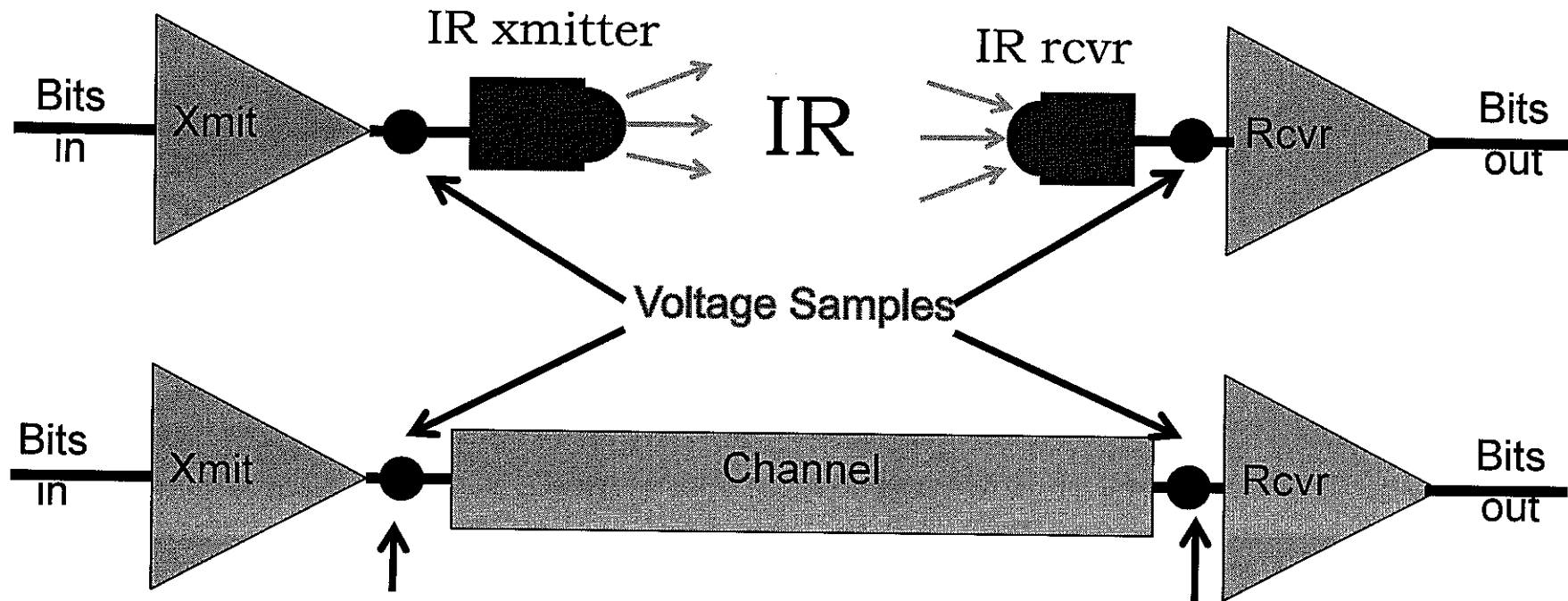




2

Bits  $\rightarrow$  Samples  $\{X\} \rightarrow$  Channel  $\rightarrow$  Samples  $\{Y\} \rightarrow$  Bits



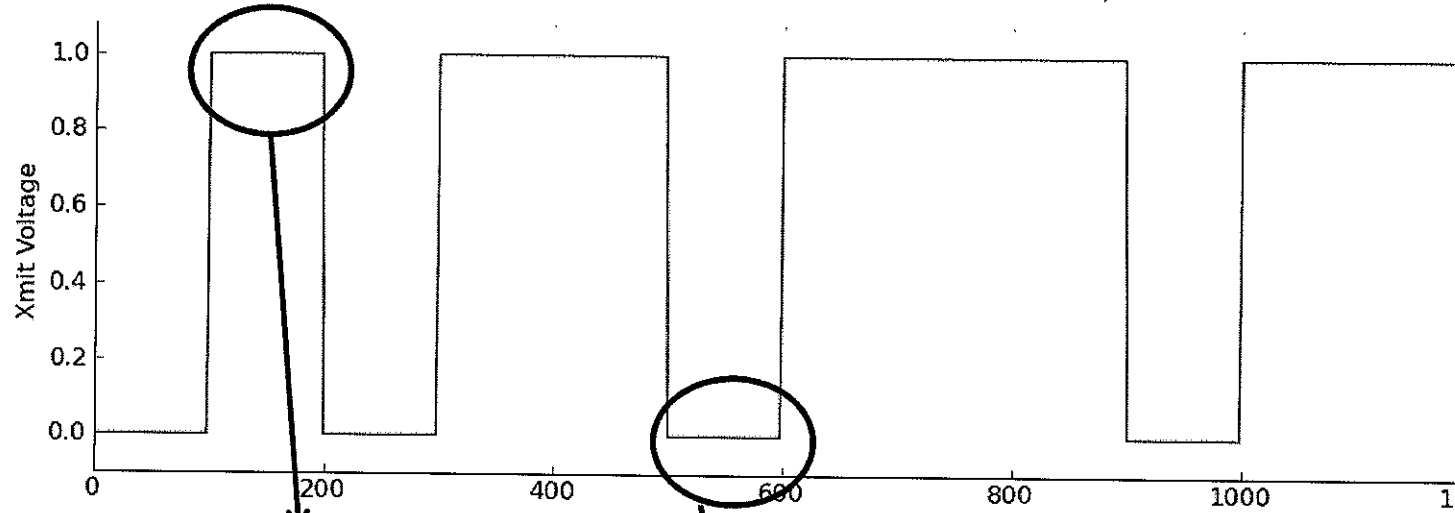
$X \equiv$  entire sequence  
 $x[n] \equiv n^{th}$  sample value

