

# INTRODUCTION TO EECS II

## DIGITAL

## COMMUNICATION

## SYSTEMS

## 6.02 Fall 2010

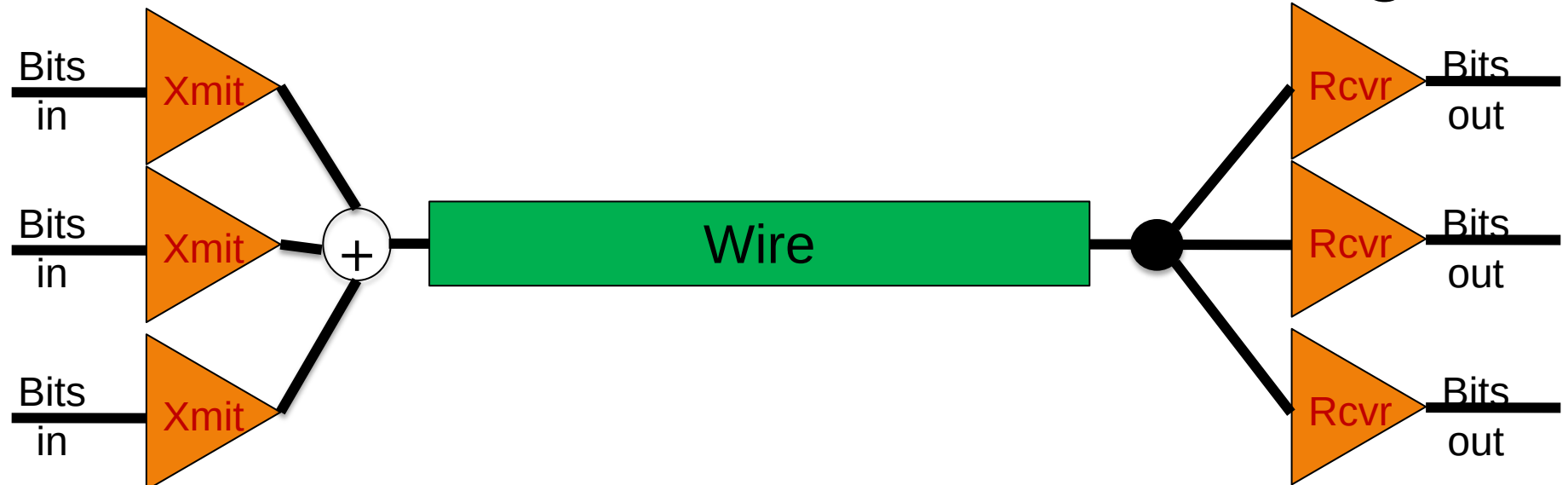
## Lecture #12

- Summary of progress
- The problem of sharing in frequency domain
- Sines  $\rightarrow$  LTI  $\rightarrow$  Sines by experiment
- Discrete-Time Complex Exponentials
- Frequency Response

# First Problem - Low BER on Wires

- Problems and Analysis Techniques:
  - Intersymbol Interference (ISI)
    - LTI Systems and Unit Sample Responses
    - Eye Diagrams
  - Noise
    - Probability Density and Cumulative Distribution Functions
    - Normal (Gaussian) random variables
- Solution Approaches
  - Techniques based on LTI models
    - Deconvolution
  - Error Detection and Correction Codes
    - Parity bits, Convolutional Code, Viterbi Algorithm

