MASSACHUSETTS INSTITUTE OF TECHNOLOGY Department of Electrical Engineering and Computer Science

6.014 Electrodynamics

Problem Set 11 Solutions

Available in Tutorials:

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Problem 11.1

$$E(x, y, z) = \hat{y} 2 j E_0 \sin(k_x x) e^{-az}$$

$$\therefore \overline{E}(x, y, z, t) = -\hat{y} 2 E_0 \sin(k_x x) e^{-az} \sin(wt)$$

$$\overline{H}(x, y, z) = \frac{\nabla \times \overline{E}}{-jwm} = \frac{j}{wm} [\hat{x} a \sin(k_x x) + \hat{z} k_x \cos(k_x x)] \cdot 2 j E_0 e^{-az}$$

$$\therefore \overline{H}(x, y, z, t) = -\frac{2 E_0}{wm} [\hat{x} a \sin(k_x x) + \hat{z} k_x \cos(k_x x)] e^{-az} \cos(wt)$$

$$d = \lambda/2$$

$$\overline{E}(wt = \mathbf{p}/2)$$

$$\overline{E}(\mathbf{w}t=0) = \overline{H}(\mathbf{w}t=\mathbf{p}/2) = 0$$

Problem 11.2

a) TE₁₀ mode: m = 1; n = 0 from which it follows that
$$k_x = \frac{p}{a}$$
; $k_y = 0$. In section
7.4 the general solutions of the magnetic and electric fields for TE and TM modes
are given. We thus have:
 $E_x = 0$ (from equation 7.4.13); $E_z = 0$ (because we are dealing with a transverse
electric field);
 $E_y(x, y, z) = -\frac{k_x}{k_0} E_0 \sin(k_x x) e^{-jk_z z} \rightarrow E_y(x, y, z, t) = -\frac{k_x}{k_0} E_0 \sin(k_x x) \cos(wt - k_z z)$

$$H_{x}(x, y, z) = \frac{E_{0}k_{x}k_{z}}{\mathbf{h}k_{0}^{2}}\sin(k_{x}x)e^{-jk_{z}z} \to H_{x}(x, y, z, t) = \frac{E_{0}k_{x}k_{z}}{\mathbf{h}k_{0}^{2}}\sin(k_{x}x)\cos(\mathbf{w}t - k_{z}z)$$

(from equation 7.4.15); $H_y = 0$ (from equation 7.4.15);

$$H_{z}(x, y, z) = -j \frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}} \cos(k_{x}x)e^{-jk_{z}z} \rightarrow H_{z}(x, y, z, t) = \frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}} \cos(k_{x}x)\sin(\mathbf{w}t - k_{z}z)$$

(from equation 7.4.15). The surface current density on the guide walls is related to the magnetic field intensity by, $\overline{J}_s = \hat{n} \times \overline{H}$, where \hat{n} is the outward normal from the wall surface and \overline{H} is the magnetic field intensity at the wall. At t=0 and evaluating the surface current densities at the walls of the waveguide we find:

@
$$x = 0$$
: $H_z(x = 0, y, z, t = 0) = -\frac{E_0 k_x^2}{h k_0^2} \sin(k_z z)$ and

