## Spectra of Large Block Matrices

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#### Outline

- 1. Introduction: An example
- 2. Problem statement
- 3. Theorem
- 4. Sketch of the proof
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# An example in wireless communication:

# Multiple Input Multiple Output (MIMO) system with ISI

In a MIMO system:

 $n_T$  transmitter antenna

 $n_R$  receiver antenna

$$Y_n = [y_{1,n}, \cdots, y_{n_R,n}]^T$$
$$= HX_n + N_n$$

received signal 
$$(y_{i,n} = \sum_{j=1}^{n_T} y_{i,j,n})$$

$$H=\left(h_{i,j}\right)_{i,j}$$

channel matrix

$$X_n = [x_{1,n}, \cdots, x_{n_T,n}]^T$$

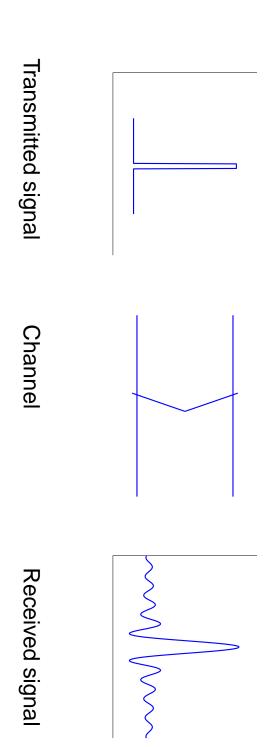
transmitted signal

 $N_n$ 

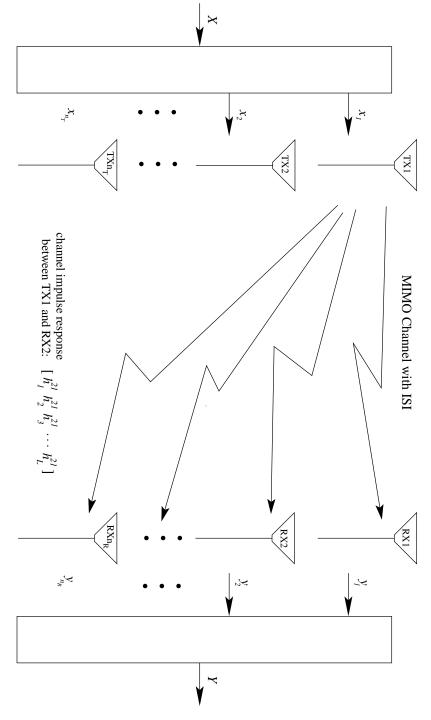
noise signal

and received at antenna i in the receiver  $h_{i,j}$  the channel effect on the signal transmitted from antenna j in the transmitter

### what happens in practice?



### A more realistic model



Block diagram of a MIMO system with ISI

Considering Intersymbol-Interference (ISI):

$$y_{i,j,n} = \sum_{l=1}^{L} x_{j,n-l} h_l^{(ij)} + n_{i,j,n}$$
 $h_{ij} = \begin{bmatrix} h_1^{(ij)} & h_2^{(ij)} & \dots & h_{L-1}^{(ij)} & h_L^{(ij)} \end{bmatrix}^T$ 

For a signal frame of K:

 $A_l = (h_l^{(ij)})_{\substack{i=1,\cdots,n_R\\j=1,\cdots,n_T}}$ 

#### But ...

the spectra of  $HH^{st}$  is the key factor in:

- Capacity of the channel,
- Structure of the receiver,

and the **trick** is to use:

$$H^*$$
 0

which has all the eigenvalues of  $HH^{st}$  plus some zeros!