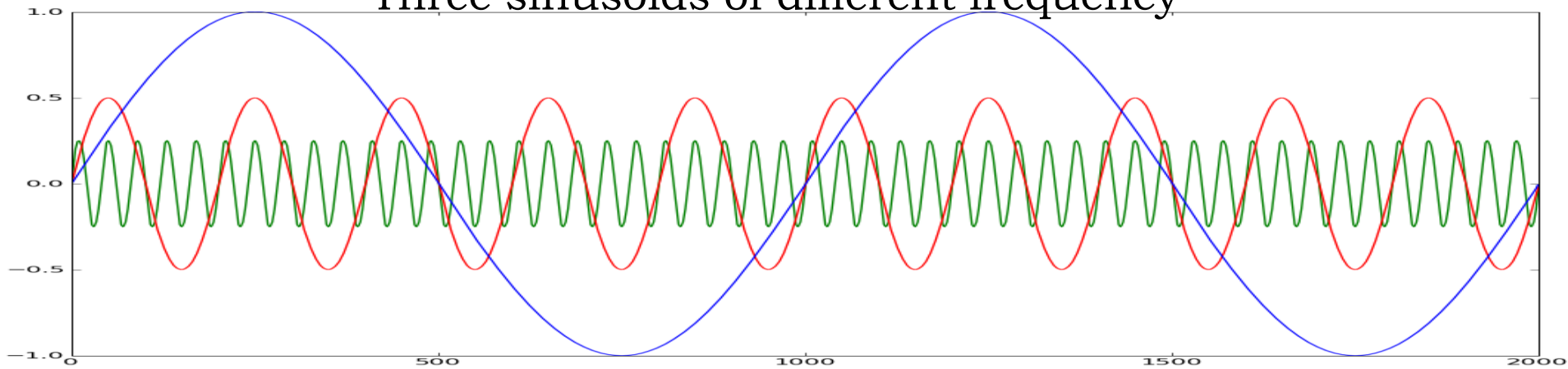
# Second Problem - Resource Sharing



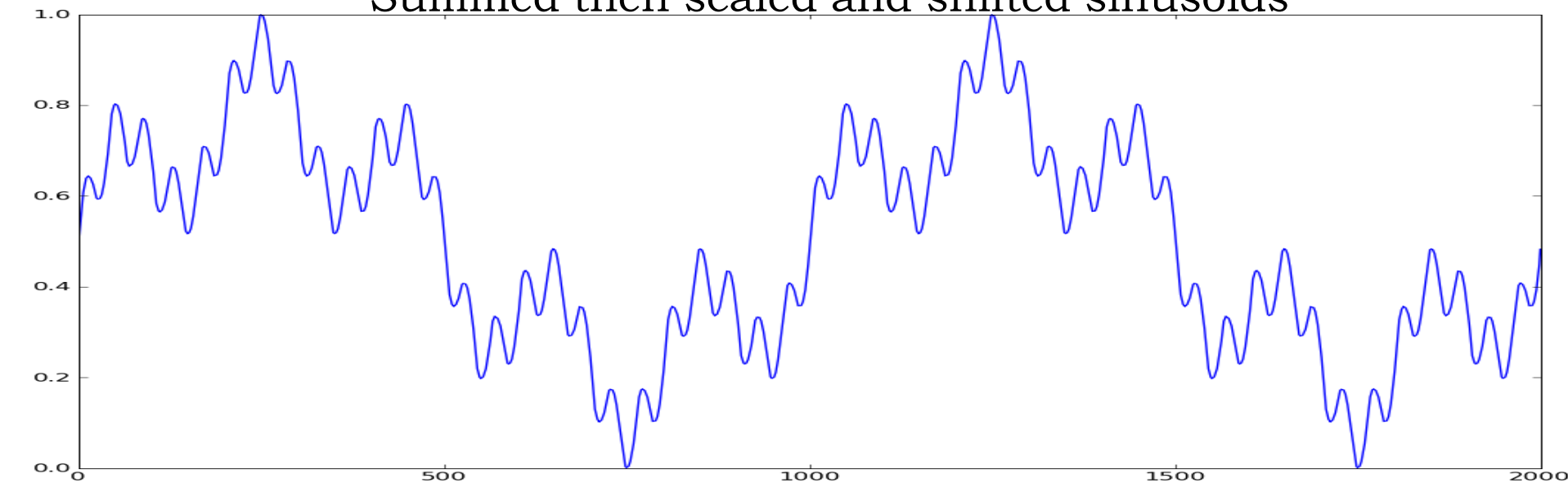
- Two Approaches
  - Time Sharing (e.g. TDMA)
    - Xmit-Rcvr pair gets a dedicated time slot or contend for slots
    - Used in both wired and wireless networks
  - Frequency Sharing (aka Frequency Division Multiplexing)
    - Each Xmit-Rcvr gets part of the spectrum (explained later)
    - Used in Broadcast TV, cell phones and WiFi.
  - Most WiFi systems use TDMA and FDM
- Next Two Weeks on issues in FDM