 $H_x(x=0, y, z, t=0) = 0$ from which it follows

$$\begin{split} \bar{J}_{s}(x=0) &= -\hat{x} \times \hat{z}H_{z}(x=0, y, z, t=0) = -\hat{y}\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}\sin(k_{z}z) \,. \\ \\ @ x = a : H_{z}(x = a, y, z, t=0) = -\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}(-1)\sin(k_{z}z) \text{ and} \\ H_{x}(x = a, y, z, t=0) = 0 \text{ from which it follows} \\ \\ \bar{J}_{s}(x = a) &= \hat{x} \times \hat{z}H_{z}(x = a, y, z, t=0) = -\hat{y}\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}\sin(k_{z}z) = \bar{J}_{s}(x=0) \,. \\ \\ @ y = 0 : H_{z}(x, y = 0, z, t=0) = -\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}\cos(k_{x}x)\sin(k_{z}z) \text{ and} \\ \\ H_{x}(x, y = 0, z, t=0) &= \frac{E_{0}k_{x}k_{z}}{\mathbf{h}k_{0}^{2}}\sin(k_{x}x)\cos(k_{z}z) \text{ from which it follows} \\ \\ \bar{J}_{s}(y = 0) &= -\hat{y} \times (\hat{x}H_{x}(x, y = 0, z, t=0) + \hat{z}H_{z}(x = 0, y, z, t=0)) \\ \\ &= \hat{x}\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}\cos(k_{x}x)\sin(k_{z}z) + \hat{z}\frac{E_{0}k_{x}k_{z}}{\mathbf{h}k_{0}^{2}}\sin(k_{x}x)\cos(k_{z}z) \\ \\ @ y = b : H_{z}(x, y = b, z, t=0) &= -\frac{E_{0}k_{x}^{2}}{\mathbf{h}k_{0}^{2}}\cos(k_{x}x)\sin(k_{z}z) \text{ and} \\ \\ \\ H_{x}(x, y = b, z, t=0) &= \frac{E_{0}k_{x}k_{z}}{\mathbf{h}k_{0}^{2}}\sin(k_{x}x)\cos(k_{z}z) \text{ from which it follows} \\ \\ \bar{J}_{s}(y = b) &= \hat{y} \times (\hat{x}H_{x}(x, y = 0, z, t=0) + \hat{z}H_{z}(x = 0, y, z, t=0)) \\ \\ \end{array}$$

$$= -\hat{x}\frac{E_0k_x^2}{hk_0^2}\cos(k_xx)\sin(k_zz) - \hat{z}\frac{E_0k_xk_z}{hk_0^2}\sin(k_xx)\cos(k_zz) = -\bar{J}_s(y=0)$$



Surface currents on guide walls for TE_{10} mode in rectangular waveguide.

(Figure is adapted from "Field and Wave Electromagnetics", D.K. Cheng, p. 554, 2nd edition)

- b) In order to break the current minimally thin slots can be placed:
 - On the in the middle of the top or bottom surfaces (e.g. when y = 0, b we

need the slot to be placed at $x = \frac{a}{2}$). Note we are not specifying the length of the slot

• Vertical orientated thin slots on the side walls will also work (e.g. when y = 0, a).

Problem 11.3





Problem 11.4

a) That the optical pulse can propagate on this line before distorting does not depend on the constants $\boldsymbol{b}_0, \boldsymbol{b}_1$. From the dispersion relationship given $(k \cong \boldsymbol{b}_0 \boldsymbol{w}_0 + \boldsymbol{b}_1 (\boldsymbol{w} - \boldsymbol{w}_0) + \boldsymbol{b}_2 (\boldsymbol{w} - \boldsymbol{w}_0)^2/2)$ we can calculate the group and the phase velocity as follows: $\frac{dk}{d\omega}\Big|_{\omega_0} = \beta_1 = 1/v_g$ and $k/\omega_0 = \beta_0 = 1/v_p$ respectively. The

dispersion relationship only pertains to the β_2 constant, and thus the distance that optical pulses can propagate on this line before distorting does not depend on β_0, β_1 .

b) Assume the velocity of the wave is the speed of light. Thus the width of the pulse

train period is $\frac{c}{f} = 30cm$. The pulse train consists of a "1" and "0" that span half this width each. Thus half the width of a pulse is 0.075m. The group velocity as a function of frequency is given as follows:

$$v_g(\omega) = \left(\frac{dk}{d\omega}\right)^{-1} \bigg|_{\omega \neq \omega_0} = \frac{1}{\beta_1 + \beta_2(\omega - \omega_0)} = \frac{1}{v_g^{-1} + \beta_2(\omega - \omega_0)}$$
. We are told that

distortion will happen when the highest velocity signals have moved ahead of the slowest signals by a distance of one-half the width of a single pulse, thus $|D_{wH} - D_{wL}|$

$$=T\left|\frac{1}{v_{g}^{-1}+\boldsymbol{b}_{2}2\boldsymbol{p}(3\times10^{14}+10^{10}-3\times10^{14})}-\frac{1}{v_{g}^{-1}+\boldsymbol{b}_{2}2\boldsymbol{p}(3\times10^{14}-10^{10}-3\times10^{14})}\right|$$

= 0.075*m*

From the above we make T the subject. The distance traveled then is

$$D = cT = \frac{3 \times 10^8 (0.075)}{\left|\frac{1}{v_g^{-1} + \boldsymbol{b}_2 2\boldsymbol{p}(10^{10})} - \frac{1}{v_g^{-1} + \boldsymbol{b}_2 2\boldsymbol{p}(-10^{10})}\right|}$$