$Y \equiv$  entire sequence  
 $y[n] \equiv n^{th}$  sample value

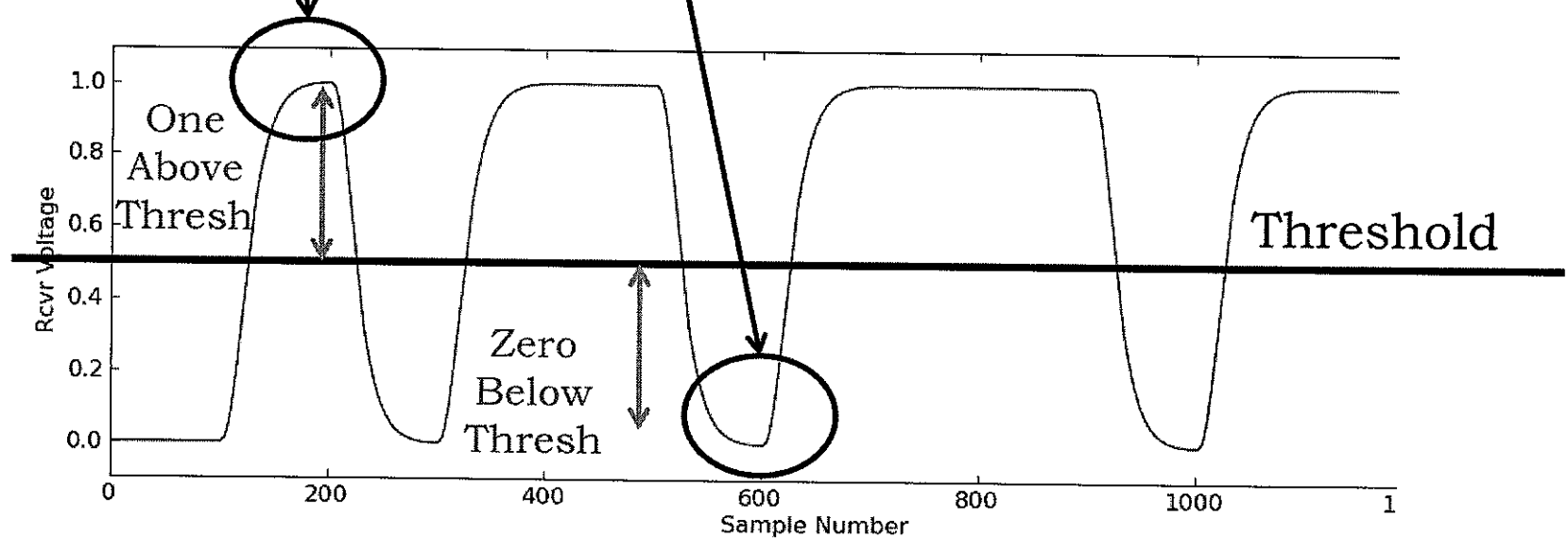
Sample Rates:   
  $\rightarrow$  4 million Samples/Second (IR Transceiver)  
  $\rightarrow$  Up to gigaSamples/Second (Fastest)