# Sinusoids and LTI Systems

Three sinusoids of different frequency

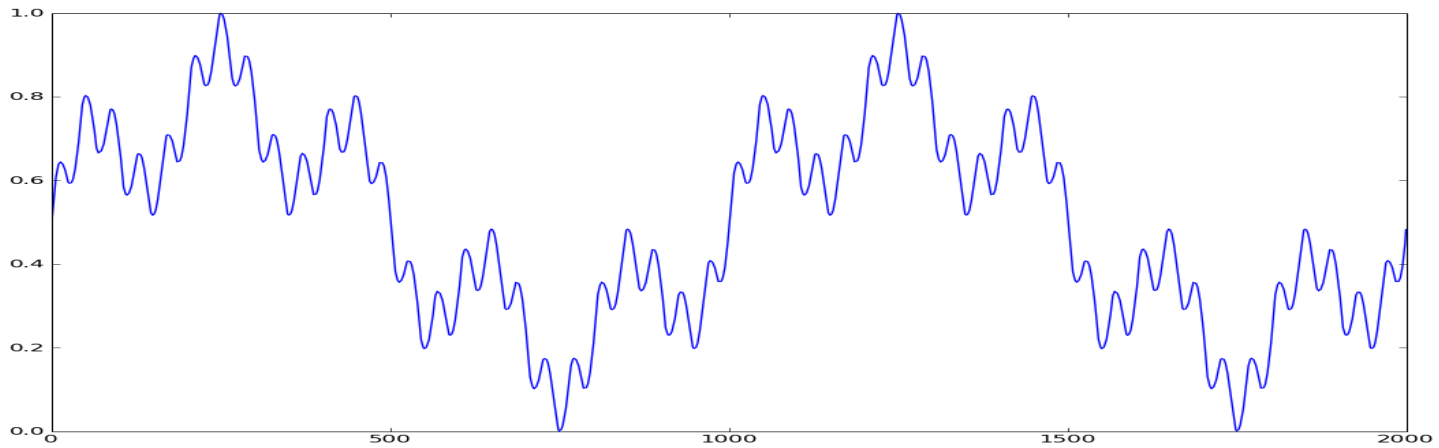


Summed then scaled and shifted sinusoids

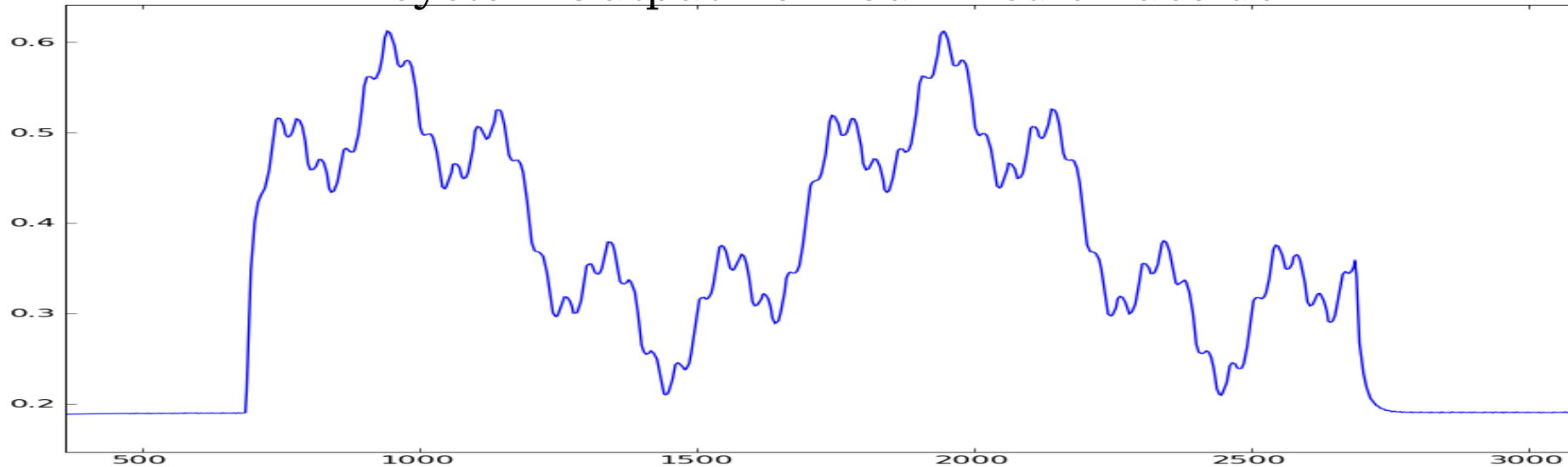


# Sinusoids and LTI Systems

Summed then scaled and shifted sinusoids

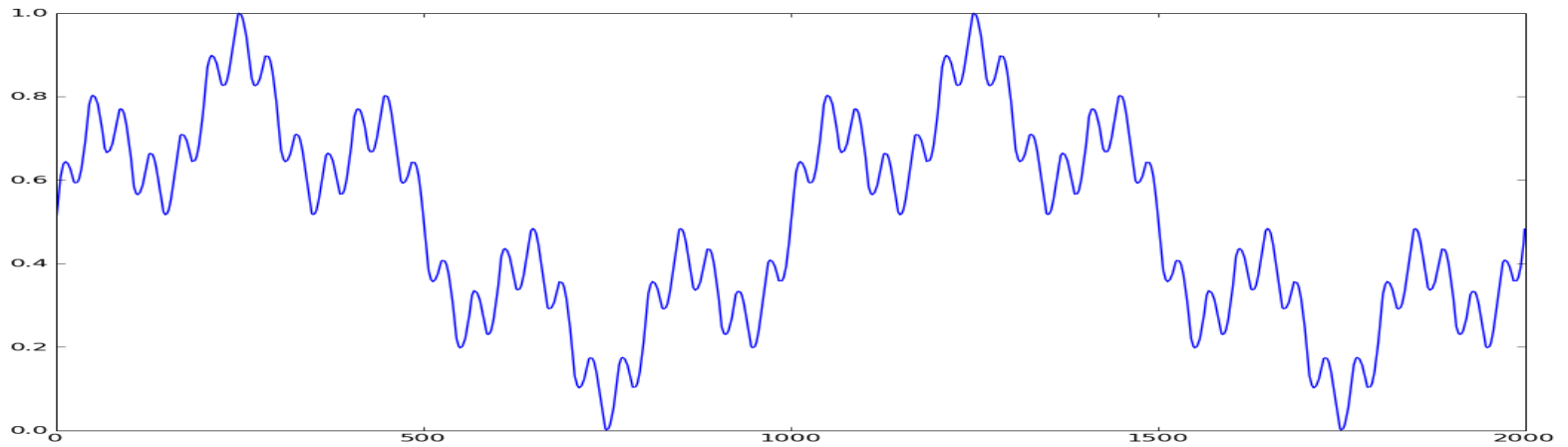


IR System Output from Summed sinusoids

