

INTRODUCTION TO EECS II

DIGITAL

COMMUNICATION

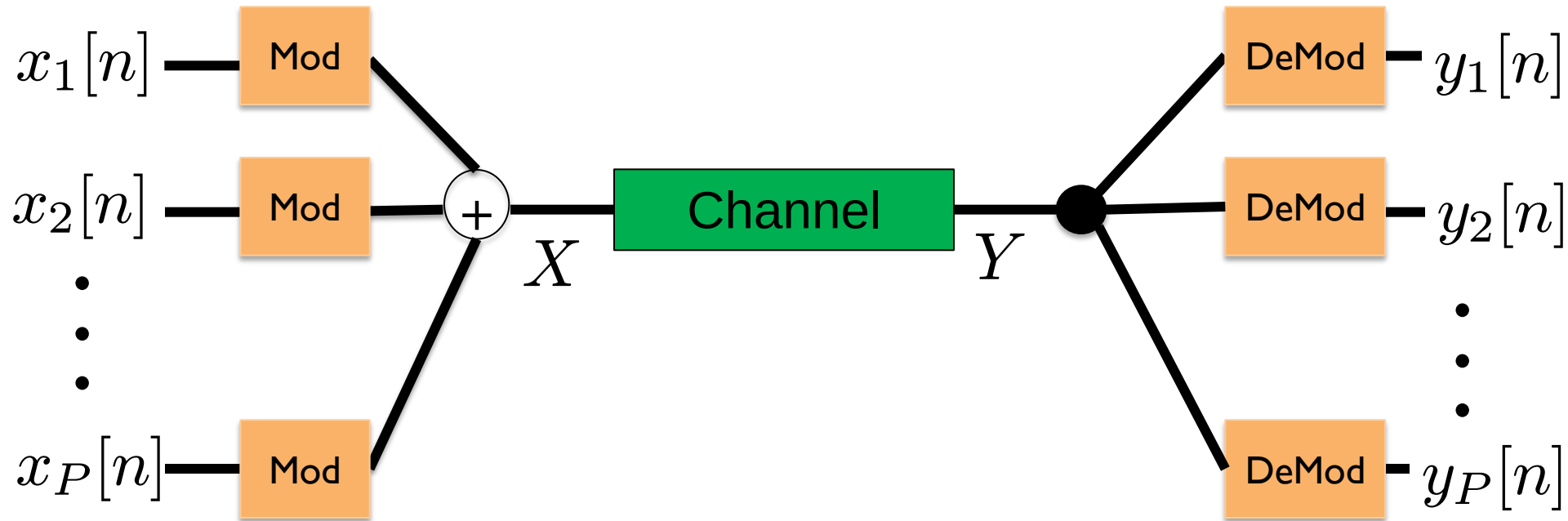
SYSTEMS

6.02 Fall 2010

Lecture #14

- Frequency Domain Sharing
- Spectrum and the Fourier Series
- Fourier Series Examples
- Rise Time and Spectrum

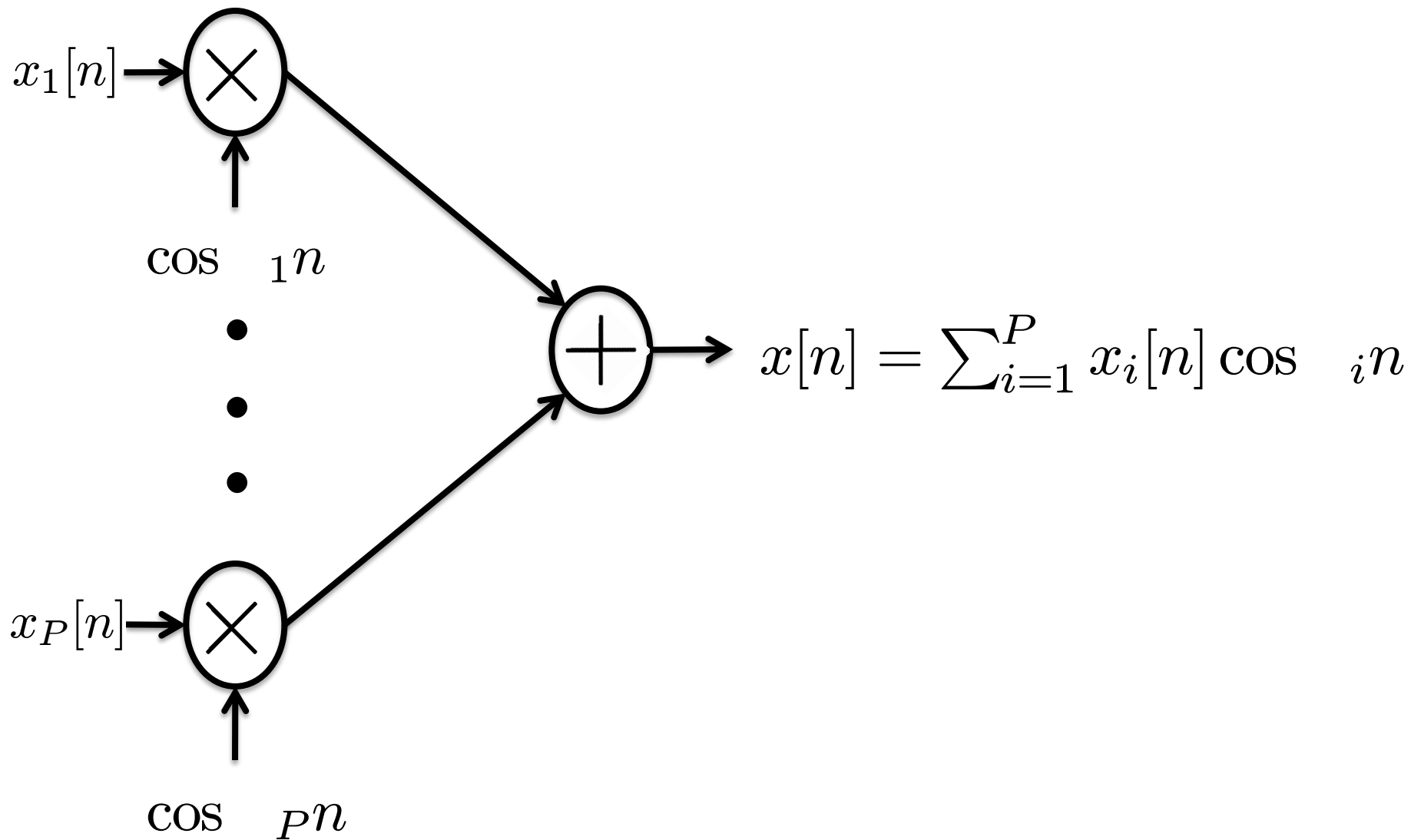
Frequency Domain Sharing - Big Picture



- Questions

- What is Modulation (Mod in figure)?
- What do typical X 's look like?
- What should the Demodulator be?
- What is the relation between $x_i[n]$ and $y_i[n]$

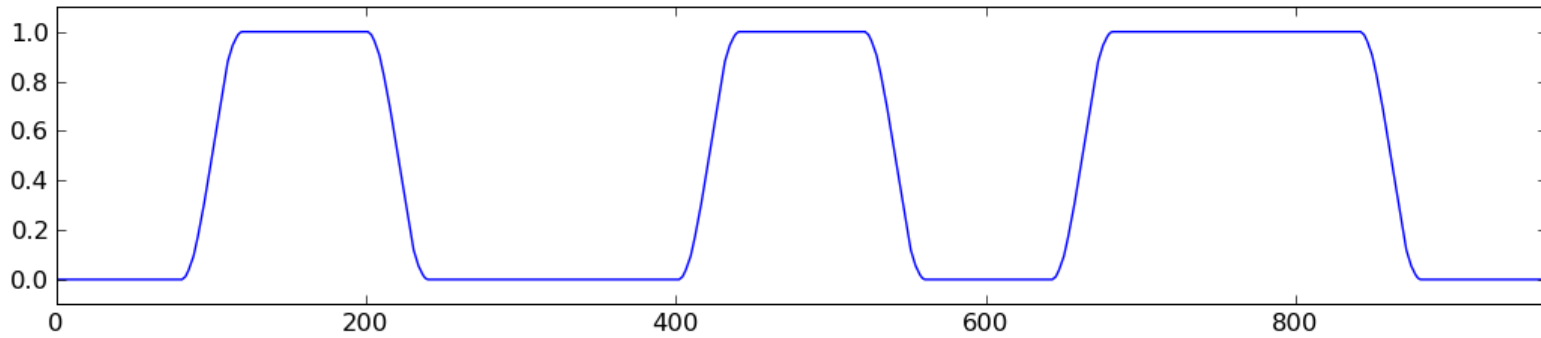
Cosine Modulation



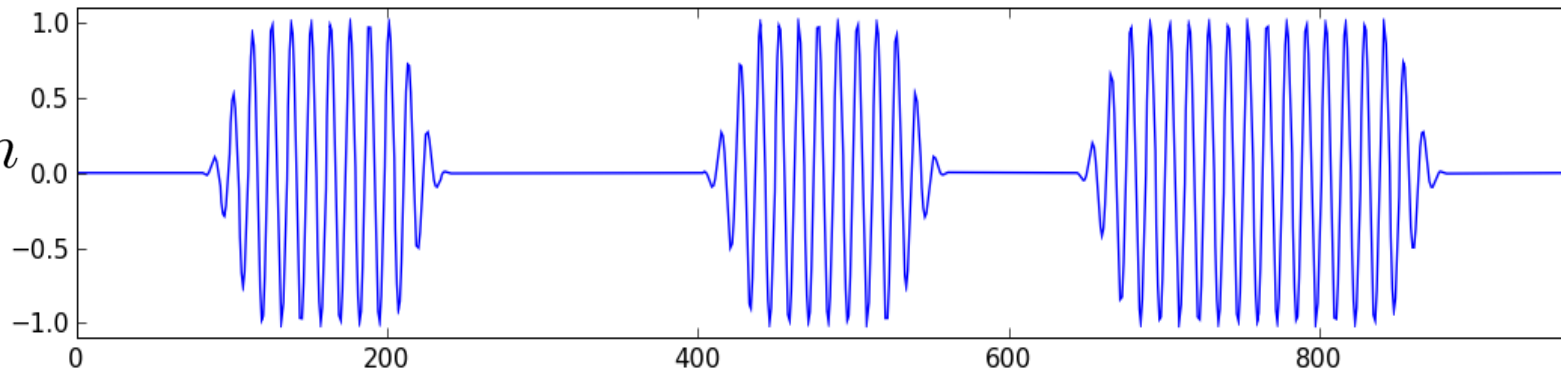
Ideal Channel Case $Y = X$

$$y[n] = \sum_{i=1}^P x_i[n] \cos \Omega_i n$$

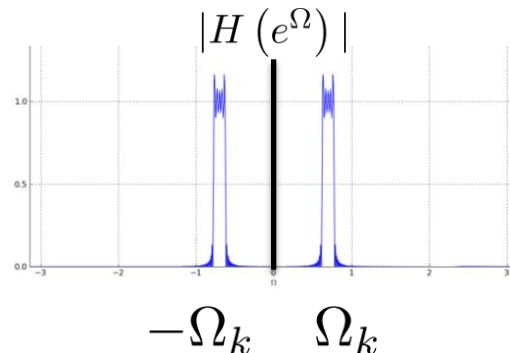
$x_i[n]$



$x_i[n] \cos 0.5n$



$$y[n] = \sum_{i=1}^P x_i[n] \cos \Omega_i n$$



?