# Excellent Channel Transmission (using 100 Samples/bit)

$x[n]$

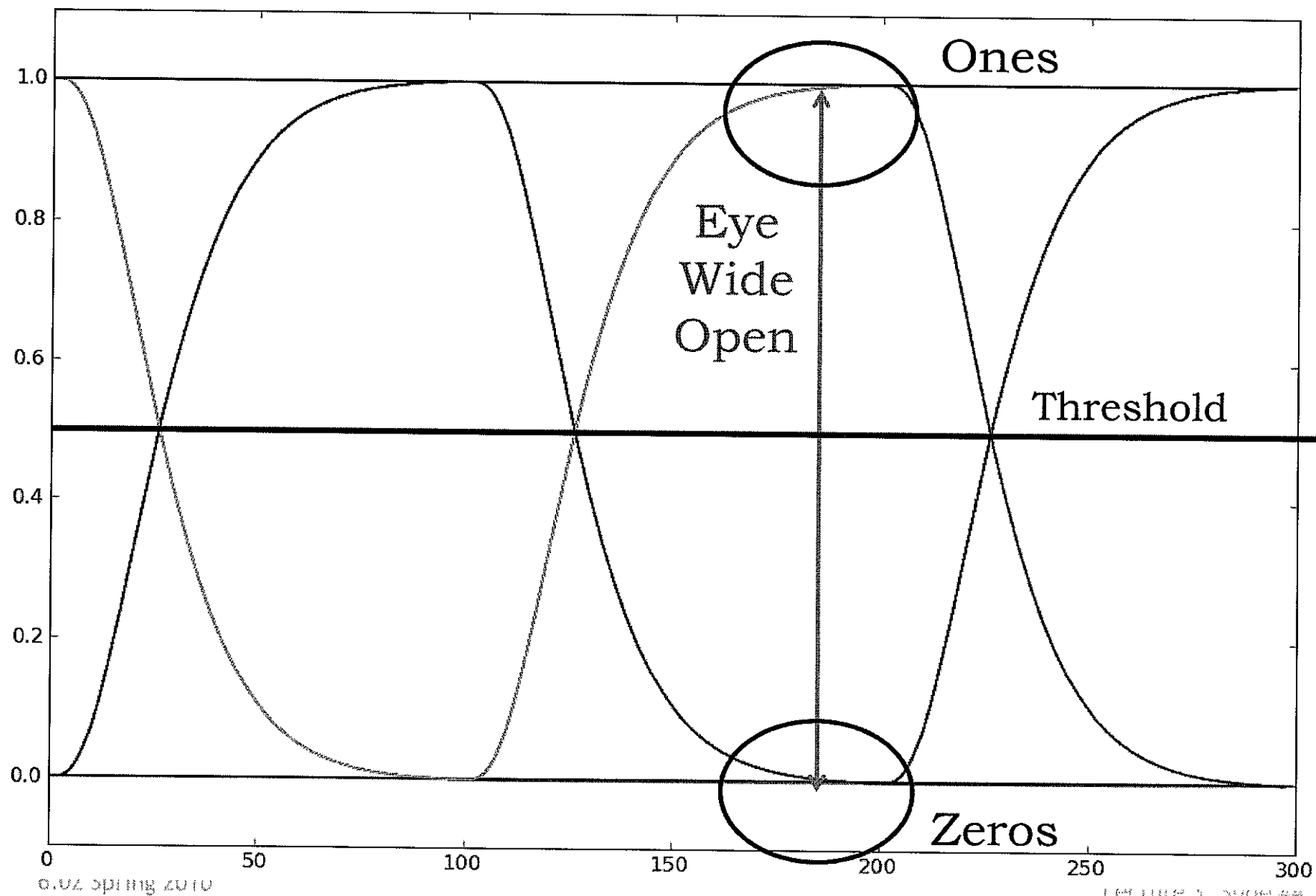


$y[n]$

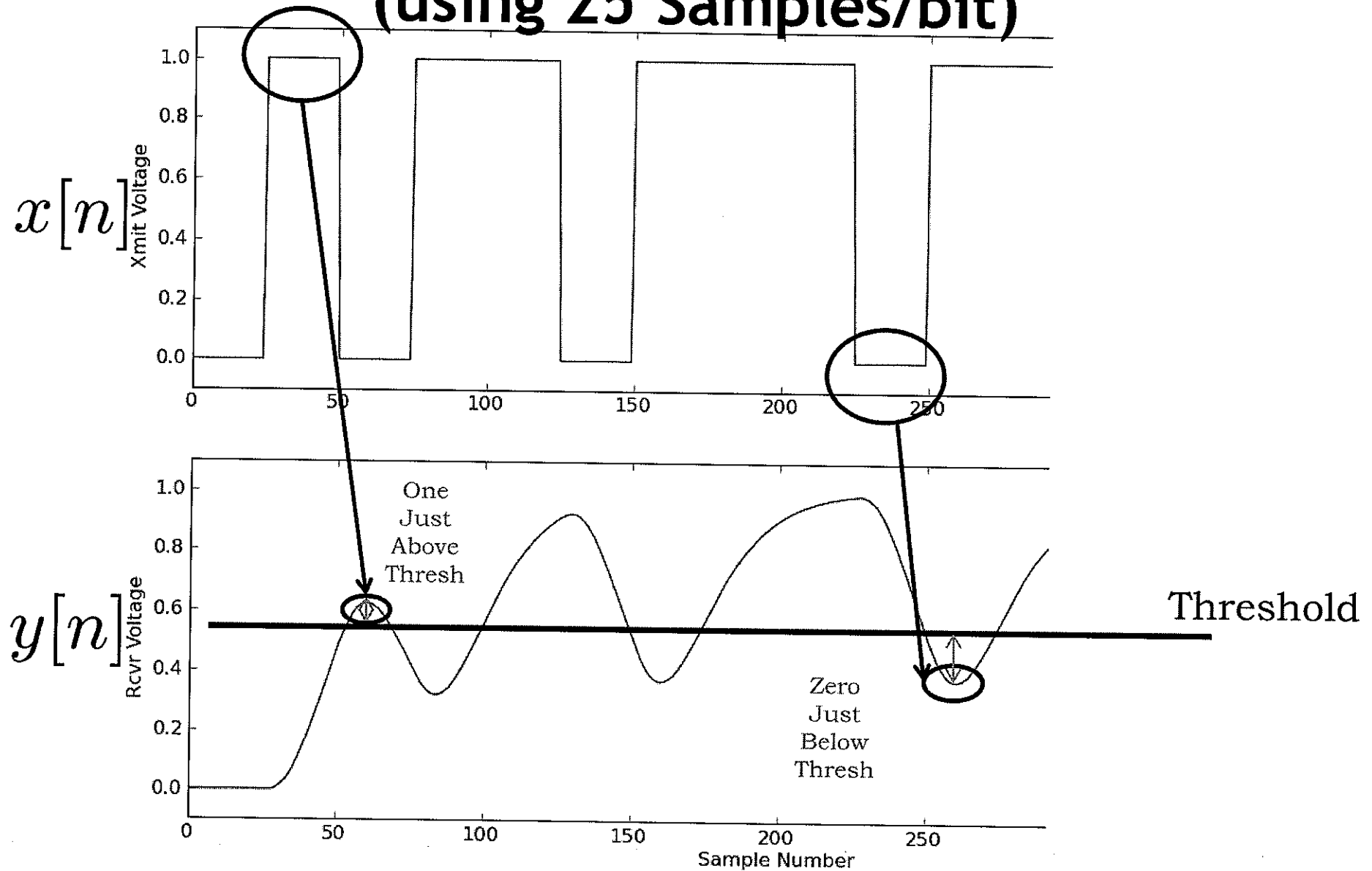


4