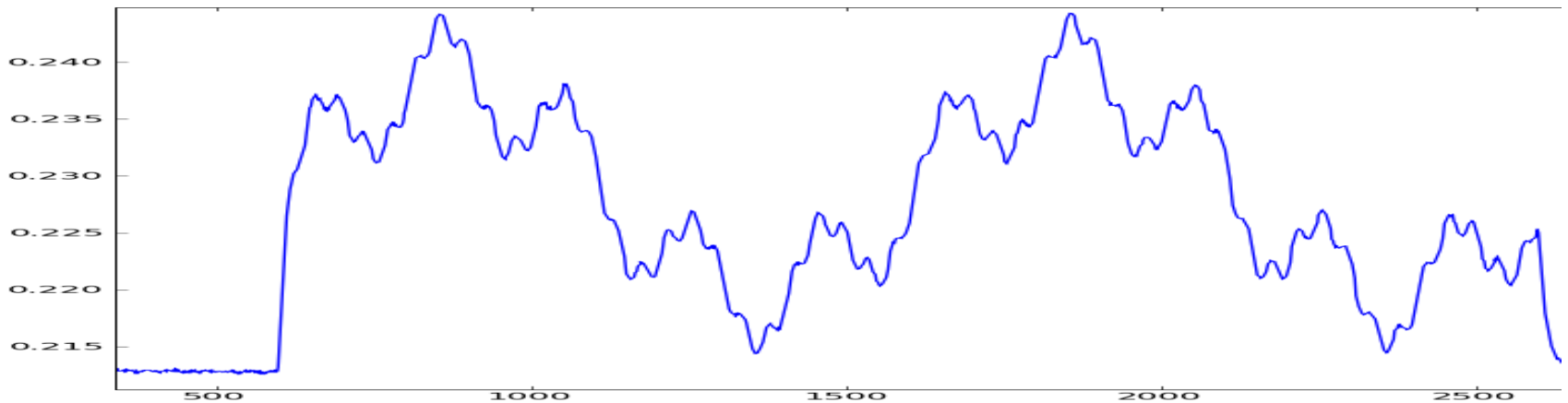


# Sinusoids and LTI Systems

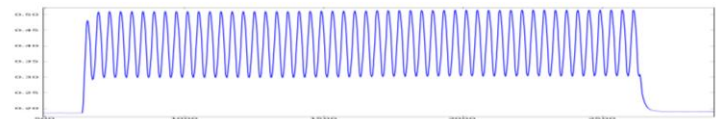
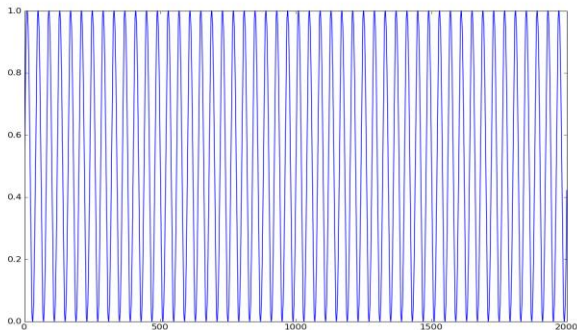
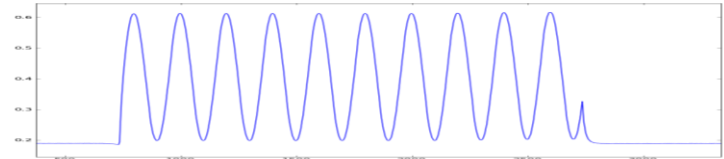
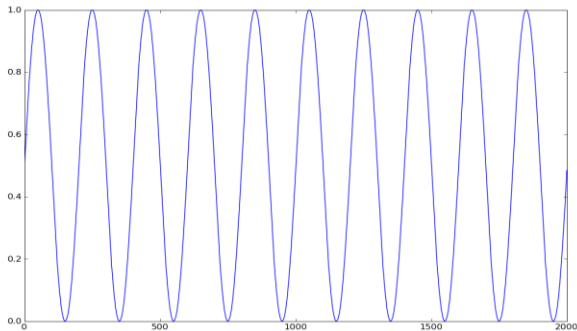
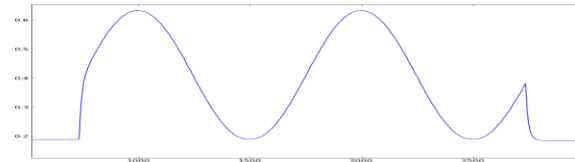
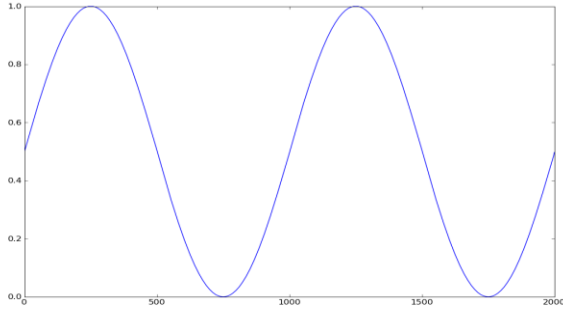
Summed then scaled and shifted sinusoids