(5)

What we have so far

If $x[n] = e^{j\Omega n}$ is the input to an LTI channel,

Then

$$y[n] = \underbrace{H(e^{j\Omega})}_{\text{Channel Frequency Response}} e^{j\Omega n}$$

Channel
Frequency
Response

Problem using F.D.M. to send messages

$$x[n] = \sum_{i=1}^p \underbrace{x_i[n]}_{i^{\text{th}} \text{ message}} \underbrace{e^{j\Omega_i n}}_{\substack{\text{frequency for the} \\ i^{\text{th}} \text{ message}}}$$

How do we analyze $x_i[n] e^{j\Omega_i n}$

$$x_i[n] e^{j\Omega_i n} \rightarrow \boxed{H} \rightarrow \cancel{H(e^{j\Omega_i})} x_i[n] e^{j\Omega_i n}$$

Why?

Need a new 1001: 1207-5-12 (5)

Periodic Approximation

For some large value of N ,

Assume $X_i[n+N] e^{j\Omega_i(n+N)} = X_i[n] e^{j\Omega_i n}$

Why? leads to tractable analysis

1st Consequence of periodic assumption

Since $e^{j\Omega_i(n+N)} = e^{j\Omega_i n}$

Then $e^{j\Omega_i n} e^{j\Omega_i N} = e^{j\Omega_i n}$

$e^{j\Omega_i N}$ must equal 1

$$e^{j\Omega_i N} = \underbrace{\cos \Omega_i N}_{\text{must be } 1} + j \underbrace{\sin \Omega_i N}_{\text{must be zero}}$$

$\Omega_i N$ must equal $2\pi, 4\pi, 6\pi, \dots$

Since $-\pi \leq \Omega_i \leq \pi$

$$\Omega_i = 0, \pm \frac{2\pi}{N}, \pm \frac{4\pi}{N}, \dots, \pm \frac{2\pi}{N} \cdot \frac{N}{2}$$

$\leftarrow$ assuming N is even!

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2nd Consequence of Periodic Assumption

If $x[n]$ is real and periodic with period N , and N is even:

No $k = \frac{N}{2}$ or $k=0$ term for $\sin$, as $\sin(0) = \sin\pi = 0$

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} a_k \sin \frac{2\pi}{N} kn + \sum_{k=0}^{\frac{N}{2}} b_k \cos \frac{2\pi}{N} kn$$

Discrete-Time Fourier Series. (DTFS)

$$\left(\frac{N}{2} - 2\right) + \left(\frac{N}{2} + 1\right) \text{ coeffs} \\ = N - 1 \text{ coeffs}$$

DTFS using Complex Exponentials

Removes a redundancy

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} \underbrace{x[k]}_{\text{complex coefficients}} e^{j \frac{2\pi}{N} kn}$$

Note: $2 \cdot \frac{N}{2} = N$ complex coeffs in DTFS, and only N real unique values for x , $x[0]$, $x[1]$, ..., $x[N-1]$.

(8)

D T F S Degrees of FreedomN is even caseIF $X[n]$ is real for all n then $X[k] e^{j \frac{2\pi}{N} kn} + X[-k] e^{-j \frac{2\pi}{N} kn}$
must be real!

$$\Rightarrow \operatorname{Re}(X[k]) = \operatorname{Re}(X[-k])$$

$$\operatorname{Im}(X[k]) = -\operatorname{Im}(X[-k])$$

So $(2 \cdot \frac{N}{2}) = N$ complexcoefficients have only N

degrees of freedom

$$\begin{matrix} N-2 \\ \text{constraints} \end{matrix} \left\{ \begin{array}{l} \operatorname{Re}(X[1]) = \operatorname{Re}(X[-1]), \dots, \operatorname{Re}(X[\frac{N}{2}-1]) = \operatorname{Re}(X[-\frac{N}{2}+1]) \\ \operatorname{Im}(X[1]) = -\operatorname{Im}(X[-1]), \dots, \operatorname{Im}(X[\frac{N}{2}-1]) = -\operatorname{Im}(X[-\frac{N}{2}+1]) \end{array} \right.$$

And

$$\begin{matrix} 1 \\ \text{constraint} \end{matrix} \left\{ \begin{array}{l} \operatorname{Re}(X[0]) = \operatorname{Re}(X[0]) \quad \operatorname{Im}(X[0]) = -\operatorname{Im}(X[0]) \end{array} \right.$$

 $\Rightarrow X[0]$ must be real

$$\begin{matrix} 3 \text{ last} \\ \text{constraints} \end{matrix} \left\{ \begin{array}{l} \text{Finally} \quad e^{j \frac{2\pi}{N} (\frac{N}{2}) n} = e^{-j \frac{2\pi}{N} (\frac{N}{2}) n} = 1^n \\ X[\frac{N}{2}] \text{ is real} \end{array} \right.$$

(4)

Computing Fourier Coefficients

Multiply by test exponential

$$e^{-j\frac{2\pi}{N}ln} x[n] = e^{-j\frac{2\pi}{N}ln} \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] e^{j\frac{2\pi}{N}kn}$$

Never zero!

DTFS

Sum both sides over $0 \leq n \leq N-1$

$$\begin{aligned} \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}ln} x[n] &= \sum_{n=0}^{N-1} \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] e^{j\frac{2\pi}{N}(k-l)n} \\ &= \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] \sum_{n=0}^{N-1} e^{j\frac{2\pi}{N}(k-l)n} \\ &= 0 \quad k \neq l \end{aligned}$$

$$= N \quad k=l$$

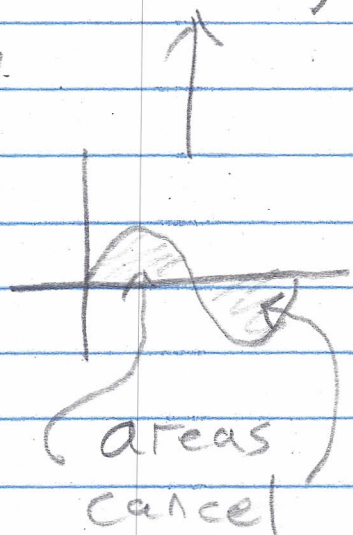
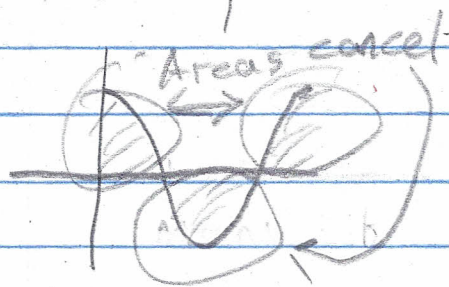
$$\Rightarrow \frac{1}{N} \sum_{n=0}^{N-1} e^{-j\frac{2\pi}{N}ln} x[n] = X[l]$$

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Why is $\sum_{n=0}^{N-1} e^{j \frac{2\pi}{N} (k-l)n} = 0 \quad k \neq l$

