathcal{V}}$ analogous to remainder in modular arithmetic.

• Generally: any fundamental region Ω satisfies $x \in \mathbb{R}^n$ uniquely expressed as

$$x = \lambda + r, \text{ where } \lambda \in \Lambda, r \in \Omega$$
$$= Q_{\Omega}(x) + x \mod_{\Omega} \Lambda.$$

Desired Properties of Good Lattices

- Denote volume of any $\mathcal{R} \subset \mathbb{R}^n$ as $V(\mathcal{R})$.
- 2nd moment per dim. of \mathcal{R} :

$$P(\mathcal{R}) = \frac{1}{n} \frac{\int_{\mathcal{R}} ||x||^2 dx}{V(\mathcal{R})}$$

- Avg energy per dim. of $U \sim unif(\mathcal{R})$.

• Normalized 2nd moment of \mathcal{R} :

$$G(\mathcal{R}) = \frac{P(\mathcal{R})}{V(\mathcal{R})^{\frac{2}{n}}}$$

- S_{n,σ^2} : the *n*-sphere with radius $\sqrt{n\sigma^2}$.
 - a) $V(S_{n,\sigma^2})^{\frac{2}{n}} \to 2\pi e \sigma^2, \ P(S_{n,\sigma^2}) \to \sigma^2,$ $\Rightarrow G(S_{n,\sigma^2}) \to \frac{1}{2\pi e}$ b) $\sigma_N^2 < \sigma^2 \Rightarrow P\left([X \sim \mathcal{N}\left(0, \sigma_N^2\right)] \notin S_{n,\sigma^2}\right) \to 0.$
- Λ_S 'good for shaping' if a):

$$G(\mathcal{V}_S)
ightarrow rac{1}{2\pi e}.$$

• Λ_C 'good for channel coding' if b):

$$\sigma_N^2 < \frac{V(\mathcal{V}_C)^{\frac{2}{n}}}{2\pi e} \Rightarrow P\left(\left[X \sim \mathcal{N}\left(0, \sigma_N^2\right)\right] \notin \mathcal{V}_C\right) \to 0.$$

Lattice Codes

• Lattice code C:

 $C = \Lambda_C \cap \mathcal{S}.$

Shaping region S imposes signaling constraint (such as power constraint for AWGN channel).

• Lattice decoder C: simply a quantizer $Q_{\Omega_C}(x)$ for Λ_C . Performs the operation

 $\lambda = Q_{\Omega_C}(y) \in \Lambda_C.$

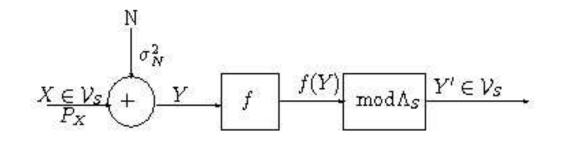
Note the decoder does not take into account the shaping region S associated with the lattice code, which simplifies the decoding process.

Previous Work on Lattice Codes

- De Buda considered a spherical lattice code where S is a sphere and is Λ_C 'good for channel coding'
- Numerous authors: S should be a thin spherical shell. Under ML decoding, the capacity is achieved.
 - But ML decoding requires finding the lattice point closest to the received signal *inside the shell*.
 - Decoding regions lose structure, have no relation to true lattice decoding.
- A spherical lattice code with a Euclidean minimum-distance decoder can achieve ¹/₂ log₂(SNR).
 - At high SNR, this essentially achieves capacity.
 - At low SNR, significant performance loss. We will discuss why 1 is missing here later.

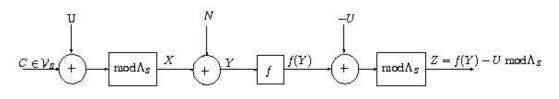
Mod-Lattice Transmission and Lattice Decoding

- Now temporarily step away from 'good for channel coding' codes Λ_C and consider Λ_S that is 'good for shaping'.
- **Desire**: \mathcal{V}_S will serve as \mathcal{S} and allow more structured encoding/decoding.



If Λ_S is 'good for shaping' $(G(\mathcal{V}_S) \to \frac{1}{2\pi e})$, $X \sim unif(\mathcal{V}_S)$, and f an MMSE estimator of X, then $\frac{1}{2}\log_2(1 + SNR)$ is achievable.

Mod-Lattice Transmission and Lattice Decoding (Cont'd)



- Introduce dither $U \sim unif(\mathcal{V}_S)$, known to both the encoder and decoder.
- Given any $C \in \mathcal{V}_S$, the channel input is $X = C + U \mod \Lambda_S$.
- $\Rightarrow X \sim \operatorname{unif}(\mathcal{V}_S) \text{ and } X \perp C.$ Why: $P_U(u)$ constant $\forall u \in \mathcal{V}_S.$ As $x \nearrow \mathcal{V}_S$, $x - c \mod \Lambda_S \nearrow \mathcal{V}_S.$ $\Rightarrow P_{X|C}(x|c) = P_U(x - c \mod \Lambda_S),$ constant $\forall x \in \mathcal{V}_S, c \in \mathcal{V}_S.$

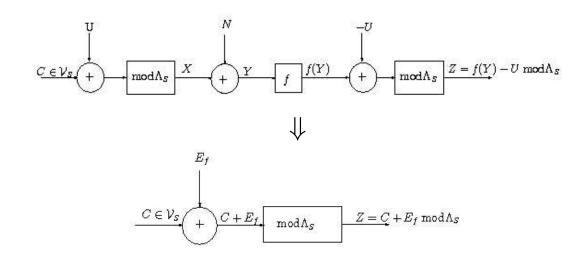
Mod-Lattice Transmission and Lattice Decoding (Cont'd)

- Dither contributes 2 nice things:
- $X \sim \text{unif}(\mathcal{V}_S)$, \Rightarrow power constraint met with equality.