IR System (off ceiling) Output from Summed sinusoids



# IR System Response to Different Sines



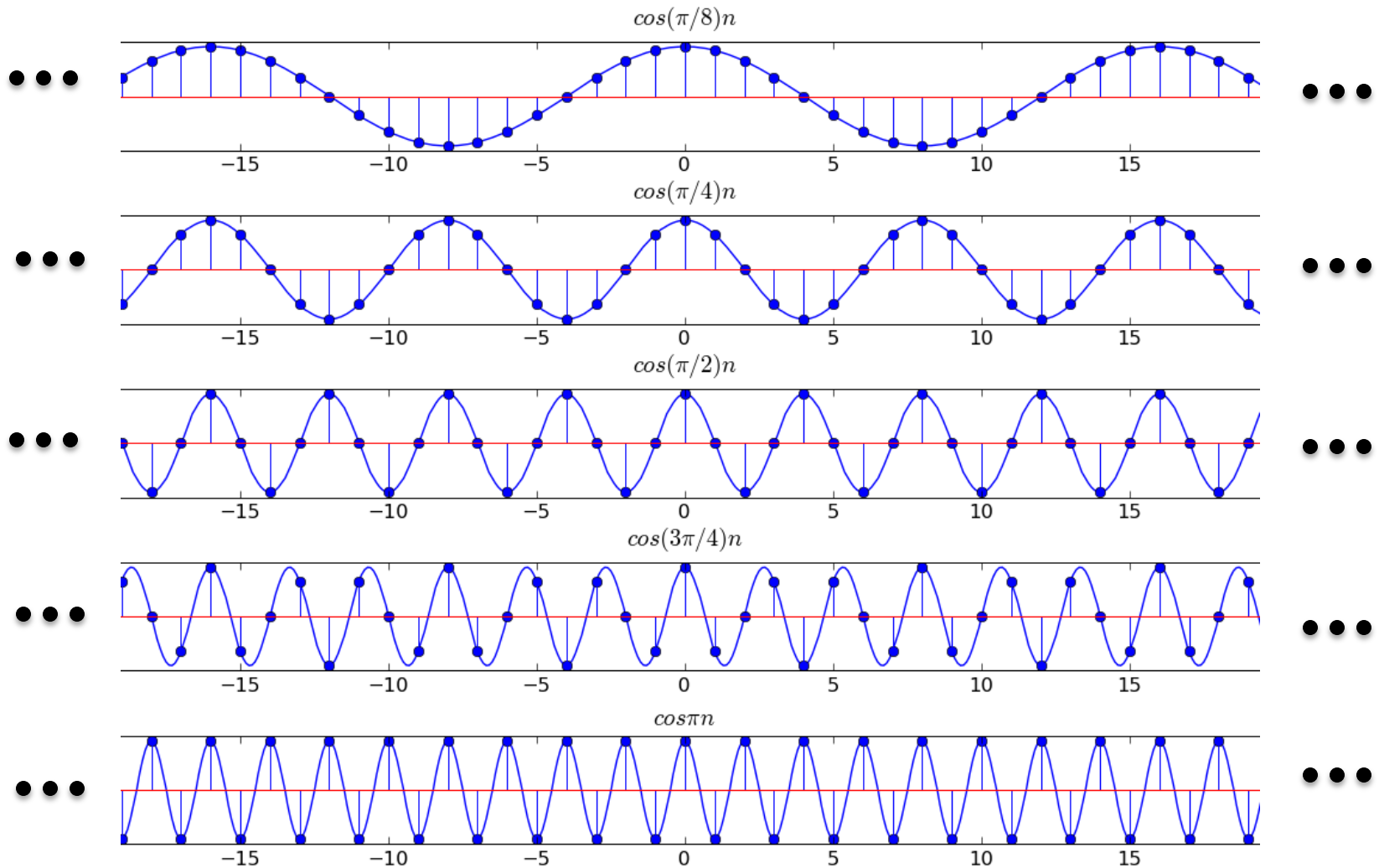
# Sinusoids in LTI Systems

- Sinusoids  $\rightarrow$  LTI Systems  $\rightarrow$  Sinusoids
  - Different Frequencies do not seem to mix
    - Xmit: encode data on different frequencies (Modulation)
    - Rcvr: decode data from different frequencies (Filtering)
  - Different Frequencies have different “Gain”
    - E.G. the IR Channel:
      - higher frequencies  $\rightarrow$  lower “gain”.
- Analyzing Sinusoids in LTI systems
  - Model as Eternal signals ( $-\infty < n < \infty$ )
  - Difficulty Computing response to  $x[n] = \cos \Omega n$
  - Simplify response computation using complex exponentials

$$\cos \Omega n = \frac{e^{j\Omega n} + e^{-j\Omega n}}{2} \quad \sin \Omega n = \frac{e^{j\Omega n} - e^{-j\Omega n}}{2j}$$



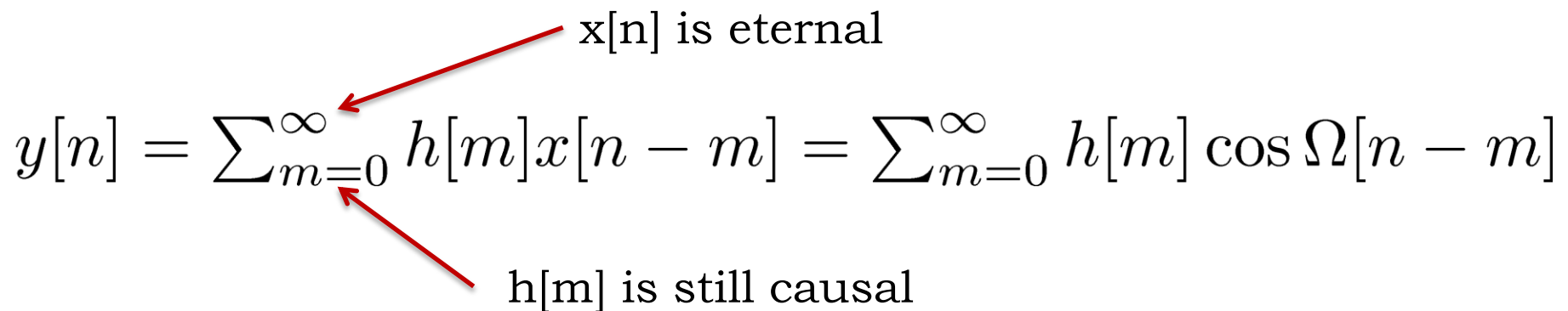
# Discrete Time Cosines $-\infty < n < \infty$



*Why is  $\pi$  highest frequency?*

# LTI Systems for Eternal Sines and Cosines

- Convolution Changes a little


$$y[n] = \sum_{m=0}^{\infty} h[m] x[n-m] = \sum_{m=0}^{\infty} h[m] \cos \Omega[n-m]$$

$x[n]$  is eternal

$h[m]$  is still causal

- If unit sample response finite length (L)

$$y[n] = \sum_{m=0}^L h[m] \cos \Omega[n-m]$$

***How do we simplify?***

# Frequency Response with Complex Exponentials

- From convolution

$$y[n] = \sum_{m=0}^L h[m] e^{j\Omega(n-m)}$$



Reorganizing

$$y[n] = \underbrace{\left( \sum_{m=0}^L h[m] e^{-j\Omega m} \right)} e^{j\Omega n}$$

A *complex* number if the sum converges



$$y[n] = H(e^{j\Omega}) e^{j\Omega n}$$

$H(e^{j\Omega})$ ,  $-\pi < \Omega < \pi$ , is the frequency response

# Going back to Cosines

- Suppose

$$x[n] = \cos \Omega n = \frac{1}{2} e^{j\Omega n} + \frac{1}{2} e^{-j\Omega n}$$



$$y[n] = \left( \sum_{m=0}^L h[m] e^{-j\Omega m} \right) \frac{e^{j\Omega n}}{2} + \left( \sum_{m=0}^L h[m] e^{j\Omega m} \right) \frac{e^{-j\Omega n}}{2}$$