### **Problem statement:**

Selfadjoint block matrix:

$$X_N = \sum_{i,j=1}^u E_{ij} \otimes A^{(i,j)}$$
  $d$  blocks of  $N imes N$ matrices in each row

where,

$$E_{ij}: d\times d \text{ Elementary matrix; entry } (i,j)=1 \text{ the rest zero }$$
 
$$A^{(i,j)}=\left(a_{rp}^{(i,j)}\right)_{r,p=1,\ldots,N}: N\times N \text{ Gaussian random matrix }$$
 
$$A^{(i,j)}=A^{(j,i)*}$$
 
$$E\left(a_{rp}^{(i,j)}\right)=0$$
 
$$E[a_{rp}^{(i,j)}\overline{a_{sq}^{(l,k)}}]=\frac{1}{n}\delta_{rs}\delta_{pq}\cdot\sigma(i,j;k,l) \quad (n=Nd)$$

#### **Notations**

ullet  $\sigma=\left(\sigma\left(i,j;k,l
ight)
ight)_{i,j,k,l=1}^{d},$  covariance function

ullet  $M_{d}\left( \mathbb{C}
ight) :d imes d$  matrices with complex entries

 $ullet \eta: M_d(\mathbb C) o M_d(\mathbb C)$  covariance mapping 1

 $[\eta(D)]_{ij} := \frac{1}{d} \sum_{k,l=1}^d \sigma(i,k;l,j) d_{kl} \quad \text{for } D = (d_{ij})_{i,j=1}^d \in M_d(\mathbb{C})$ 

 $ullet \operatorname{tr}_d\Bigl(D\Bigr) := rac{1}{d} \sum_{i=1}^d [D]_{ii} \quad ext{normalized trace}$ 

#### **Theorem**

Cauchy transform  $G(\boldsymbol{z})$  is determined by For  $N o \infty$ , the n imes n matrix  $X_N$  has a limiting eigenvalue distribution whose

$$G(z) = \operatorname{tr}_d(\mathcal{G}(z)),$$

by the facts that where  $\mathcal{G}\left(z
ight)$  is an  $M_d\left(\mathbb{C}
ight)$ -valued analytic function on  $\mathbb{C}^+$  uniquely determined

1. 
$$\lim_{|z|\to\infty,\Im(z)>0} z\mathcal{G}(z) = I_d$$

2. 
$$z\mathcal{G}\left(z\right) = I_d + \eta(\mathcal{G}\left(z\right)) \cdot \mathcal{G}\left(z\right) \quad \forall z \in \mathbb{C}^+$$

### Sketch of proof-1

Wick formula: For Gaussian family of  $x_1, x_2, \cdots, x_n$ :

$$E\left[x_{i(1)}\cdots x_{i(k)}\right] = \sum_{\pi \in \mathcal{P}_2(k)} \prod_{(r,s) \in \pi} E\left[x_{i(r)}x_{i(s)}\right],$$

where  $\mathcal{P}_{2}\left(k
ight)$  is the set of all pairings of the set  $\{1,\cdots,k\}$ .

$$\Rightarrow \lim_{N \to \infty} E[\operatorname{tr}_n(X_N^m)] = \sum_{\pi \in NC_2(m)} \mathcal{K}_{\pi},$$

where

$$\mathcal{K}_{\pi} := rac{1}{d^{m/2+1}} \sum_{i(1),...,i(m)=1}^{a} \prod_{(p,q)\in\pi} \sigma\Big(i(p),i(p+1);i(q),i(q+1)\Big)$$

At the operator-valued level:  $\mathcal{K}_{\pi}:=\mathrm{tr}_{d}(\kappa_{\pi}), \kappa_{\pi}\in M_{d}\left(\mathbb{C}
ight)$  $\Rightarrow \lim_{N\to\infty} E[\operatorname{tr}_n(X_N^m)] = \operatorname{tr}_d \Big\{ \sum_{\pi\in NC_2(m)} \kappa_\pi \Big\}$ 

### Non-crossing partions!

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$$\eta \left( \eta \left( \eta(I_d) \right) \right)$$

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 $\eta \Big( \eta(I_d) \cdot \eta(I_d) \Big)$ 

### Sketch of proof-2

- $\mathcal{E}(X^m) := \sum_{\pi \in NC_2(m)} \kappa_{\pi}$  $\Rightarrow \lim_{N \to \infty} E[\operatorname{tr}_n(X_N^m)] = \operatorname{tr}_d(\mathcal{E}(X^m)),$ X is called "operator-valued semicircular element".
- ullet  $\mathcal{M}(z) = \sum_{m=0}^{\infty} \mathcal{E}[X^m]z^m,$  generating power series

$$\Rightarrow \mathcal{M}(z) = I_d + z^2 \eta \Big( \mathcal{M}(z) \Big) \cdot \mathcal{M}(z)$$

ullet  $\mathcal{G}(z):=rac{1}{z}\mathcal{M}(1/z)$  operator-valued Cauchy transform

$$\Rightarrow z\mathcal{G}(z) = I_d + \eta(\mathcal{G}(z)) \cdot \mathcal{G}(z)$$