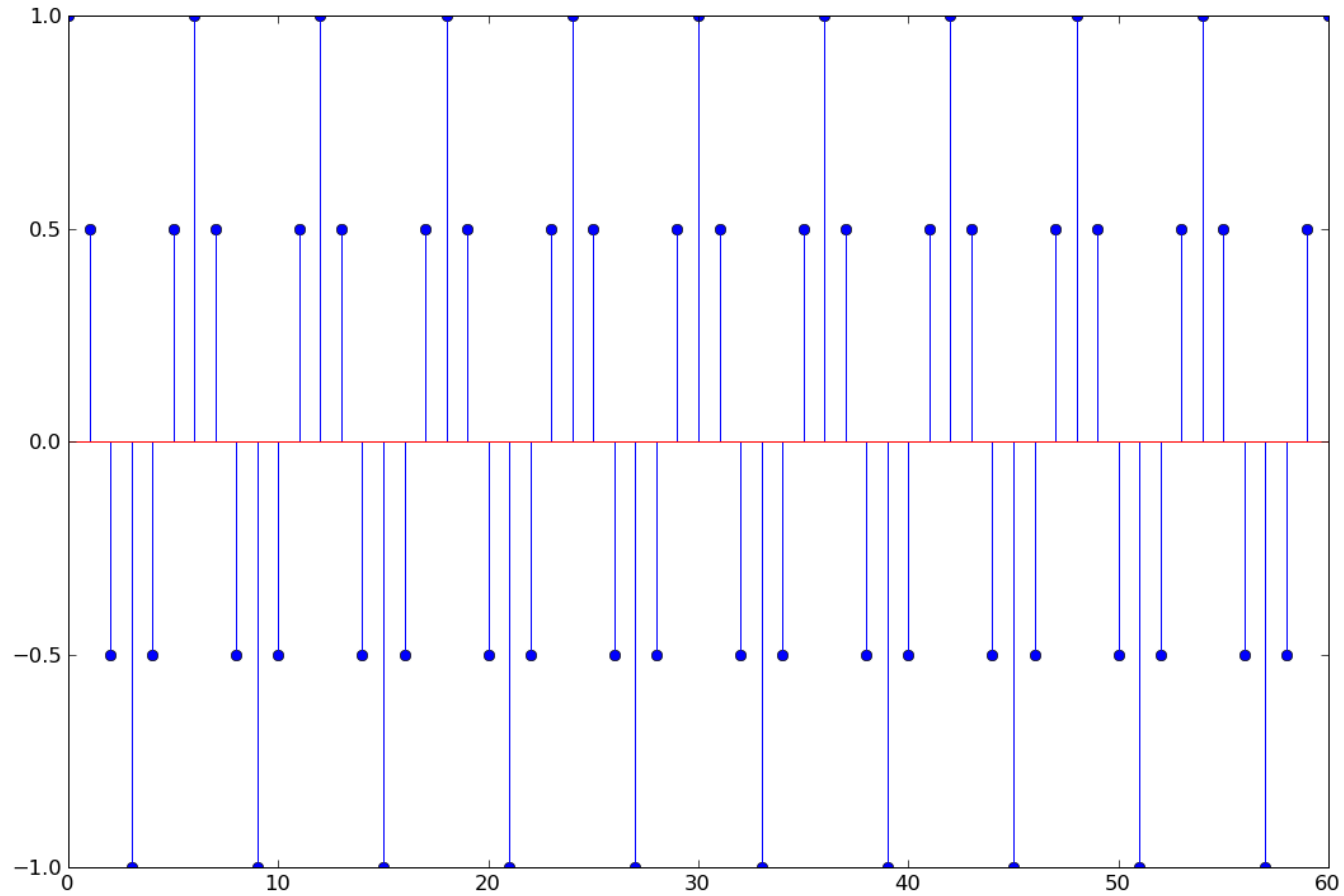
Suppose $k-l = 1$

$$\sum_{n=0}^{N-1} e^{j \frac{2\pi}{N} n} = \sum_{n=0}^{N-1} \left(\cos \frac{2\pi}{N} n + j \sin \frac{2\pi}{N} n \right)$$



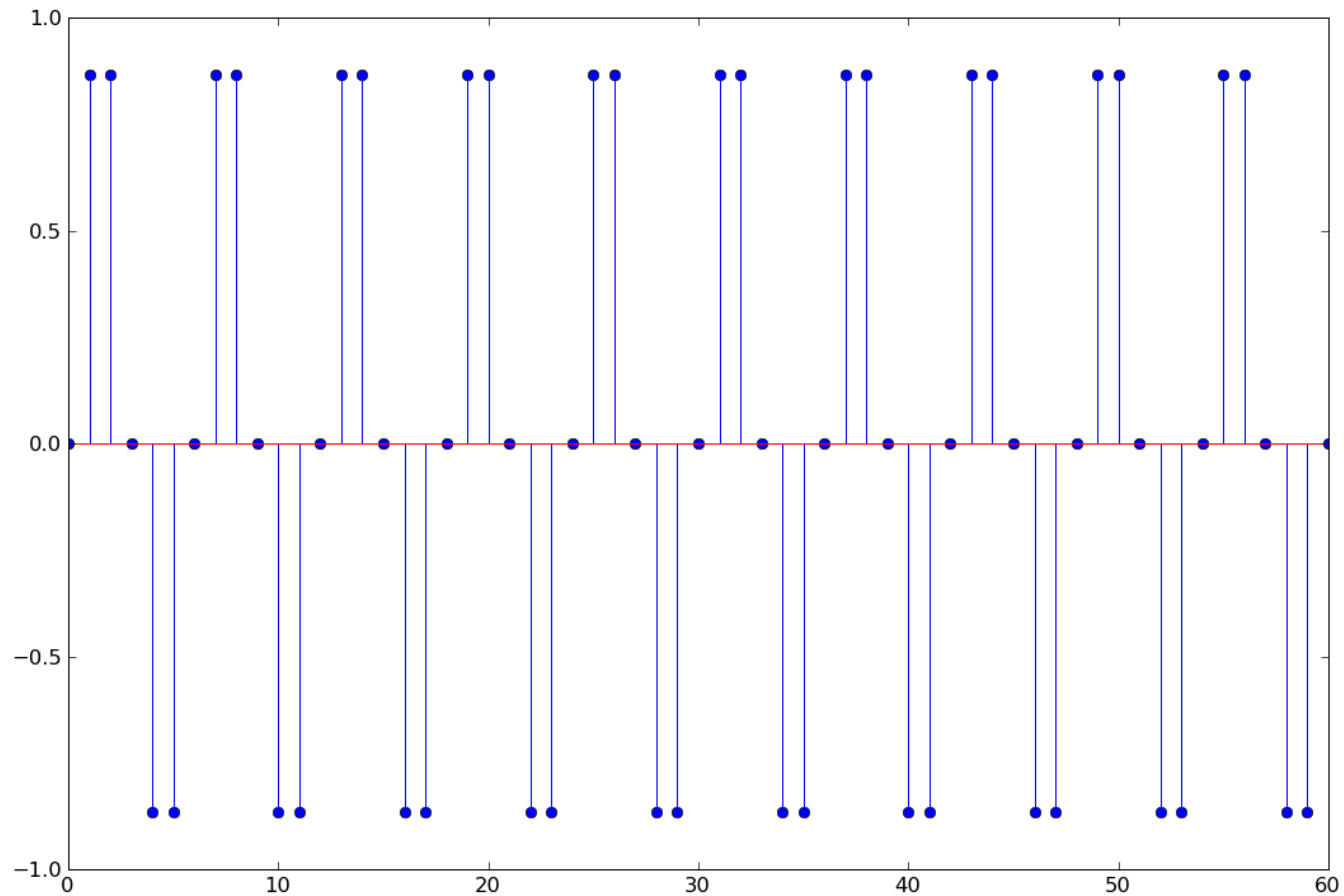
Cosine in Time

$$x[n] = \cos \frac{\pi}{3} n = \frac{1}{2} e^{j \frac{\pi}{3} n} + \frac{1}{2} e^{-j \frac{\pi}{3} n}$$



Sine in Time

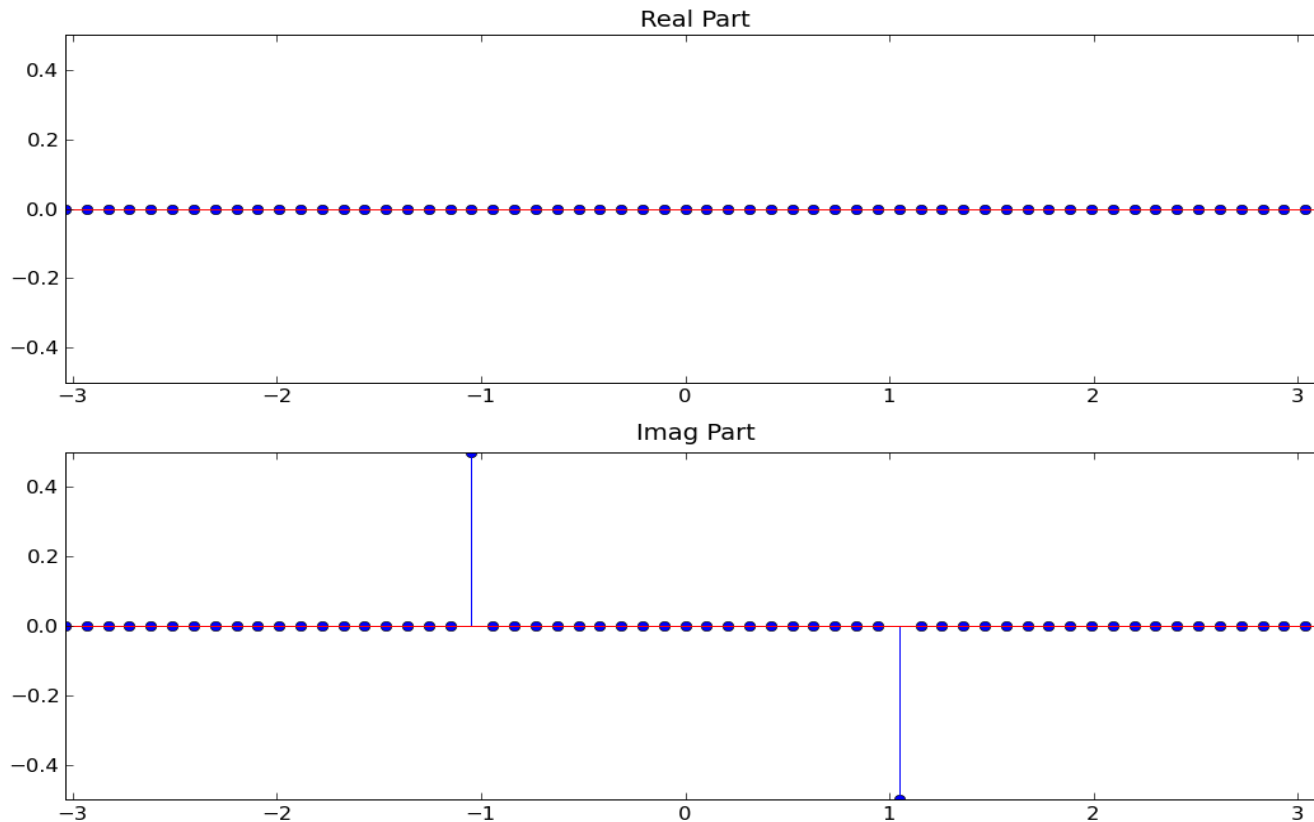
$$x[n] = \sin \frac{\pi}{3}n = \frac{-j}{2}e^{j\frac{\pi}{3}n} + \frac{j}{2}e^{-j\frac{\pi}{3}n}$$