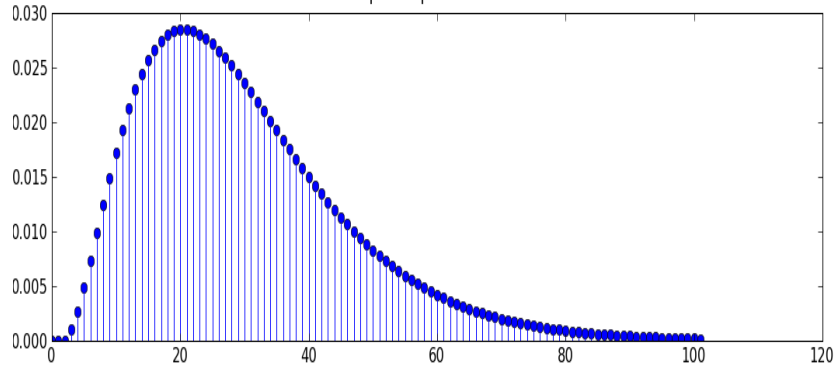


$$y[n] = \frac{1}{2} H(e^{j\Omega}) e^{j\Omega n} + \frac{1}{2} H(e^{-j\Omega}) e^{-j\Omega n}$$

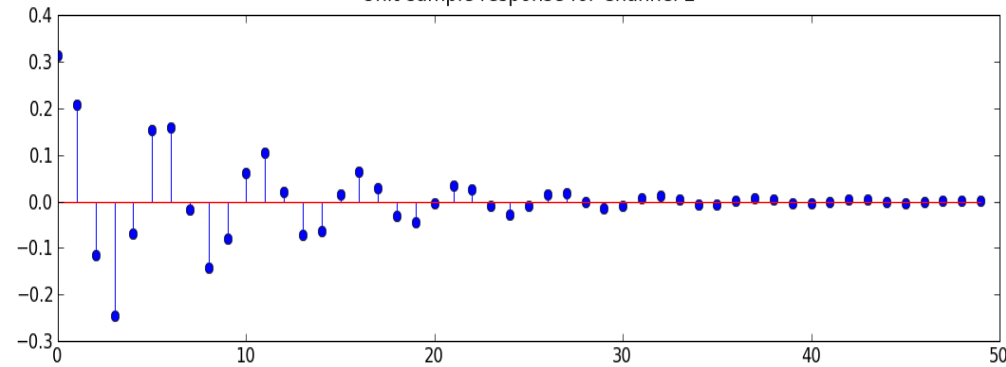
- Just using LTI superposition property!