# **Example: Block Toeplitz matrices**

$$X = \frac{1}{\sqrt{3N}} \begin{bmatrix} A & B & C \\ B & A & B \\ C & B & A \end{bmatrix}, \quad X = \frac{1}{\sqrt{4N}} \begin{bmatrix} A & B & C & D \\ B & A & B & C \\ C & B & A & B \\ D & C & B & A \end{bmatrix}$$

A,B,C,D,E selfadjoint Gaussian random matrices

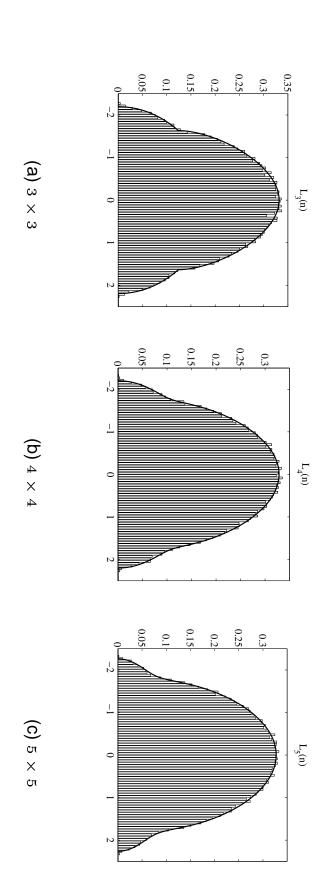
Step I: 
$$\mathcal{G} = egin{bmatrix} f & 0 & h \\ 0 & g & 0 \\ h & 0 & f \end{bmatrix}$$

Step II : 
$$\eta(\mathcal{G})=\frac{1}{3}$$
 
$$0 \qquad 2f+g+0 \qquad 0 \qquad g+2h$$
 
$$0 \qquad 2f+g+2h \qquad 0 \qquad 2f+g$$

Step III: Solve the following system of equations:

$$z\mathcal{G}(z) = I_d + \eta(\mathcal{G}(z)) \cdot \mathcal{G}(z) \Rightarrow \begin{cases} zf = 1 + \frac{g(f+h)+2(f^2+h^2)}{3} \\ zg = 1 + \frac{g(g+2(f+h))}{3} \\ zh = \frac{4fh+g(f+h)}{3} \end{cases}$$

Step IV: Apply Stieltjes inversion formula



Block Toeplitz matrices spectra( $N=100~{
m for}~100~{
m realizations}$ )

as the number of blocks gets larger, the spectra goes to semicircular!

# MIMO channel with ISI (K=4 and L=4)

Wishart type block matrices, like  $H^{st}H$  with

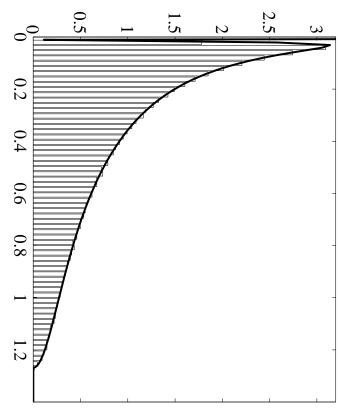
$$H = \begin{bmatrix} A & B & C & D & 0 & 0 & 0 \\ 0 & A & B & C & D & 0 & 0 & 0 \\ 0 & 0 & A & B & C & D & 0 \\ 0 & 0 & 0 & A & B & C & D & 0 \end{bmatrix}$$

A,B,C,D non-seldadjoint Gaussian random matrices.

we use :

$$\begin{bmatrix} 0 & H \\ H^* & 0 \end{bmatrix}$$

but we cancel the additional zeros!



 $H_nH_n^{st}/n$  for K=4 and L=4 over its histogram for N=100, based on 100Superimposed theoretical density of the eigenvalues of complex normal realizations.