Sine Fourier Series, 2 Imaginary values

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] e^{j \frac{2\pi}{N} kn} = \frac{-j}{2} e^{j \frac{\pi}{3} n} + \frac{j}{2} e^{-j \frac{\pi}{3} n}$$

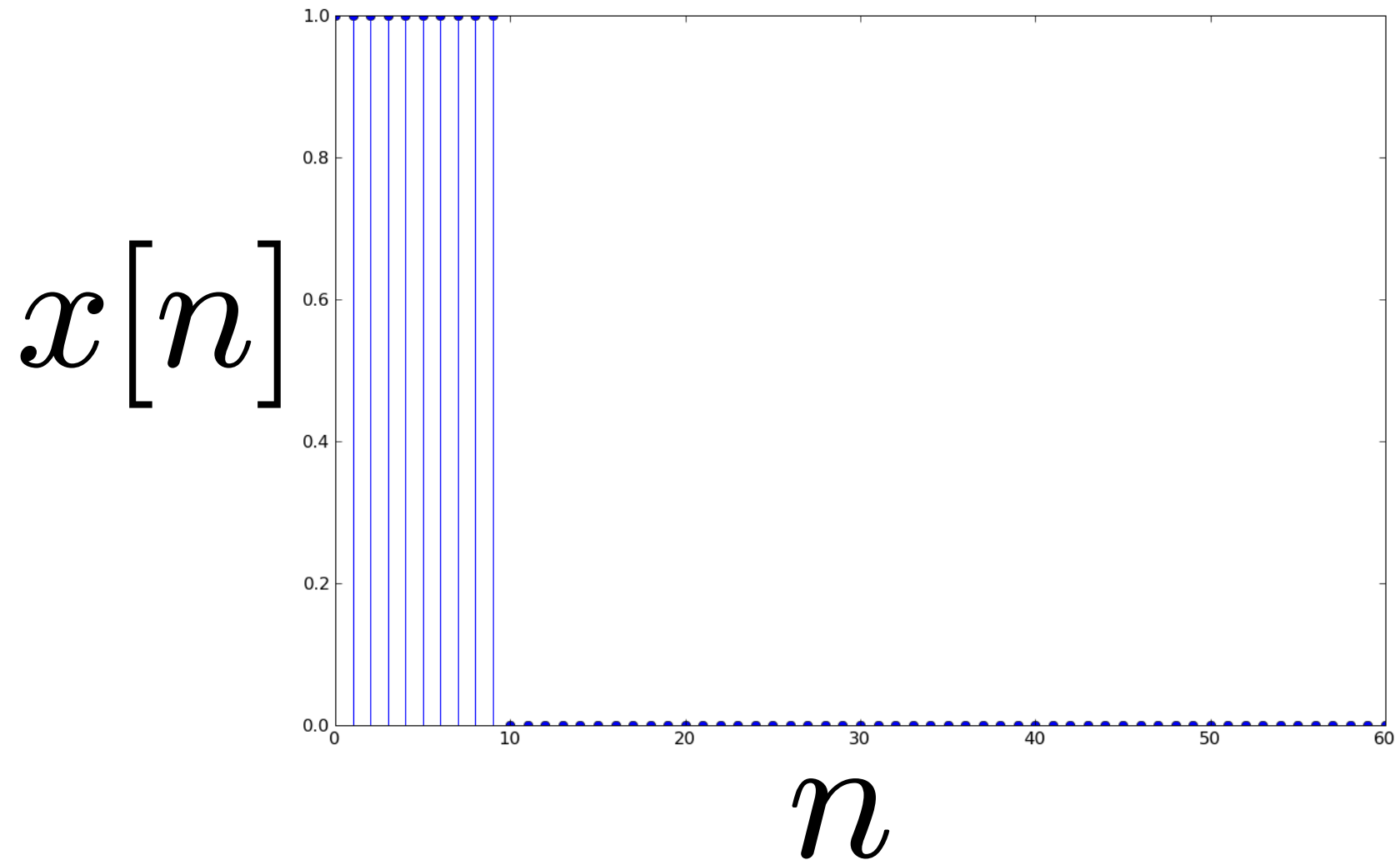
$X[k]$



$$\frac{2\pi}{N} k$$

Note
horizontal
axis ranges
over
-pi to pi!

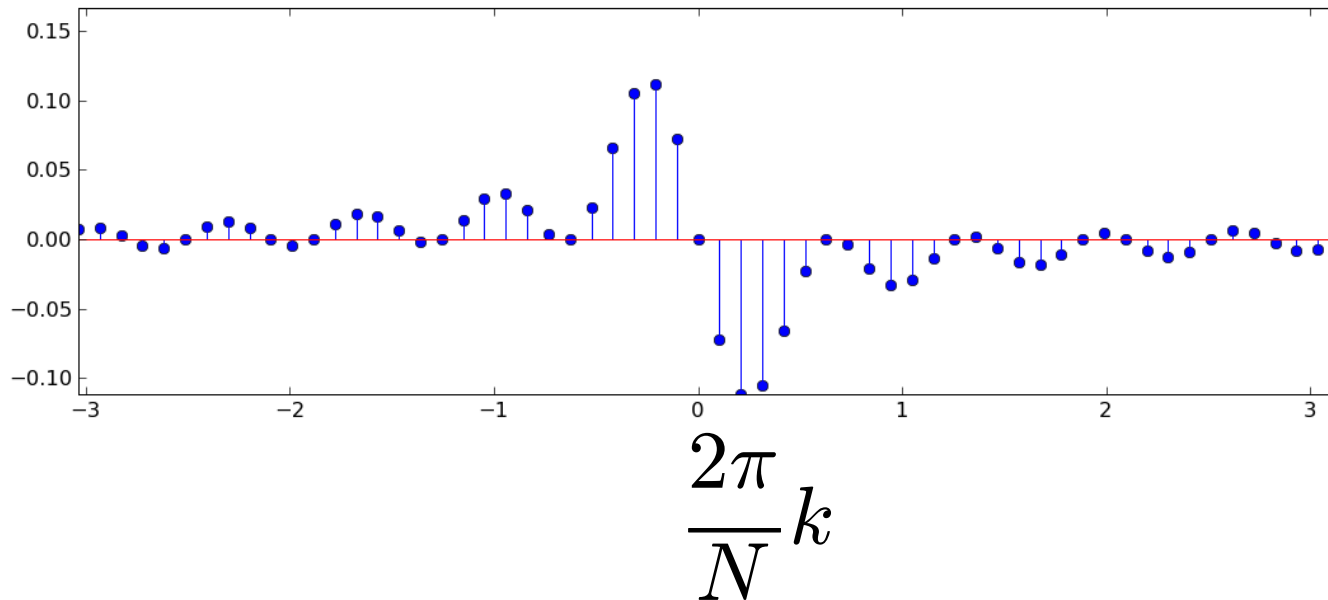
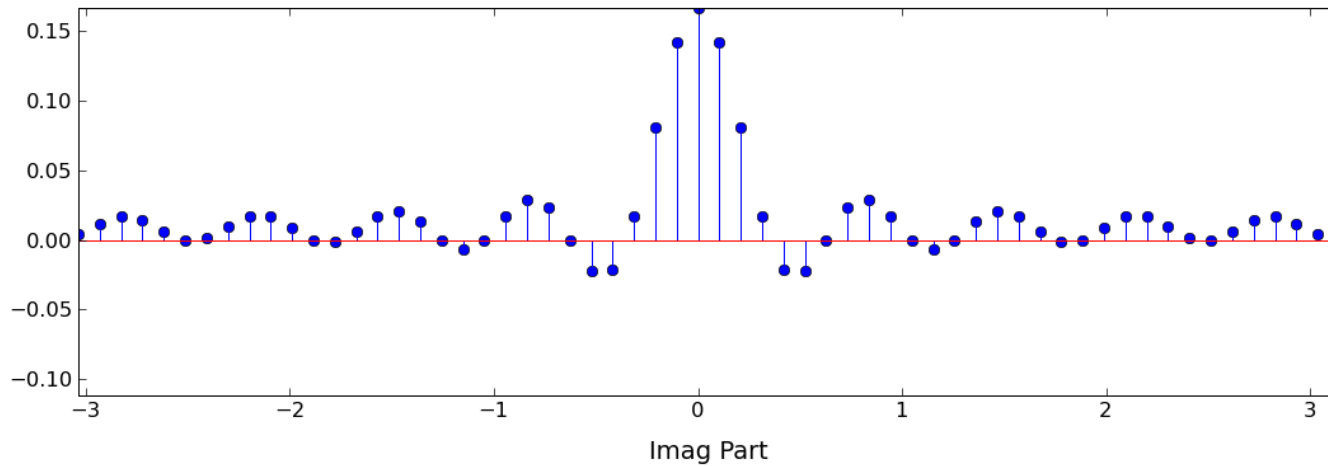
Fast Rise Pulse