# Excellent Channel Transmission Eye Diagram

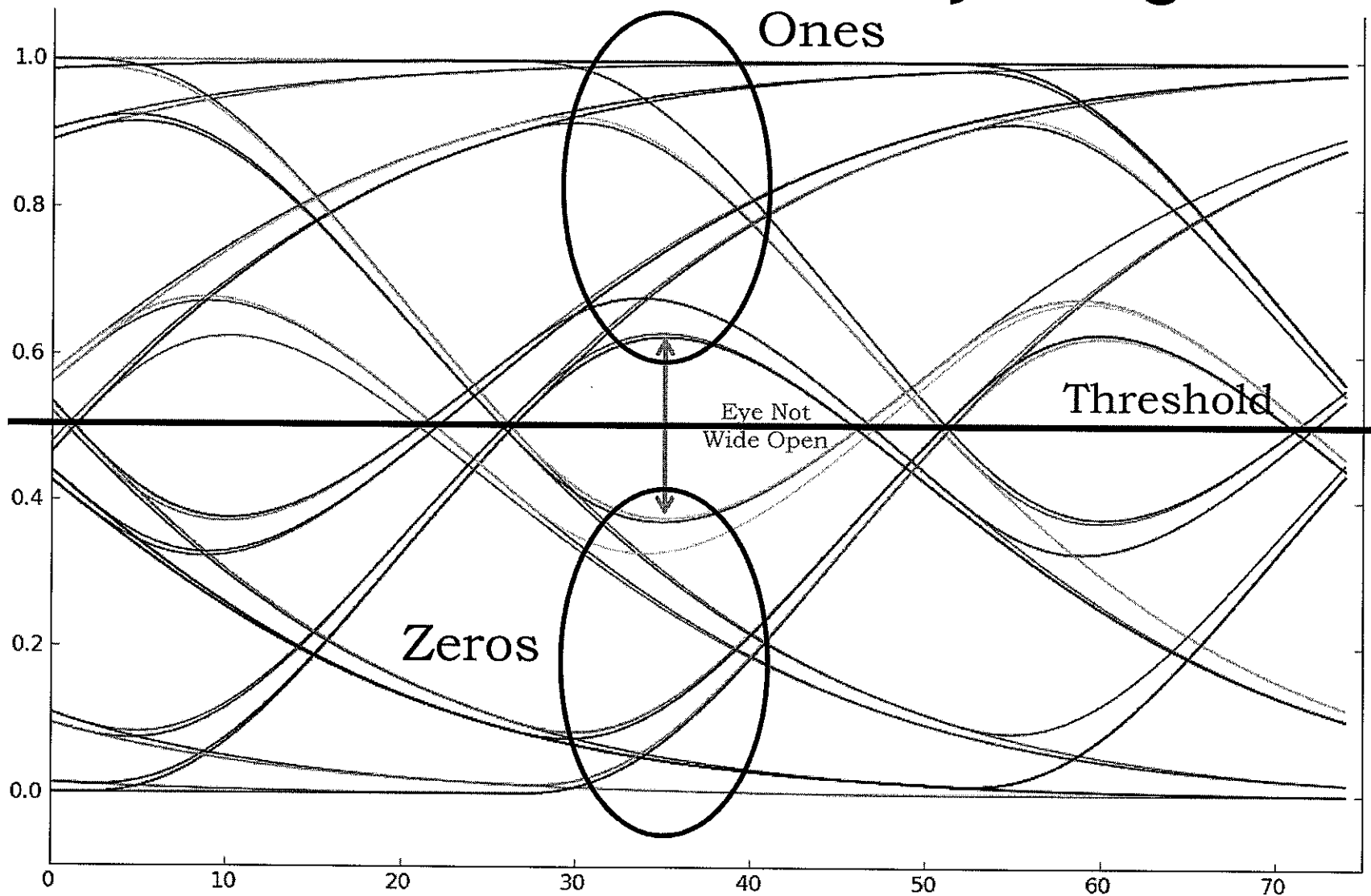


# Okay Channel Transmission (using 25 Samples/bit)



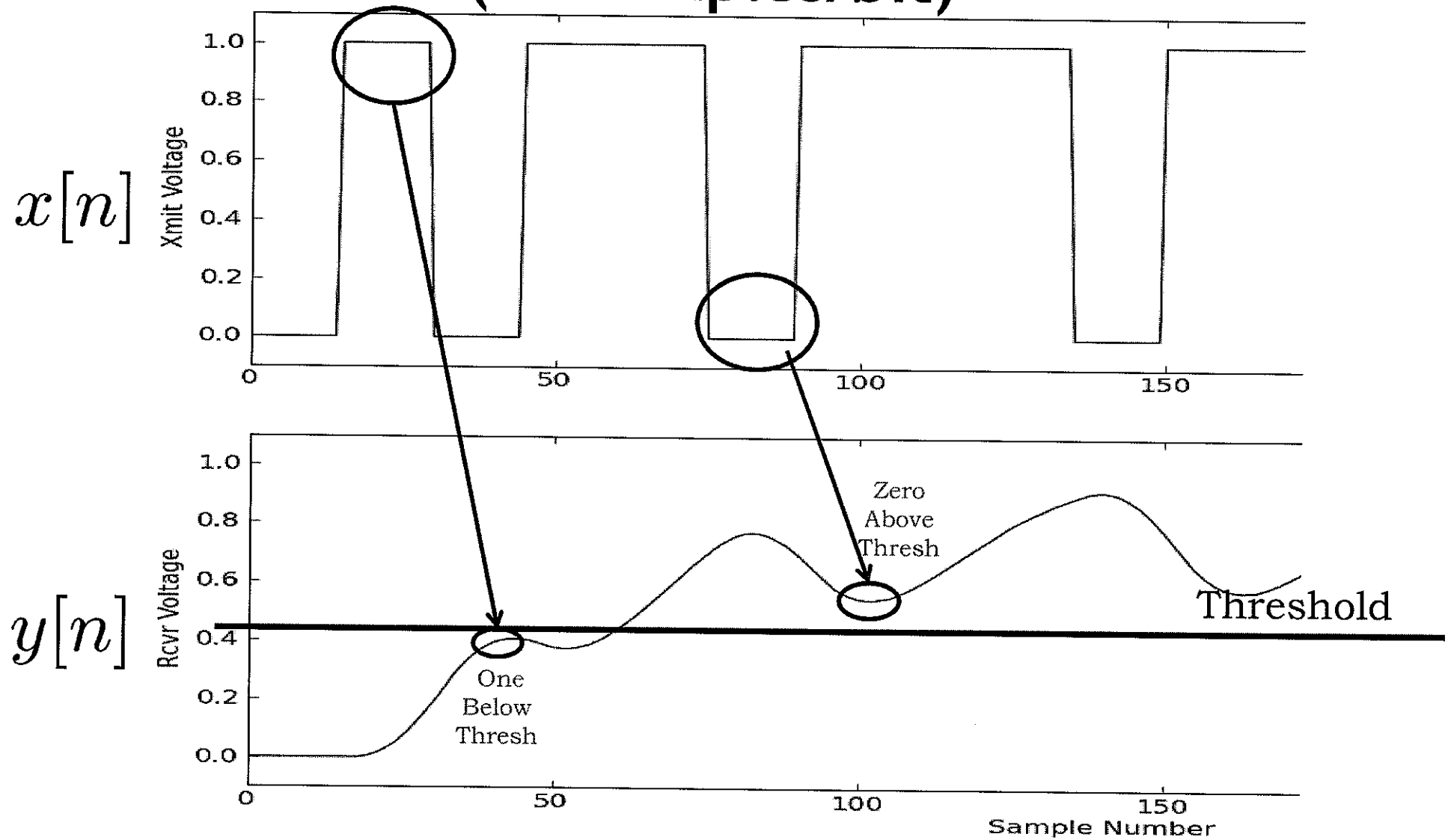