# Channel 1 and 2 Frequency Responses

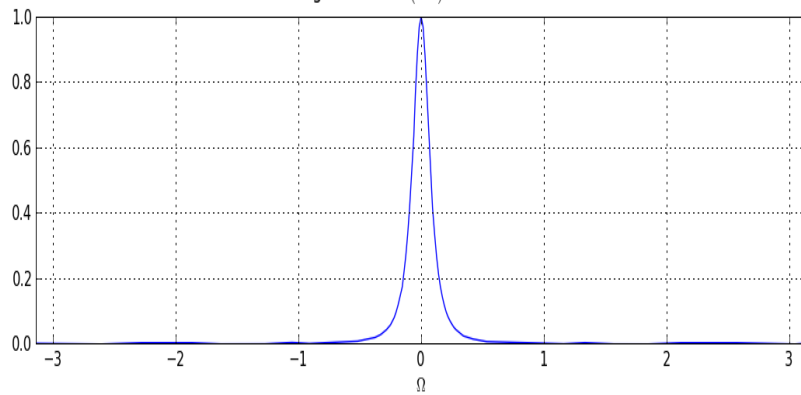
Unit-sample response for Channel 1



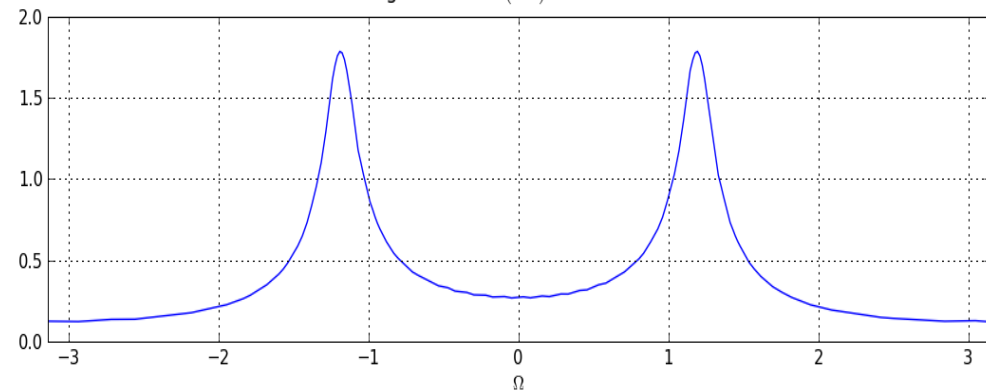
Unit-sample response for Channel 2



Magnitude of  $H(e^{j\Omega})$  for Channel 1



Magnitude of  $H(e^{j\Omega})$  for Channel 2

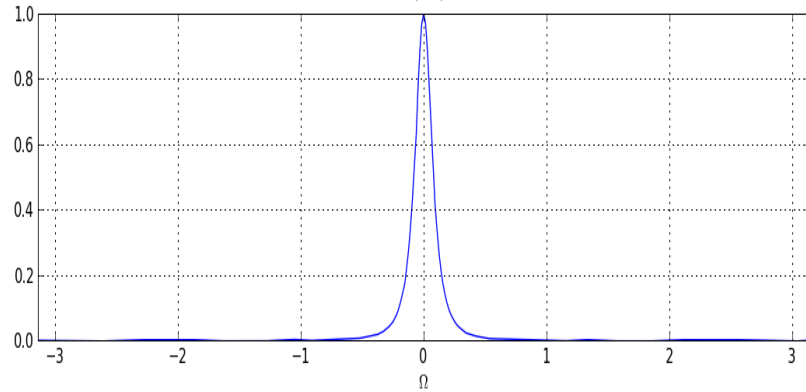


# Channel 1 and 2 Frequency Responses

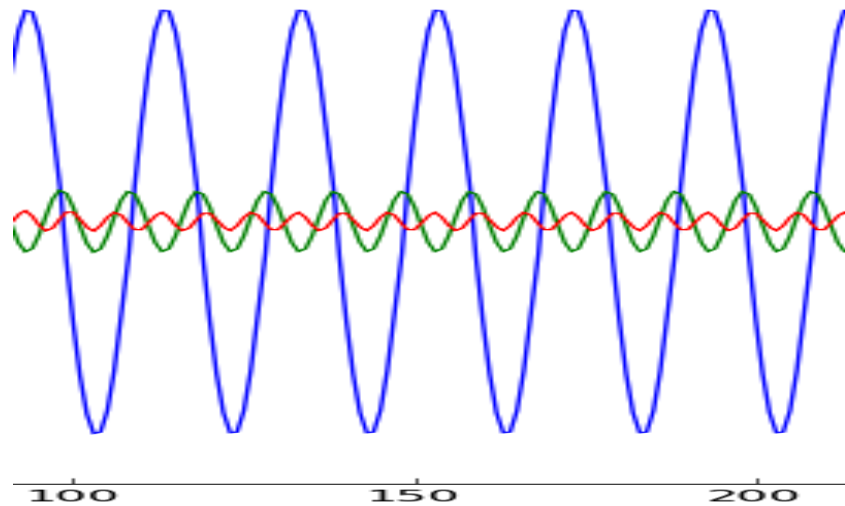
$$x[n] = \cos(\Omega n)$$

$$\Omega = \left\{ \frac{\pi}{10}, \frac{2\pi}{10}, \frac{3\pi}{10} \right\}$$

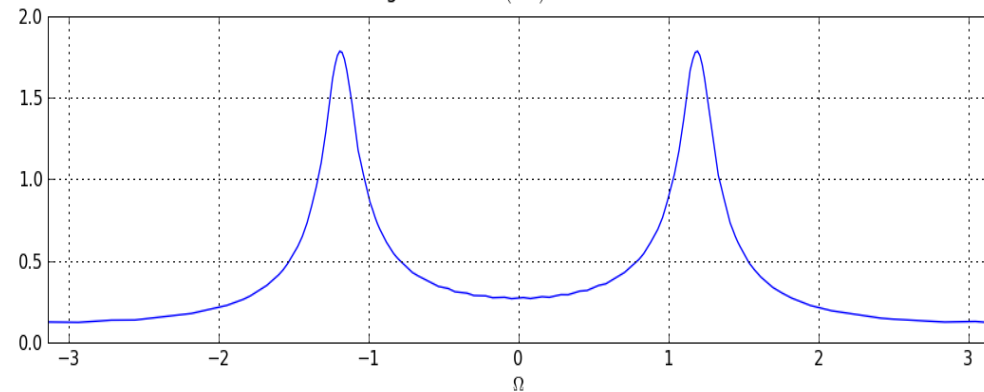
Magnitude of  $H(e^{j\Omega})$  for Channel 1