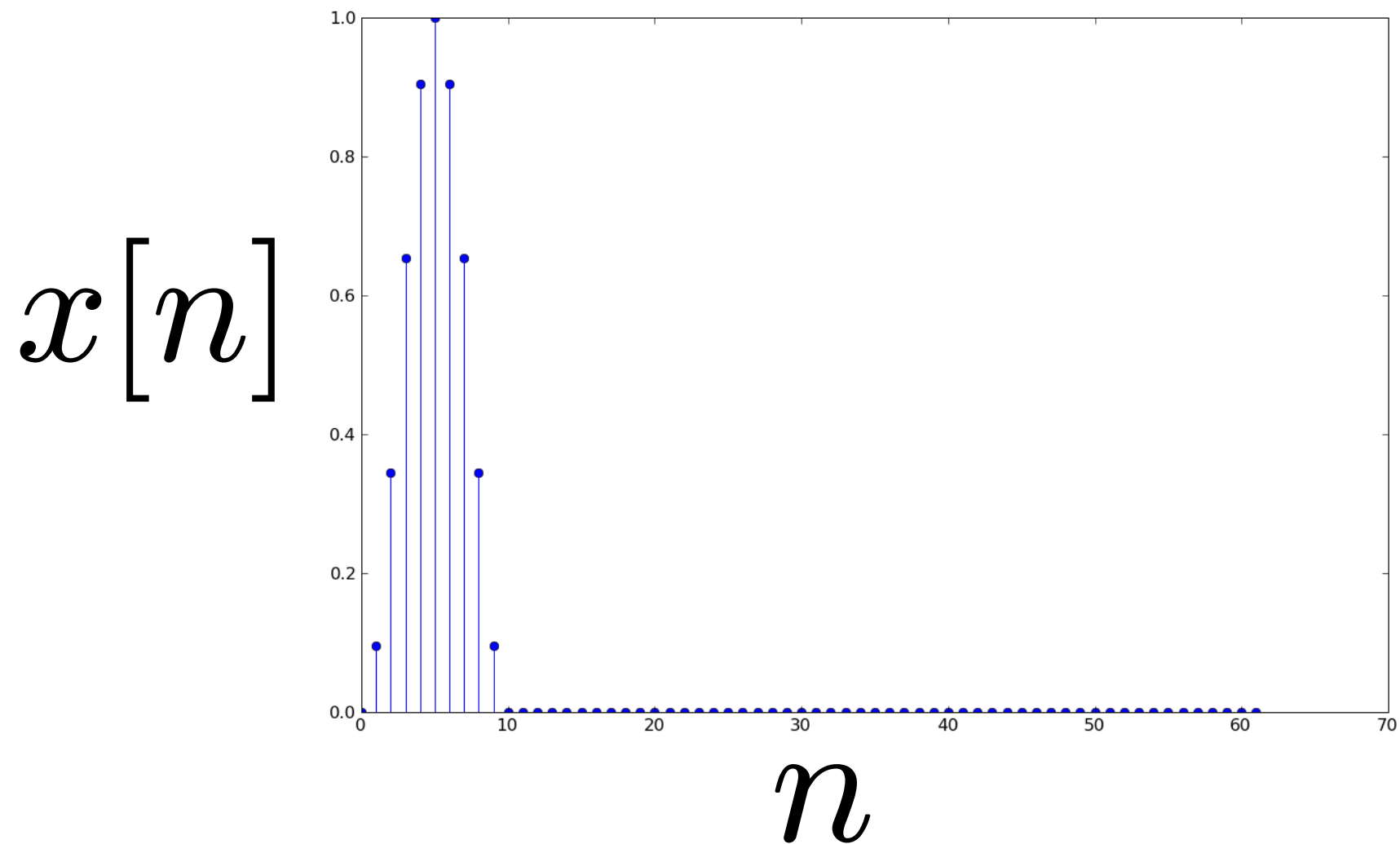
Spectrum of Fast Rise Pulse

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] e^{j \frac{2\pi}{N} kn}$$

$X[k]$



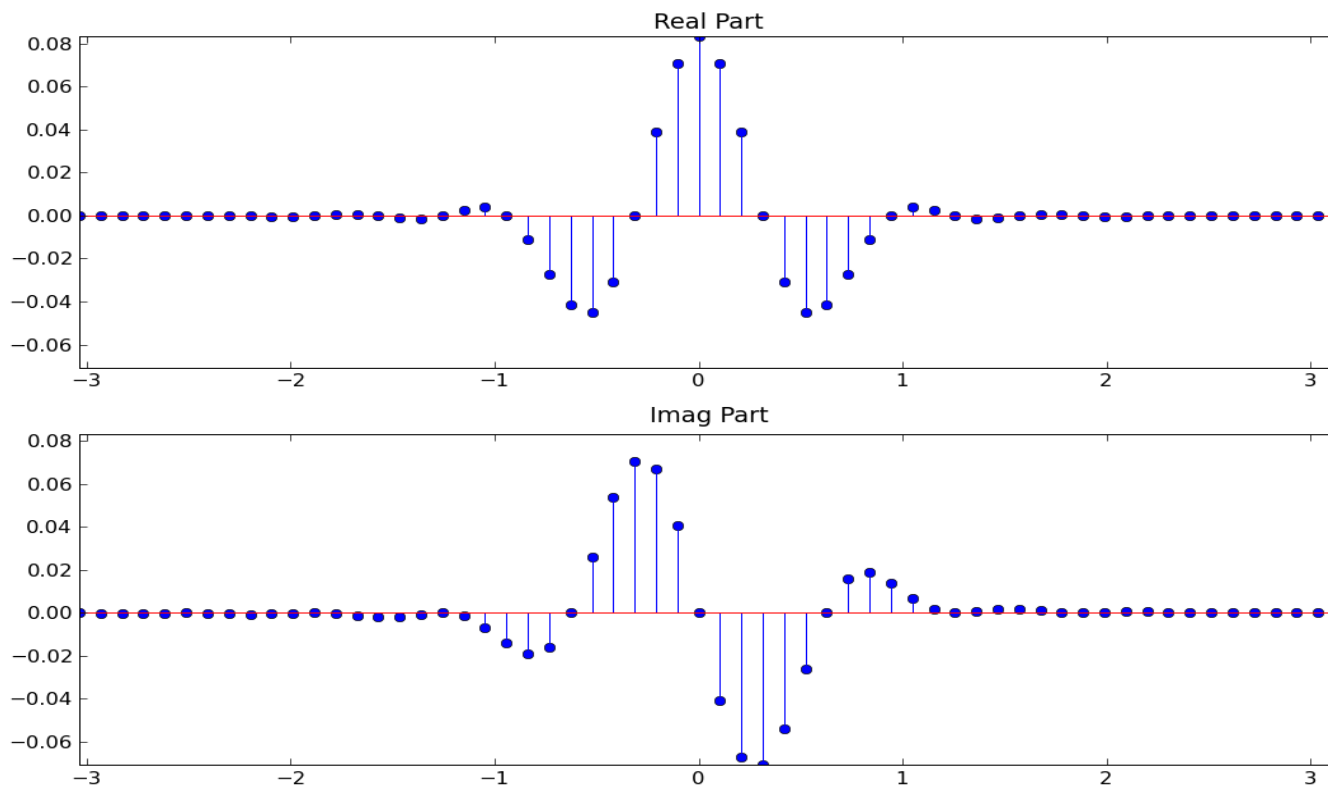
Slow Rise Pulse



Spectrum Slow Rise Pulse

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k] e^{j \frac{2\pi}{N} kn}$$

$X[k]$



$$\frac{2\pi}{N}k$$

Key Fourier Series Advantage



$$y[n] = \sum_{m=0}^n h[m]x[n-m]$$

OR

$$x[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} X[k]e^{j\frac{2\pi}{N}kn} \quad \longrightarrow \quad y[n] = \sum_{k=-\frac{N}{2}}^{\frac{N}{2}-1} H(e^{j\frac{2\pi}{N}k})X[k]e^{j\frac{2\pi}{N}kn}$$

An orange arrow points from the input equation to the output equation. A second orange arrow points from the term $H(e^{j\frac{2\pi}{N}k})$ in the output equation to the Fourier series definition below.

$$H(e^{j\omega}) = \sum_{m=0}^L h[m]e^{-j\omega m}$$

(19)

ExamplesSuppose $N = 20$ 1) What are the allowed Ω_i 's

$$\Omega_i = 0, \pm \frac{\pi}{10}, \pm \frac{\pi}{5}, \pm \frac{3\pi}{10}, \dots, \pm \frac{4\pi}{5}, \pm \frac{9\pi}{10}, \pm \pi$$

We will only
use $\pm \pi$ Ω

$$\left(e^{j \frac{2\pi}{N} \cdot \frac{N}{2} n} = e^{j \pi n} = e^{j \frac{2\pi}{N} \cdot \frac{N}{2} n} = e^{j \pi n} \right)$$

$$2) \quad X[n] = \cos\left(\frac{3\pi}{5} n + \phi\right)$$

$$= \frac{1}{2} \left(e^{j \left(\frac{2\pi}{20} \cdot 6n + \phi \right)} \right)$$

$$+ \frac{1}{2} \left(e^{-j \left(\frac{2\pi}{20} \cdot 6n + \phi \right)} \right)$$

$$= \frac{e^{j\phi}}{2} e^{j \frac{2\pi}{20} 6n} + \frac{e^{-j\phi}}{2} e^{-j \frac{2\pi}{20} 6n}$$

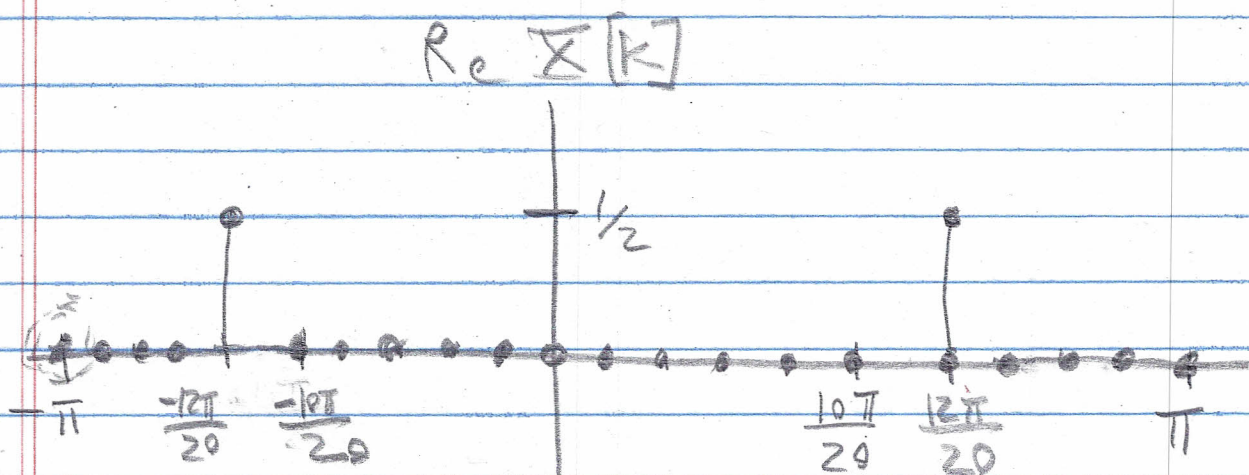
$$X[6] = \frac{e^{j\phi}}{2} \quad X[6] = \frac{e^{-j\phi}}{2}$$

$$X[k] = 0 \quad k \neq 6$$

(20)