# Okay Channel Transmission Eye Diagram

6



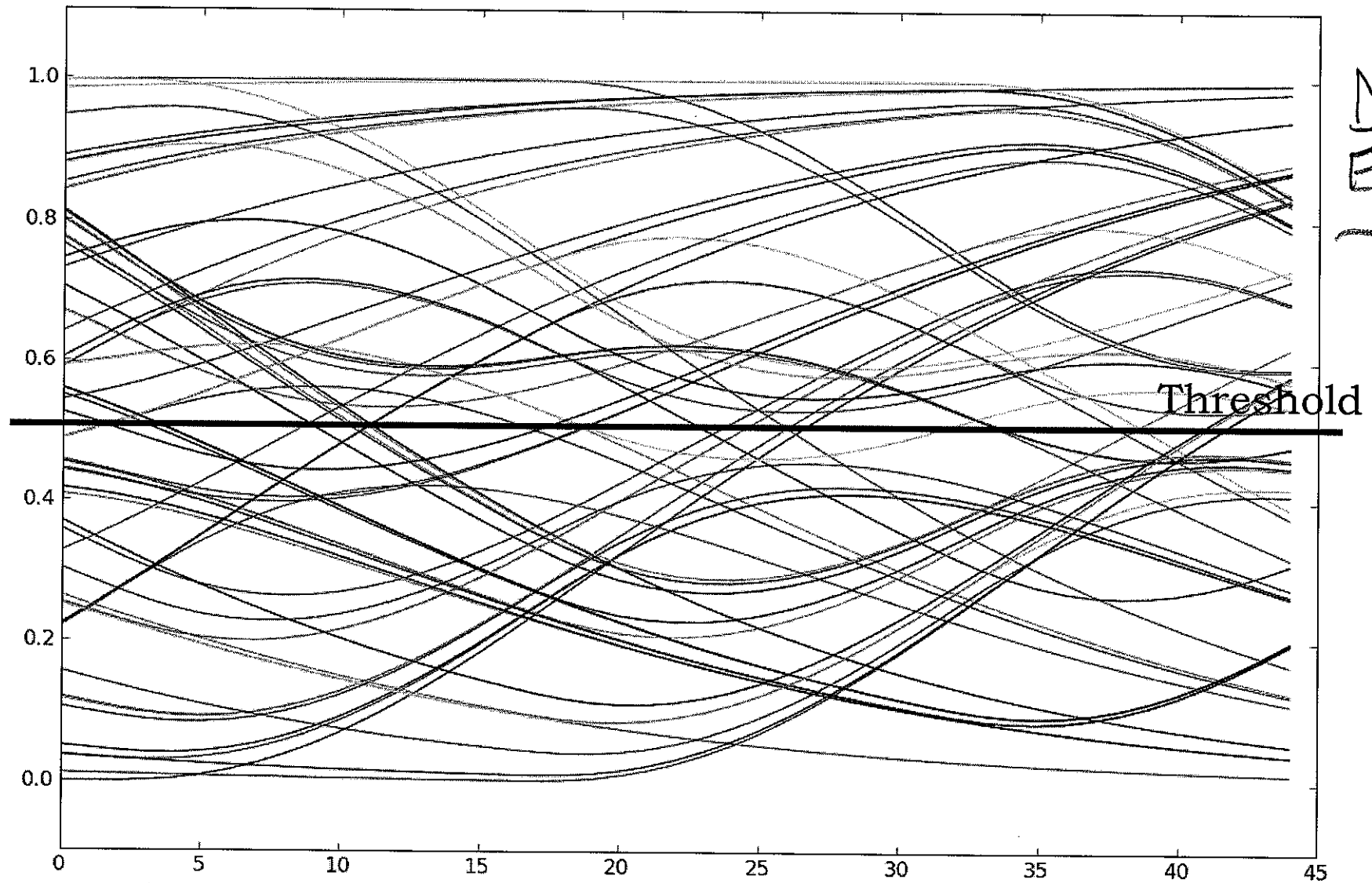
7

# Unusable Channel Transmission (15 Samples/bit)



8

# Unusable Channel Transmission Eye Diagram



No  
Eye!