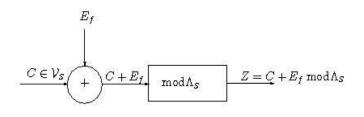
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$$X \perp C$$
; $C \leftrightarrow X \leftrightarrow Y \Rightarrow (Y, X) \perp C$.

•
$$\Rightarrow E_f = f(Y) - X \perp C.$$

• $Z = C + E_f \mod \Lambda_S$. \Rightarrow now an additive noise channel:



Equivalent Channel Model

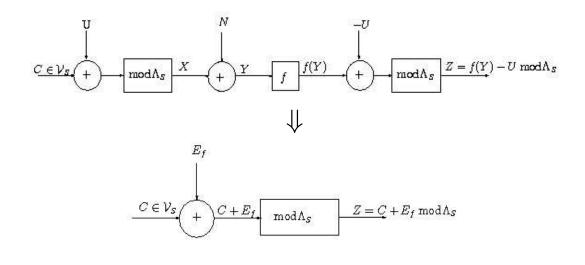


$$\begin{array}{ll} C &\sim & \mathrm{unif}\left(\mathcal{V}_{S}\right) \; \mathrm{optimal} \; \Rightarrow Z \sim \mathrm{unif}\left(\mathcal{V}_{S}\right).\\ E_{f}' \; \triangleq \; E_{f} \; \mathrm{mod}\Lambda_{S} \; .\\ \mathbf{C} \; \geq \; \mathbf{C}(\Lambda_{S},f) = \frac{1}{N} \left[h(Z) - h(Z|C)\right]\\ &= \; \frac{1}{N} \left[\log_{2} V(\Lambda_{S}) - h(E_{f}')\right]\\ &= \; \frac{1}{2} \log_{2} 2\pi e P_{X} - \frac{1}{2} \log_{2} 2\pi e G(\mathcal{V}_{S}) - \frac{1}{N} h(E_{f}')\\ &\geq \; \frac{1}{2} \log_{2} 2\pi e P_{X} - \frac{1}{2} \log_{2} 2\pi e G(\mathcal{V}_{S}) - \frac{1}{N} h(E_{f}) \end{aligned}$$

$$\quad \text{EPI:} \; \frac{1}{N} h(E_{f}) \leq \log_{2} 2\pi e P_{E_{f}}.\\ &\Rightarrow \mathbf{C}(\Lambda_{S},f) \; \geq \; \frac{1}{2} \log_{2} \frac{P_{X}}{P_{E_{f}}} - \frac{1}{2} \log_{2} 2\pi e G(\mathcal{V}_{S}).\\ &\bullet \; \Lambda_{S} \; \text{'good for shaping':} \; G(\mathcal{V}_{S}) \rightarrow \frac{1}{2\pi e}. \end{array}$$

$$\Rightarrow \mathbf{C} \ge \mathbf{C}(\Lambda_S, f) \ge \frac{1}{2} \log_2 \frac{P_X}{P_{E_f}}.$$

MMSE Estimation



$$\mathbf{C} \ge \mathbf{C}(\Lambda_S, f) \ge \frac{1}{2}\log_2 \frac{P_X}{P_{E_f}}$$

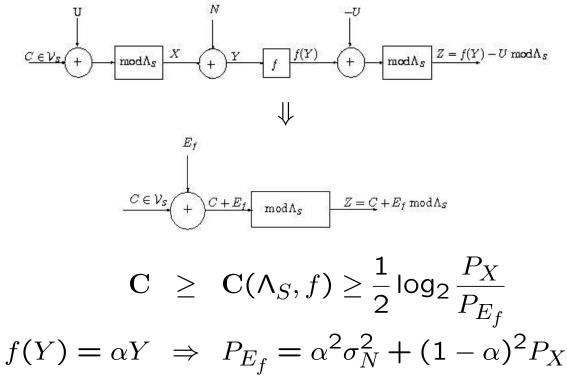
• Let
$$f(Y) = \hat{X}(Y) = \alpha Y$$
:

$$E_f = \alpha Y - X = \alpha N - (1 - \alpha)X$$

$$\Rightarrow P_{E_f} = \alpha^2 \sigma_N^2 + (1 - \alpha)^2 P_X$$

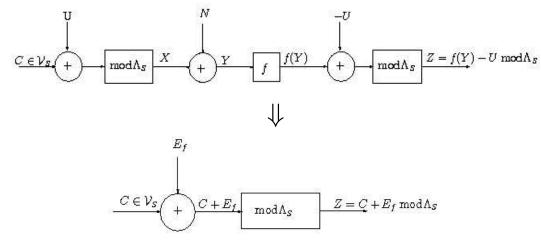
• minimize $P_{E_f} \Leftrightarrow$ choose α^* to be linear MMSE estimate:

$$\alpha^* = \frac{P_X}{P_X + \sigma_N^2} = \frac{SNR}{1 + SNR}$$
$$\Rightarrow P_{E_f}^* = \frac{P_X \sigma_N^2}{P_X + \sigma_N^2}$$
$$\Rightarrow C(\Lambda_S, f^*) = \frac{1}{2} \log_2(1 + SNR).$$

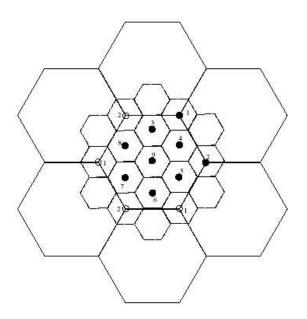


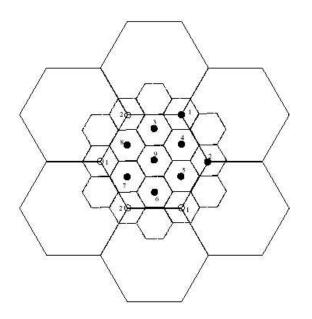
- Dither U used in non-symmetric way:
 - At encoder, simply added to codeword, followed by ${\rm mod}\Lambda_S$
- At decoder, Y is scaled followed by dither subtraction and $mod\Lambda_S$ operation
- Prev. ways of using $\text{mod}\Lambda_S$: no scaling $\Leftrightarrow \alpha = 1 \Rightarrow C(\Lambda_S, f) = \frac{1}{2}\log_2(SNR)$
- $\alpha^* \neq 1$: estimator is *biased*.
- MMSE scaling minimizes $var(E_f)$ and increases 'effective' SNR by factor $\frac{SNR+1}{SNR}$.

Nested Lattice Codes



- Desire: use structured coding scheme to signal $C \in \mathcal{V}_S$. Consider lattice codes.
- Fine Λ_C : 'good for channel coding'.
- Shape with \mathcal{V}_S , Λ_S 'good for shaping'.
- Nested lattice code: $\Lambda_S \subset \Lambda_C$.





$$\mathcal{C} = \{\Lambda_C \mod \Lambda_S\} = \{\Lambda_C \cap \mathcal{V}_S\}$$
$$R = \frac{1}{n}\log_2|\mathcal{C}| = \frac{1}{n}\log_2\frac{V(\mathcal{V}_S)}{V(\mathcal{V}_C)}$$

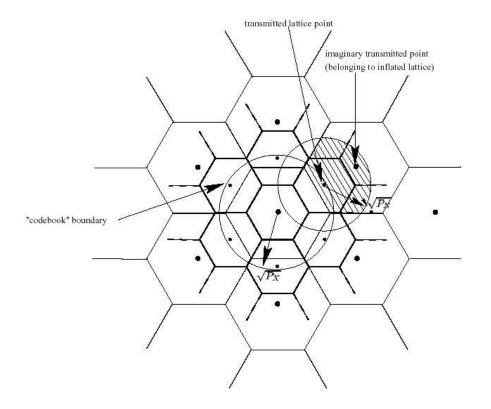
- Erez, Zamir show that nested lattice codes with desired properties exist for all SNRs.
- ML decoding with nested lattices is equivalent to lattice decoding.
 - ML decoder's quantizer:

 $\Omega_C^* = \{ e : f_{E_f}(e) \ge f_{E_f}(e - c \mod \Lambda_S) \forall c \in \mathcal{C} \}$

- Note that $\Omega_C^* \neq \mathcal{V}_S$. Using \mathcal{V}_S instead suffices and can achieve capacity.

Discussion

• Inflated lattice



- Geometry
 - Force αY to lie in same sphere as X: $\tilde{\alpha} = \sqrt{\frac{SNR}{SNR+1}}$. \Rightarrow not the right intuition
- But since $\alpha^* = \frac{SNR}{SNR+1} < \tilde{\alpha}$, with high prob. from LLN, no information loss in $\alpha^* Y \to \alpha^* Y \mod \Lambda_S$ transformation.

Other Coding problems with Gaussian Distributions

Costa's 'Dirty paper coding'

Y = S + X + N

 ${\cal S}$ known to encoder, not to decoder.

- Constructively and trivially addressed with Erez/Zamir technique: add α^*S to channel input
- Wyner-Ziv: rate-distortion bound achieved with these codes.
- Error exponents: α^* only optimal as $R \rightarrow \mathbf{C}$.
 - Lower rates: α^* suboptimal.
 - Random coding error exponent can be achieved at all rates with proper choice of α .
- MIMO flat fading channels: generalization of these codes achieves diversity-multiplexing tradeoff.