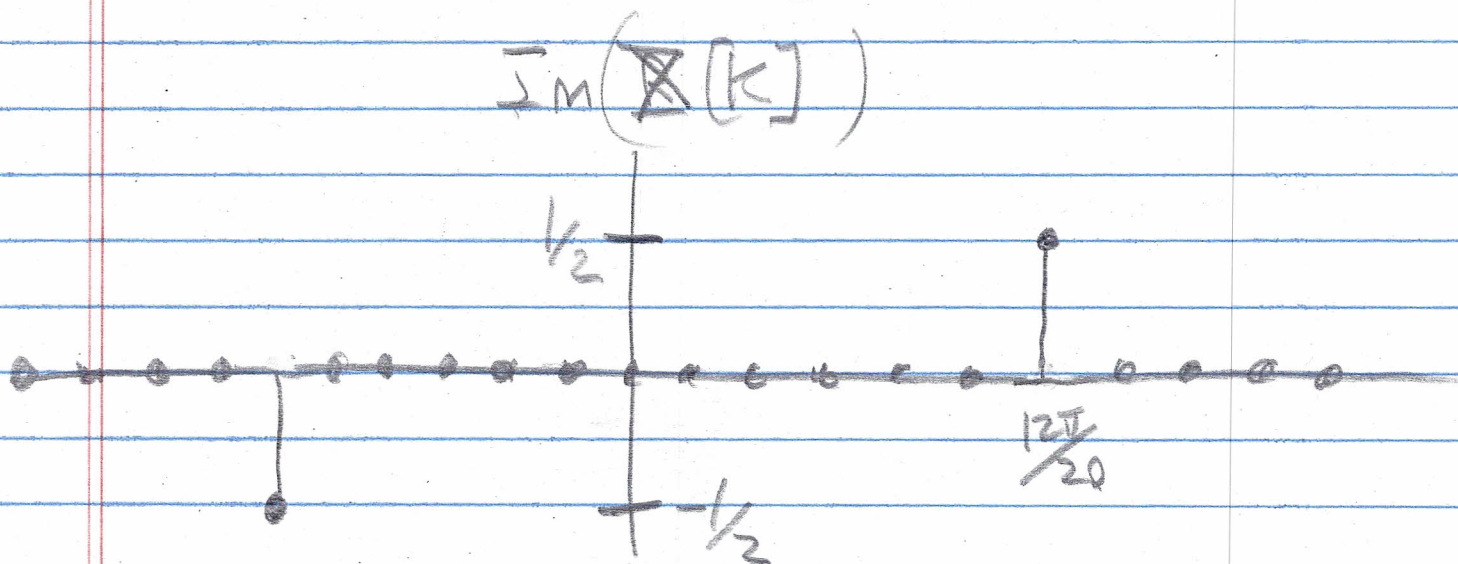
$$\text{If } \phi = 0$$

$$\text{Im}(X[k]) = 0$$



$$\text{If } \phi = \pi/2 \quad e^{j\phi/2} = \frac{1}{2}j \quad e^{-j\phi/2} = -\frac{1}{2}j$$

$$\text{Re}(X[k]) = 0$$

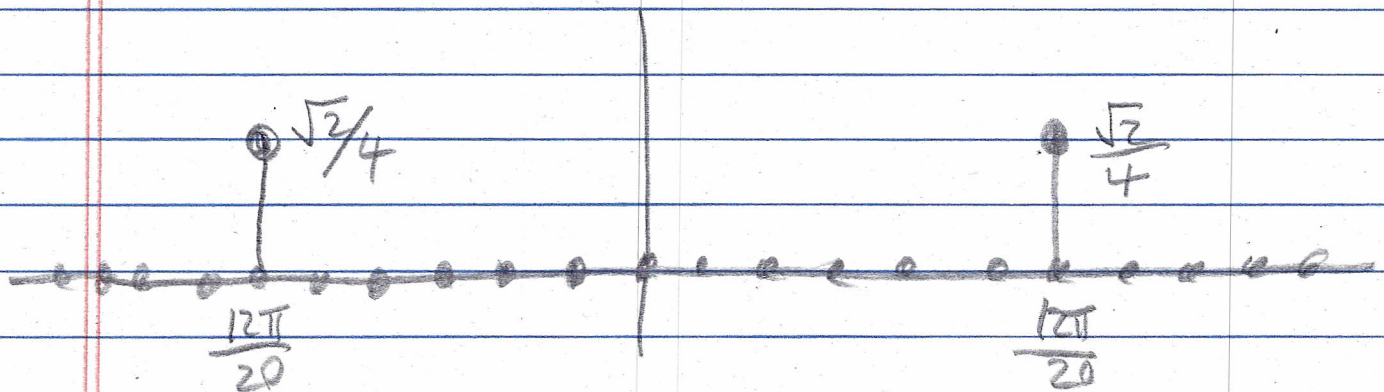


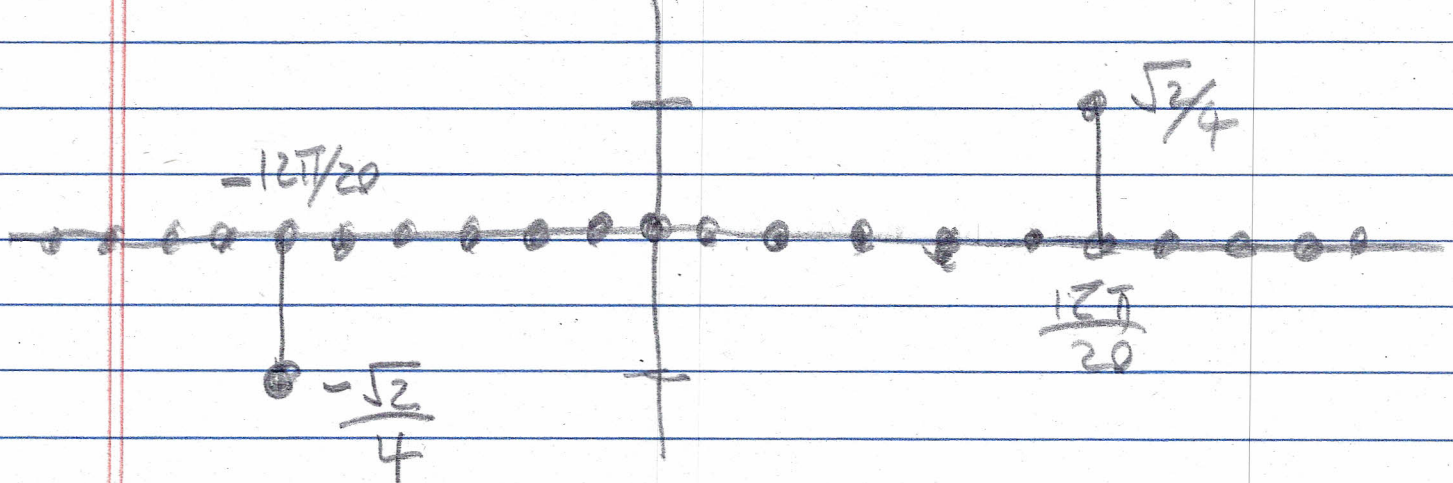
Note $\cos(\omega n + \pi/2) = -\sin(\omega n)$

(21)

$$\underline{\phi = \pi/4} \quad \frac{1}{2} e^{j\phi} = \frac{\sqrt{2}}{4} + \frac{\sqrt{2}}{4} j$$

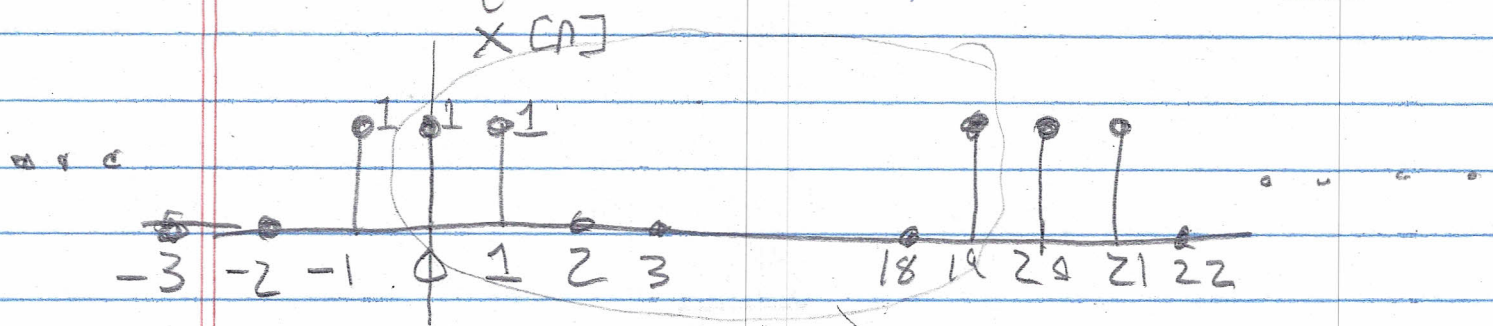
$$\frac{1}{2} e^{-j\phi} = \frac{\sqrt{2}}{4} - \frac{\sqrt{2}}{4} j$$

$$\text{Re}(X[k])$$


$$\text{Im}(X[k])$$


(22)

Square Pulse Case (N=20)