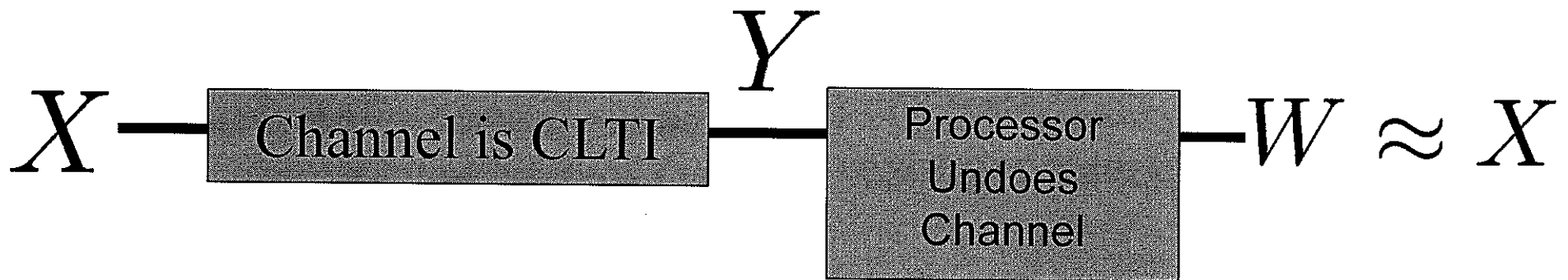
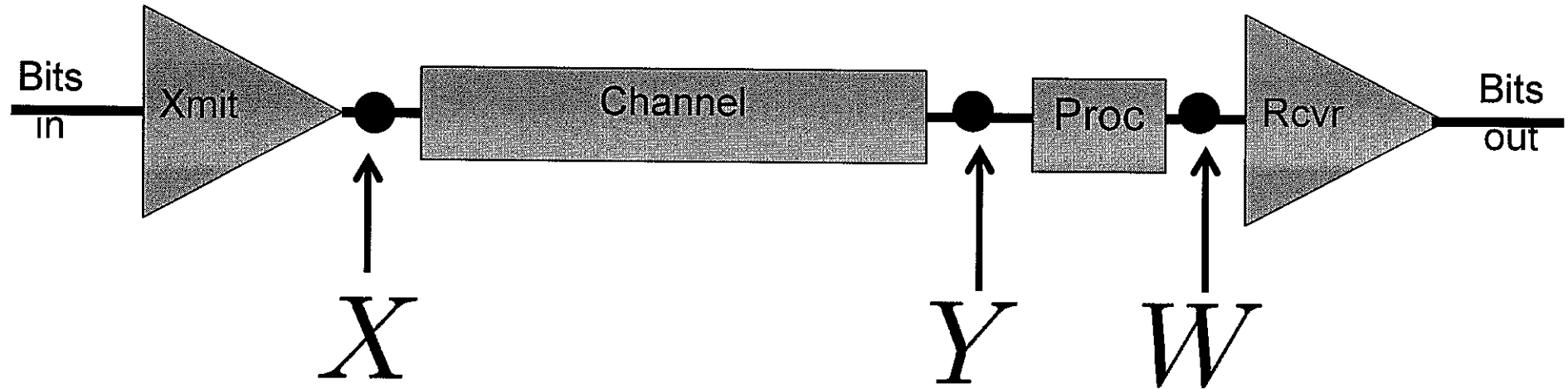


# What does an Eye Diagram Tell You?

9

- How many samples to use per bit
  - More samples/bit:
    - More open eye → Easier detection ↑
    - Fewer bits/second → Slower bit rate ↓
- Where to put the 1-0 threshold
  - The threshold should evenly divide the upper and lower parts of the eye.
- How to estimate errors
  - Next Week.

# Faster Bit Rate with Post-Channel Processing?



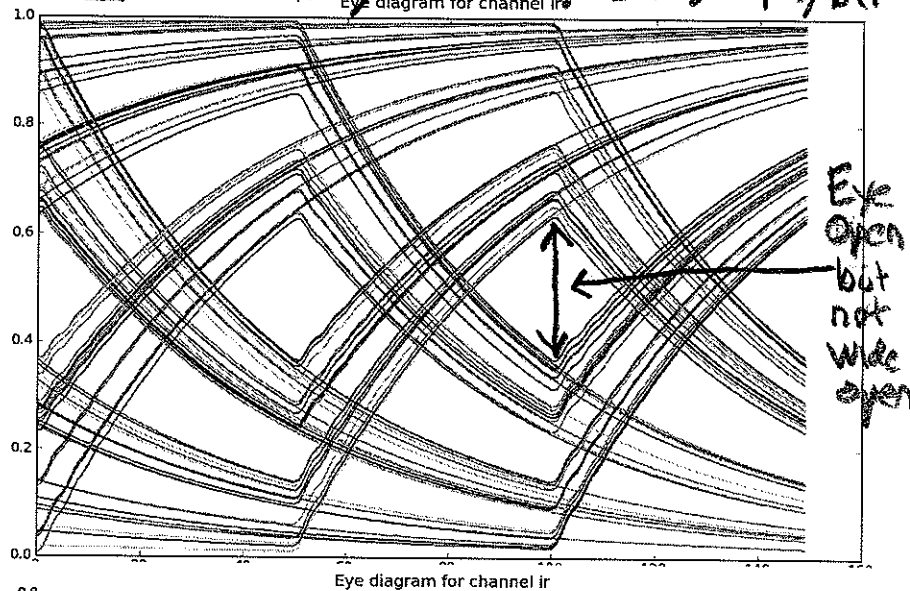
- Model Wire as Causal, Linear Time-Invariant (CLTI)
- Use Unit Sample Response to Undo Channel Effects