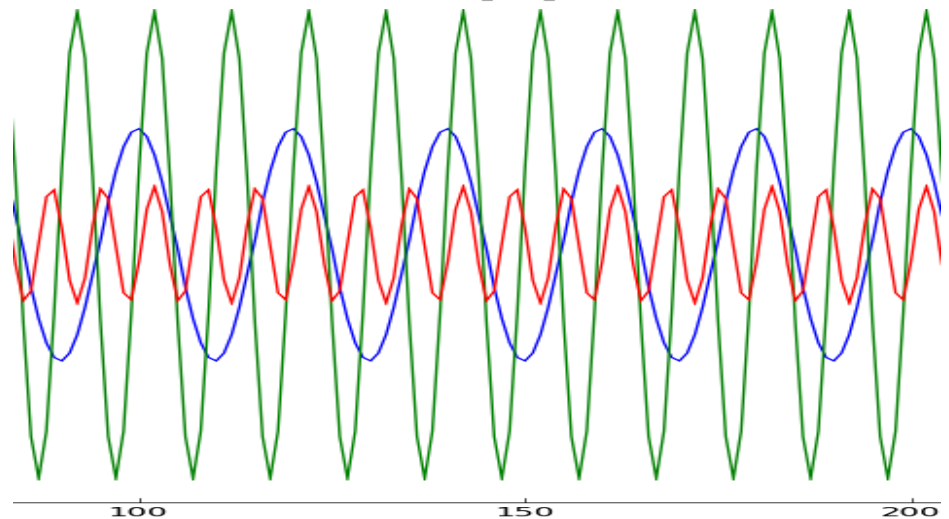
$y[n]$



Magnitude of  $H(e^{j\Omega})$  for Channel 2



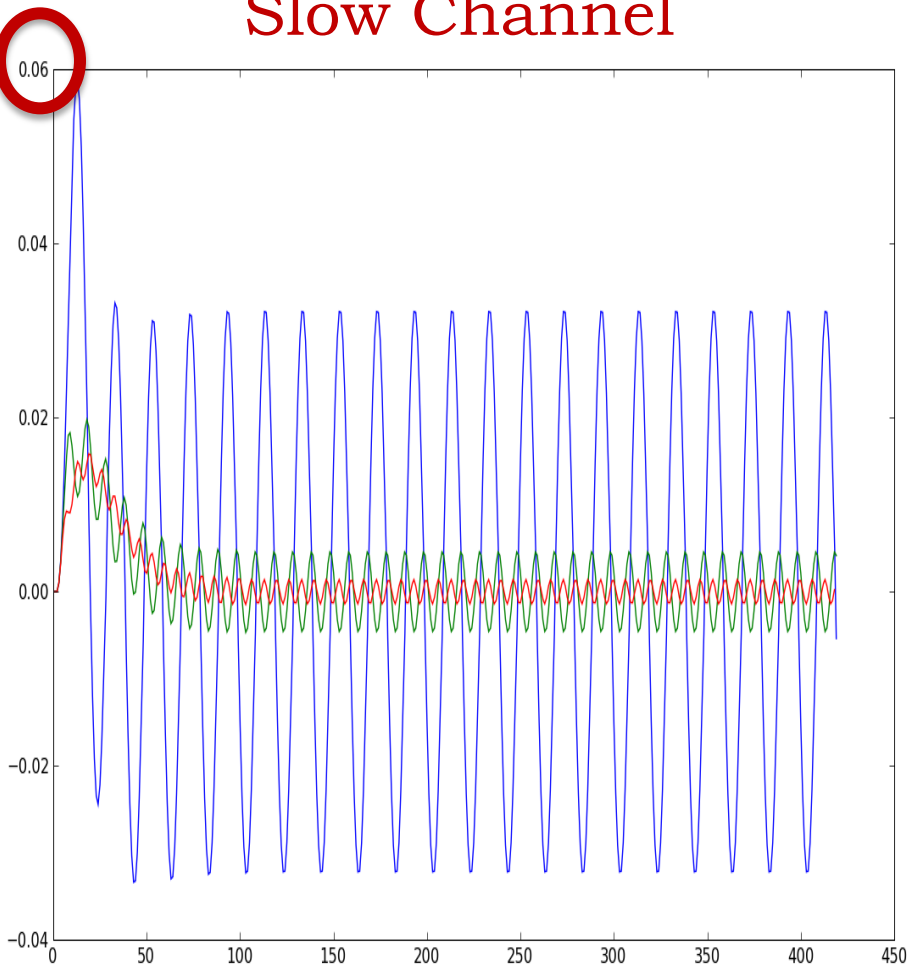
$y[n]$



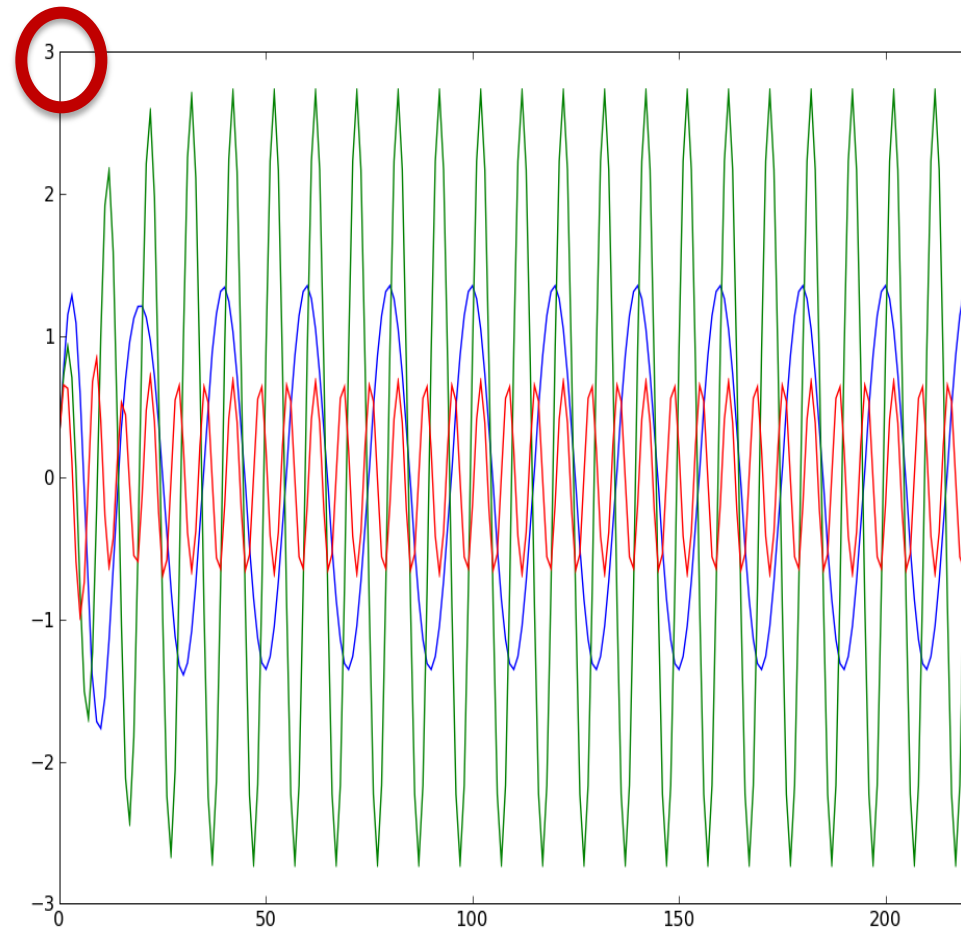
# Non-eternal Example: Cosine starts at zero

$$x[n] = \cos(\Omega n)u[n] \quad \Omega = \left\{ \frac{\pi}{10}, \frac{2\pi}{10}, \frac{3\pi}{10} \right\}$$

Slow Channel



Fast Channel



# Summary and Next Time

- Frequency Division Multiplexing

- Eternal frequencies do not mix when input to LTI systems
- Frequency division multiplexing

$$x[n] = A_1 e^{j\Omega_1 n} + \dots + A_K e^{j\Omega_K n}$$

- Frequency Response

$$H(e^{j\Omega}) \quad \Omega \in [-\pi, \pi]$$

- Next Two Weeks

- Need Filters at receiver to separate messages.
- How do we analyze more complicated messages

$$x[n] = x_1[n] e^{j\Omega_1 n} + \dots + x_K[n] e^{j\Omega_K n}$$

- Do the modulated messages still stay separated?
- Fourier Analysis and Spectrum
- Modulation Schemes