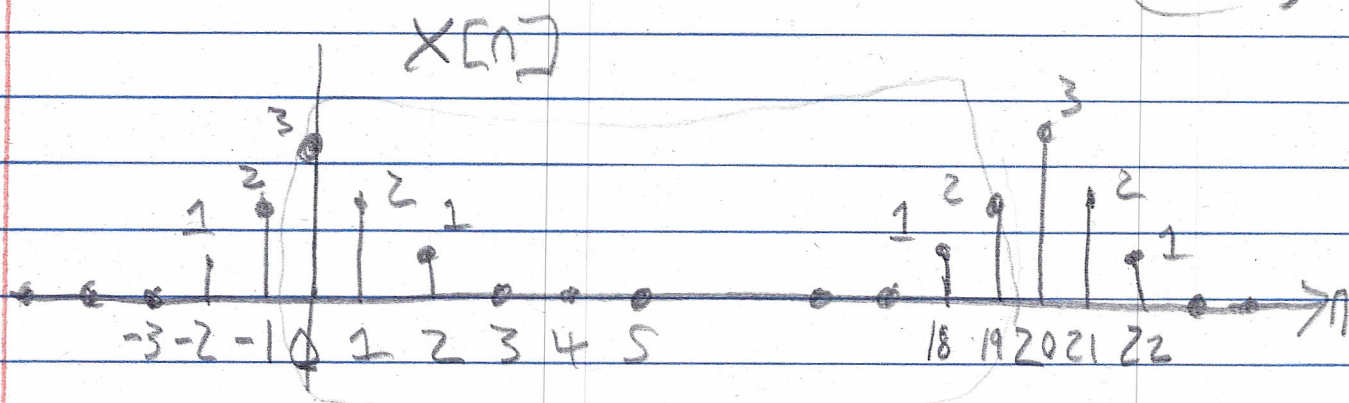
Repeats
periodically

$$\begin{aligned}
 X[k] &= \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi}{20} k n} \\
 &= \frac{1}{20} \left(1 \cdot e^{-j \frac{2\pi}{20} k \cdot 0} + 1 \cdot e^{-j \frac{2\pi}{20} k \cdot 1} \right. \\
 &\quad \left. + 1 \cdot e^{-j \frac{2\pi}{20} k \cdot 19} \right) \\
 &= \frac{1}{20} \left(1 + e^{-j \frac{2\pi}{20} k} + e^{-j \frac{2\pi \cdot 19}{20} k} \right) \\
 &= e^{+j \frac{2\pi}{20} k}
 \end{aligned}$$

$$= \frac{1}{20} \left(1 + 2 \cos\left(\frac{2\pi}{20} k\right) \right)$$

For arbitrary N $X[k] = \frac{1 + 2 \cos\left(\frac{2\pi}{N} k\right)}{N}$

(23)

Triangular Pulse Case (N=20)

$$\begin{aligned}
 X[k] &= \frac{1}{20} \left(3 \cdot 1 + 2 \cdot e^{-j \frac{2\pi}{20} k} + 1 \cdot e^{-j \frac{4\pi}{20} k} \right. \\
 &\quad \left. + 1 \cdot e^{-j \frac{2\pi \cdot 18}{20} k} + 2 \cdot e^{-j \frac{2\pi \cdot 19}{20} k} \right) \\
 &= \frac{1}{20} \left(3 + 4 \cos\left(\frac{2\pi}{20} k\right) + 2 \cos\left(\frac{4\pi}{20} k\right) \right)
 \end{aligned}$$

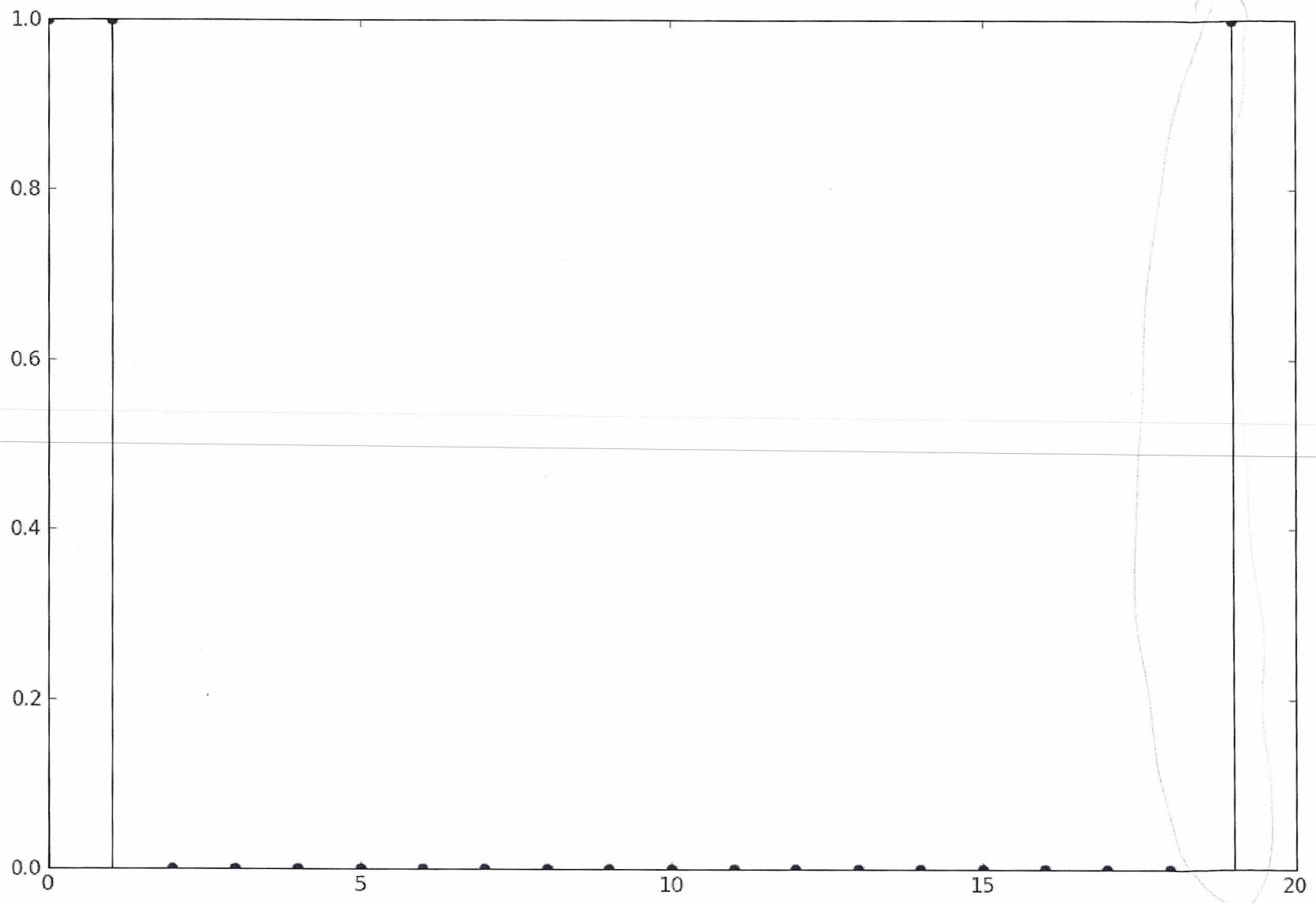
Note Compare to product of square pulses

$$\begin{aligned}
 &\left(\frac{1}{20} \left(1 + e^{-j \frac{2\pi}{20} k} + e^{+j \frac{2\pi}{20} k} \right) \right) \cdot \\
 &\left(\frac{1}{20} \left(1 + e^{-j \frac{2\pi}{20} k} + e^{+j \frac{2\pi}{20} k} \right) \right) \\
 &= \frac{1}{20^2} \left(3 + 2 e^{-j \frac{2\pi}{20} k} + e^{-j \frac{4\pi}{20} k} \right. \\
 &\quad \left. + e^{+j \frac{4\pi}{20} k} + 2 e^{+j \frac{2\pi}{20} k} \right) \\
 &= \frac{1}{(20)^2} \left(3 + 4 \cos\left(\frac{2\pi}{20} k\right) + 2 \cos\left(\frac{4\pi}{20} k\right) \right)
 \end{aligned}$$