If  $W = X$  then only  
need 1 sample/bit  
(Perfect deconvolution)

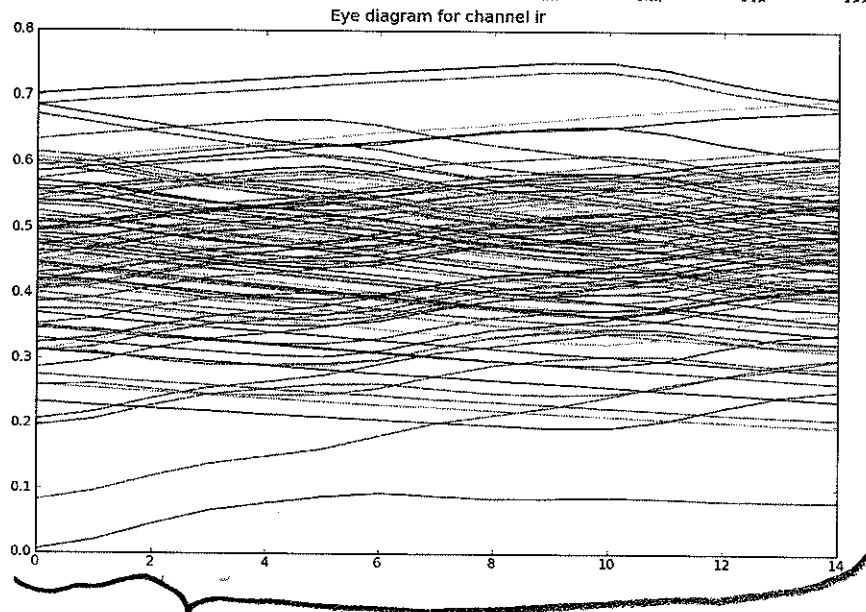
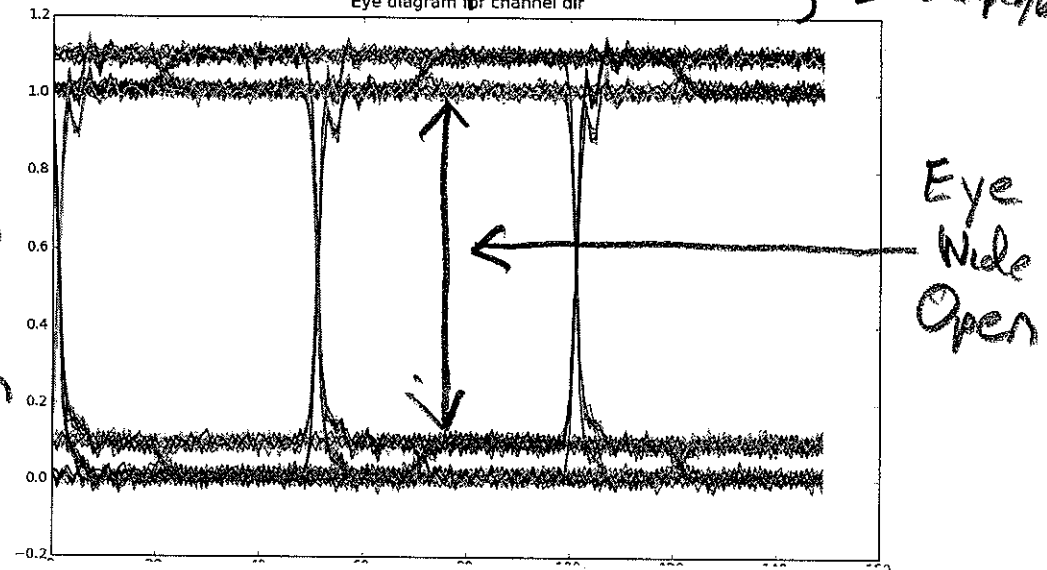
10A

