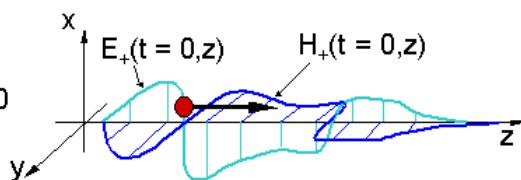


## WIRELESS COMMUNICATIONS

### Nature of waves: (review)

$$\nabla \times \bar{E} = -\mu_0 \frac{\partial \bar{H}}{\partial t} \quad \nabla \cdot \bar{E} = \rho / \epsilon_0 = 0$$

$$\nabla \times \bar{H} = \epsilon_0 \frac{\partial \bar{E}}{\partial t} \quad \nabla \cdot \bar{H} = 0$$



$$\text{Yields Wave Equation: } \nabla^2 \bar{E} - \mu_0 \epsilon_0 \frac{\partial^2 \bar{E}}{\partial t^2} = 0$$

### Many (infinite number) solutions to Wave Equation:

e.g.  $E_x(z,t) = E_+(z - ct) + E_-(z + ct)$  where  $c = 1/\sqrt{\mu_0 \epsilon_0}$   
 = forward + backward traveling wave  
 = arbitrary function  $E_{+/-}(z, t)$

e.g.  $E_x(z,t) = E_0 \cos(\omega t - kz)$  [arg = constant] marks spot ● on waveform  
 If  $\omega t = kz$ , then  $d/dt (\omega t - kz)$  yields  $dz/dt = \omega/k = c$   
 $H_y(z,t) = (E_0/\eta_0) \cos(\omega t - kz)$

### In general:

$$\bar{E}(\vec{r},t) = \sum_i \bar{E}_i(\vec{r},t) = \text{sum of uniform plane waves in various directions } i$$

L3-1

## NATURE OF WAVES--POLARIZATION

Most communications is narrowband, and therefore the signals are roughly sinusoidal.

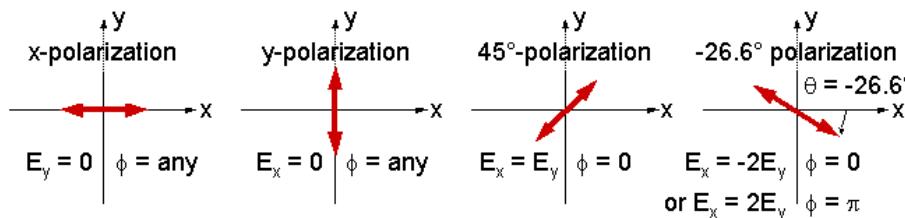
Z-moving waves can generally have electric field components in both x and y directions.

$$\text{e.g. Let } \bar{E}(\vec{r},t) = \hat{x} E_x \cos(\omega t - kz) + \hat{y} E_y \cos(\omega t - kz - \phi)$$

Plots of  $\bar{E}(z=0,t)$

Cases:

#### Linear polarization



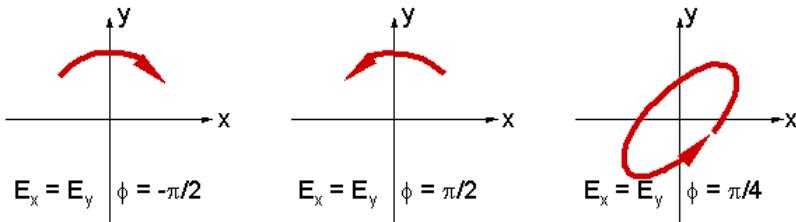
L3-2

### Main points of L3: Polarized waves and phasors

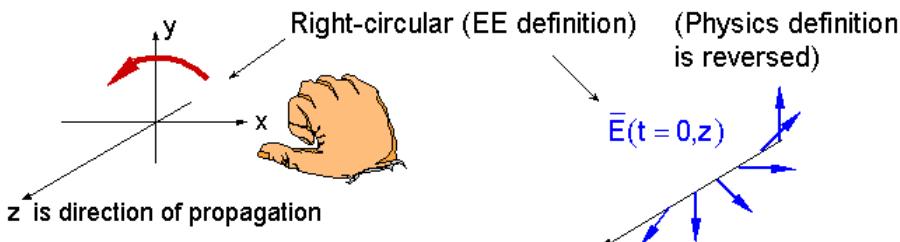
- e.g. +z wave:  $\bar{E}(\vec{r},t) = \hat{x} E_x \cos(\omega t - kz) + \hat{y} E_y \cos(\omega t - kz - \phi)$ ,  $\phi$ ,  $E_x/E_y$  determine pol.
- If  $\phi = 0$ , then any linear polarization can be chosen using  $E_x$ ,  $E_y$
- If  $(\phi = \pm\pi/2)$  and  $(E_x = E_y) \Rightarrow$  circular polarization; else  $\Rightarrow$  elliptical polarization
- Convention:  $\bar{E}(\vec{r},t) = R_e \{ \bar{E}(\vec{r},t) e^{j\omega t} \}$
- Example:  $\bar{E}(\vec{r}) = 5(\hat{x} + j\hat{y}) e^{-jkz} \Rightarrow \bar{E}(\vec{r},t) = 5[\hat{x} \cos(\omega t - kz) - \hat{y} \sin(\omega t - kz)]$
- Antennas are polarized and respond to  $\bar{E}$  or  $\bar{H}$ , or to both

## CIRCULAR AND ELLIPTICAL POLARIZATION

Let  $\bar{E}(r,t) = \hat{x}E_x \cos(\omega t - kz) + \hat{y}E_y \cos(\omega t - kz - \phi)$



Handedness:



L3-3

## EXAMPLES OF POLARIZED WAVES

Right Circular Polarization at  $t = 0$

$\bar{E} \times \bar{H}$  is in direction of power flow  
( $+z$  here)

Field pattern translates as a rigid body in  $+z$  direction

y-polarization at  $t = 0$

L3-4

## POLARIZED WAVES--REPRESENTATION

### Time Domain Example:

$$\bar{E}(r,t) = \hat{x}E_x \cos(\omega t - kz - \phi_x) + \hat{y}E_y \cos(\omega t - kz - \phi_y)$$

where  $k = \omega k = 2\pi/\lambda$

### Frequency Domain (Phasors):

$\underline{E}e^{-jkz}$  represents the same wave as above! ( $\bar{E} = \hat{x}\underline{E}_x + \hat{y}\underline{E}_y$ )

i.e.  $\bar{E}(r,t) = \operatorname{Re} \{ \underline{E}e^{-jkz} e^{j\omega t} \} = \operatorname{Re} \{ (\hat{x}\underline{E}_x + \hat{y}\underline{E}_y) e^{j(\omega t - kz)} \}$

where  $\operatorname{Re} \{ \hat{x}\underline{E}_x e^{j(\omega t - kz)} \} = \operatorname{Re} \{ \hat{x}\underline{E}_x e^{-j\phi_x} e^{j(\omega t - kz)} \}$

$$= \hat{x}\underline{E}_x \operatorname{Re} \{ e^{j(\omega t - kz - \phi_x)} \} = \hat{x}\underline{E}_x \cos(\omega t - kz - \phi_x)$$

[We defined  $E_x = |\underline{E}_x|$ ]

L3-5

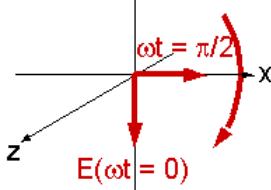
## EXAMPLES OF PHASOR REPRESENTATION

### Using $\bar{E}(r,t) = \operatorname{Re} \{ \underline{E}(r)e^{j\omega t} \}$ :

- 1)  $\bar{E}(r) = 5\hat{y}e^{-jkz}$  implies  $\bar{E}(r,t) = 5\hat{y}\cos(\omega t - kz)$  [y-polarization]
- 2)  $\bar{E}(r) = 5(\hat{x} + \hat{y})e^{-jkz}$  implies  $\bar{E}(r,t) = 5(\hat{x} + \hat{y})\cos(\omega t - kz)$  [45°-polarization]
- 3)  $\bar{E}(r) = 5(\hat{x} + j\hat{y})e^{-jkz}$  implies  $\bar{E}(r,t) = 5[\hat{x}\cos(\omega t - kz) - \hat{y}\sin(\omega t - kz)]$   
 [Recall  $e^{j\omega t} = \cos \omega t + j \sin \omega t$ ] [left-circular-polarization]