Square pulse case

2+

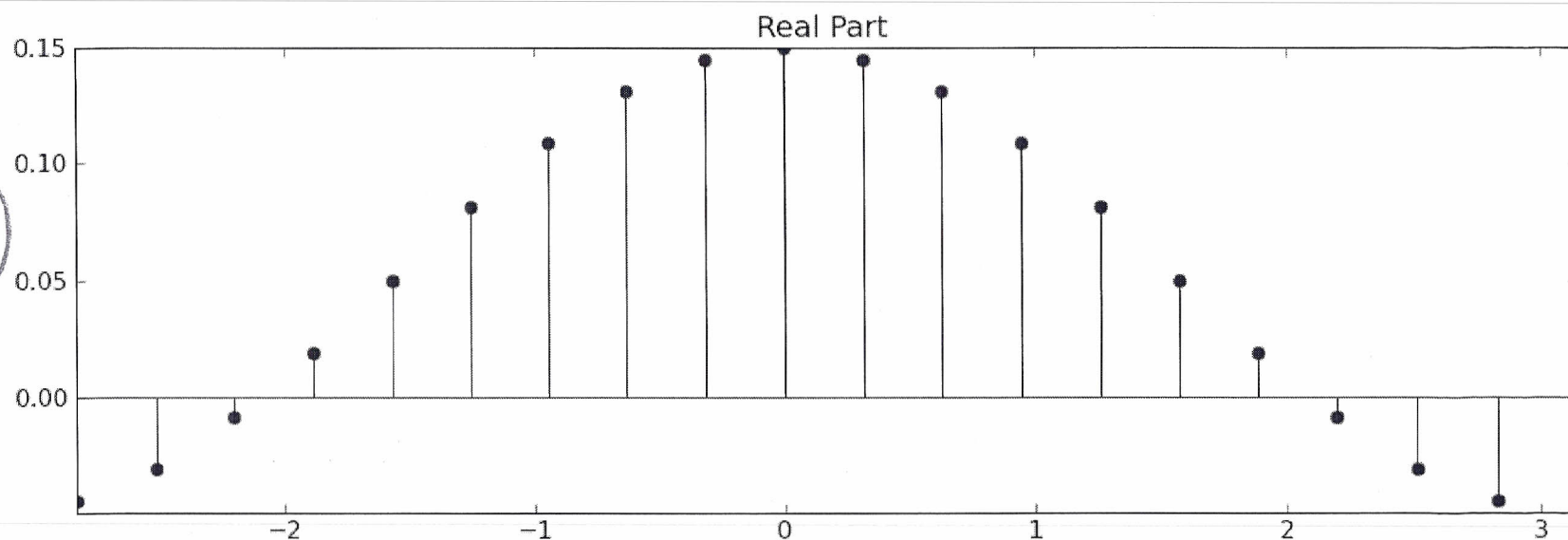
$x[n]$



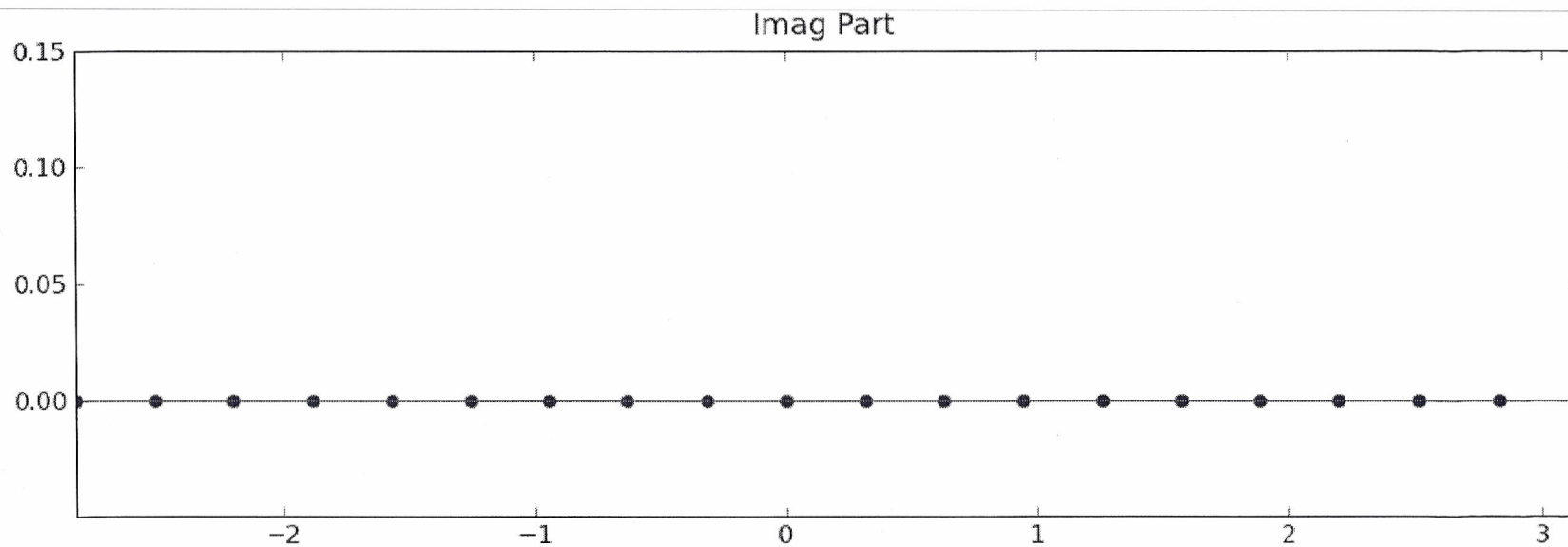
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Square Pulse DTFS Coefficients

$\text{Re}\{X[k]\}$

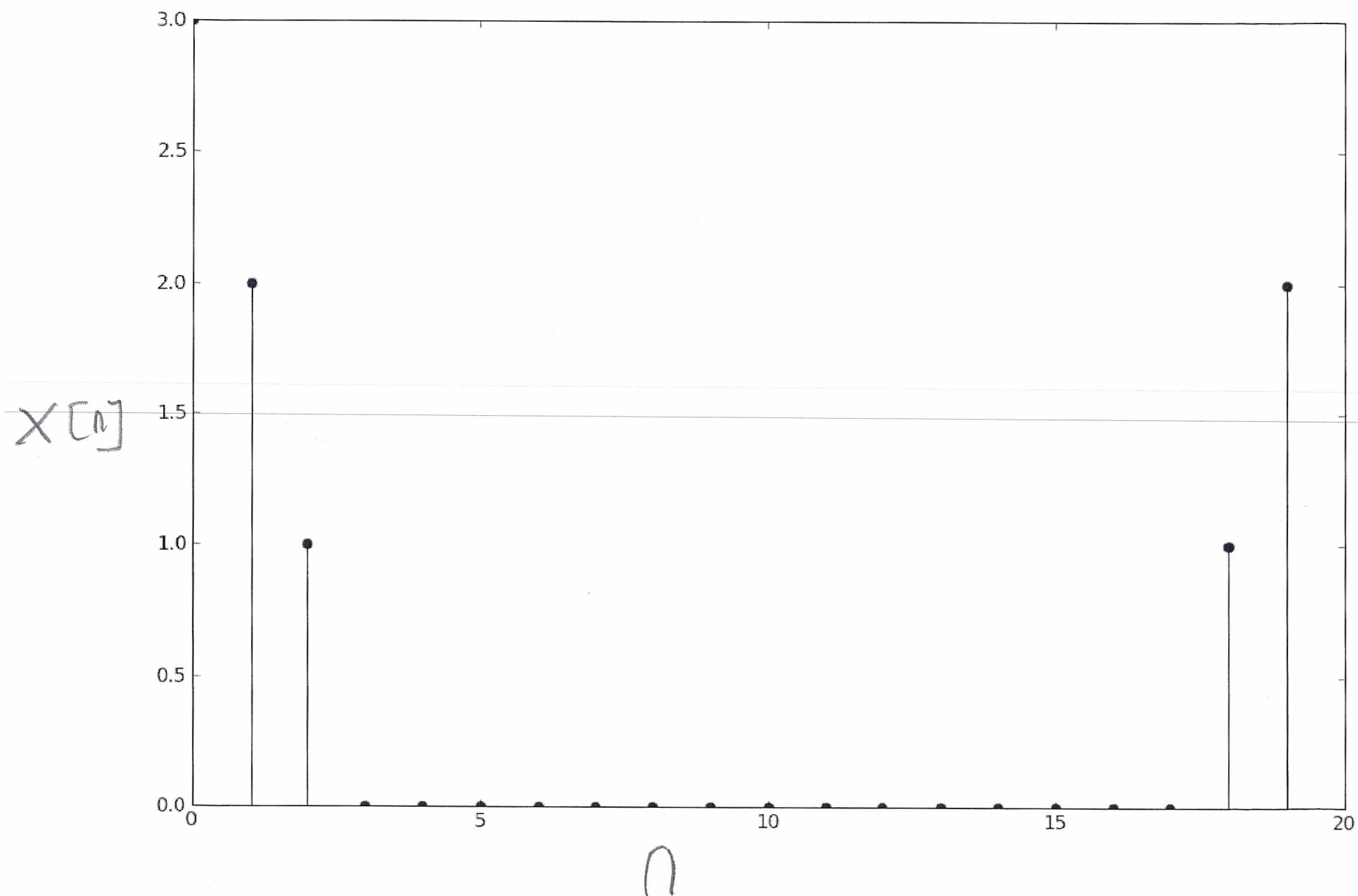


$\text{Im}\{X[k]\}$



28

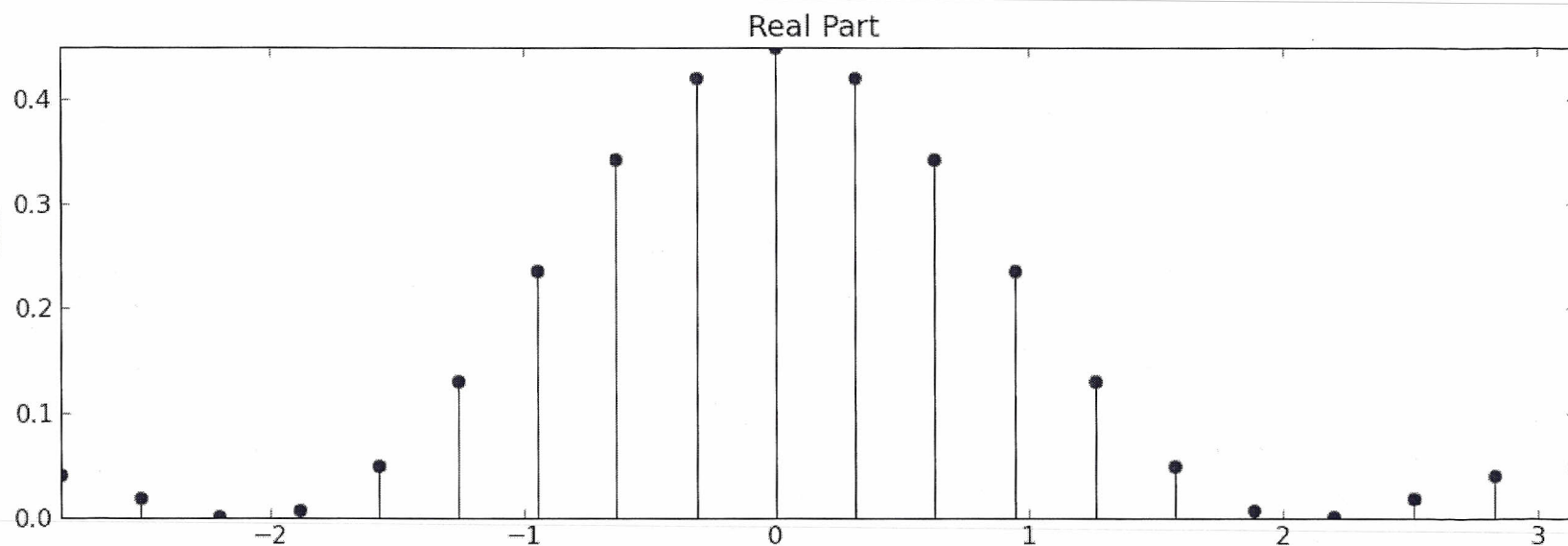
Triangular pulse Case



Triangular Pulse DTFS coefficients

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$\text{Re}(X[k])$



$\text{Im}(X[k])$