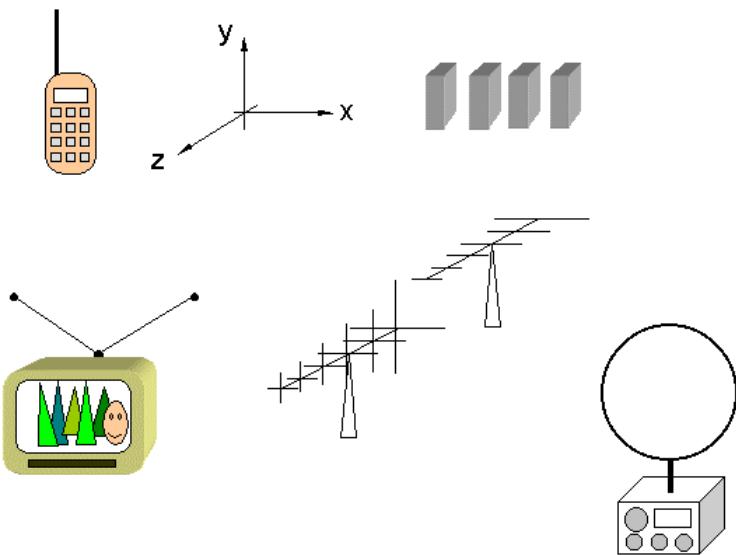
### Why left-circular?

Let  $z = 0$ .  $\operatorname{Re} \{ (\hat{x} + j\hat{y})e^{j\omega t} \} = \hat{x}$  at  $t = 0$   
 $= \operatorname{Re} \{ (\hat{x} + j\hat{y})j \} = -\hat{y}$  at  $\omega t = \pi/2$



L3-6

## POLARIZED TRANSMITTERS AND RECEIVERS



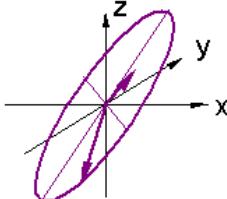
L3-7

## POLARIZATION AT A POINT IN SPACE

**Assume many waves superimpose at a single point:**

All waves are at  $\omega$

Arrive from many directions with arbitrary polarizations



**Wave #1 alone at that point produces:**

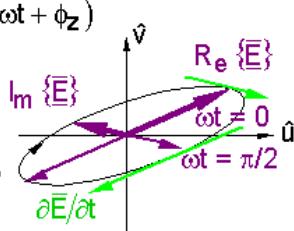
$$\bar{E}_1(t) = \hat{x}E_{x1}\cos(\omega t + \phi_{x1}) + \hat{y}E_{y1}\cos(\omega t + \phi_{y1}) + \hat{z}E_{z1}\cos(\omega t + \phi_{z1})$$

**All waves together yield  $\bar{E}(t) = \sum_i \bar{E}_i(t)$ :**

$$\bar{E}(t) = \hat{x}E_x\cos(\omega t + \phi_x) + \hat{y}E_y\cos(\omega t + \phi_y) + \hat{z}E_z\cos(\omega t + \phi_z)$$

$$\bar{E} = \hat{x}E_x + \hat{y}E_y + \hat{z}E_z = R_e \{\bar{E}\} + jI_m \{\bar{E}\}$$

The real and imaginary parts of  $\bar{E}$  define a plane  $\hat{u}, \hat{v}$   
Since there is no single direction of propagation here,  
We cannot say "right" or "left" elliptical



L3-8