# A Glimpse at IR Channel Results

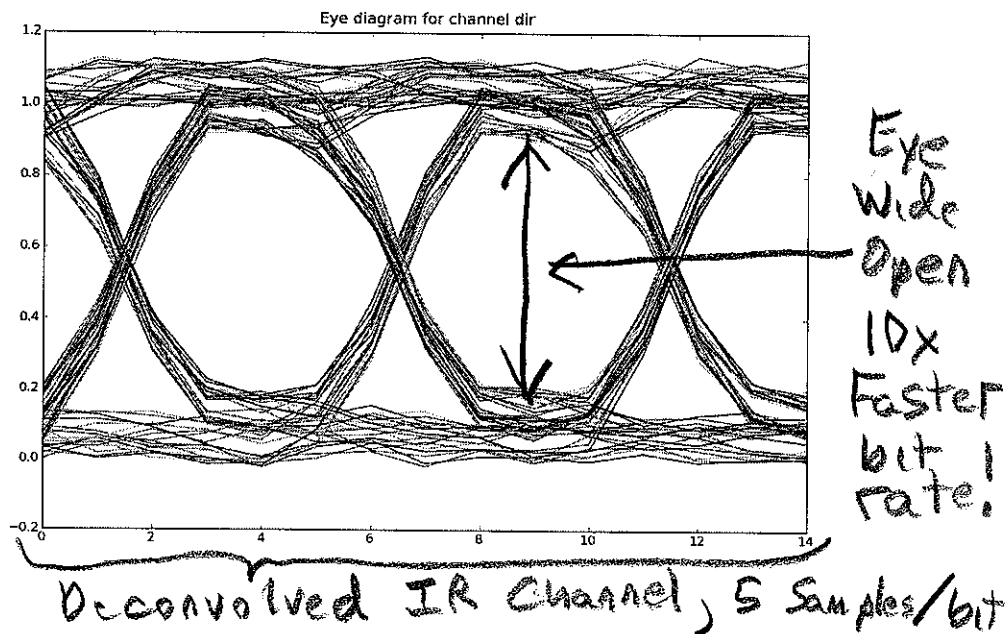
IR Channel, Light On, 50 samples/bit



IR Channel plus Deconvolver, 50 samples/bit



IR Channel, 5 samples/bit



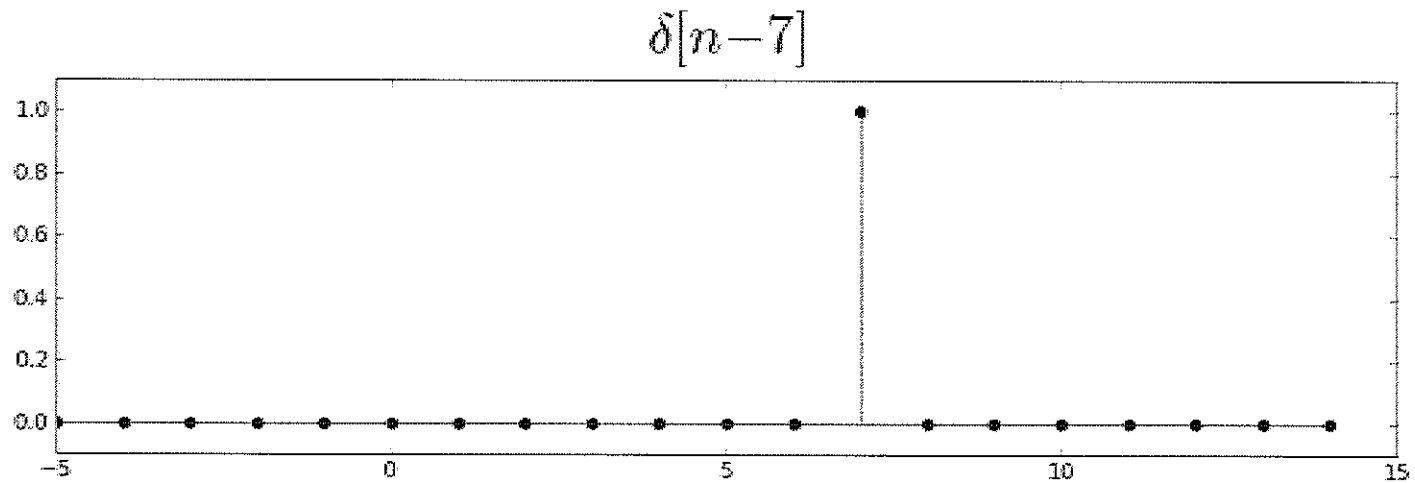
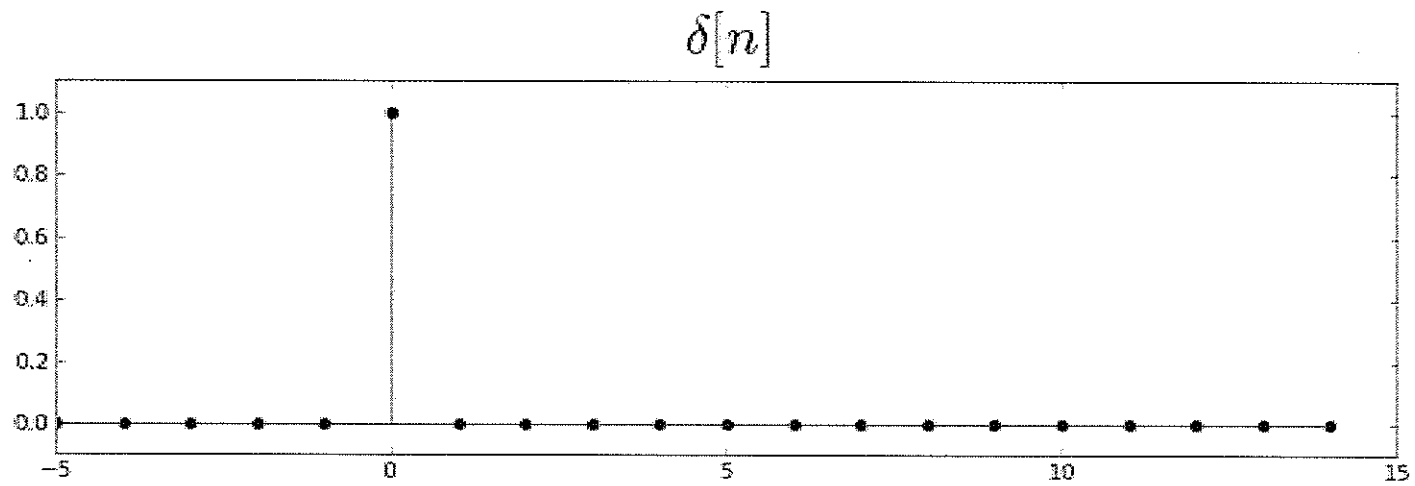
# Causality, Time-Invariance and Linearity



- Causality (out does not change before in)
  - If  $x[n] = x[-\infty]$  for  $n < N$ 
    - Then  $y[n] = y[-\infty]$  for  $n < N$ .
- Time-Invariance (out shifts when in shifts)
  - If  $in = x[n] \rightarrow out = y[n]$ 
    - Then  $in = x[n - k] \rightarrow out = y[n - k]$
- Linearity
  - If  $in = x_a[n] \rightarrow out = y_a[n], in = x_b[n] \rightarrow out = y_b[n]$ 
    - Then  $in = Ax_a[n] + Bx_b[n] \rightarrow out = Ay_a[n] + By_b[n]$

12

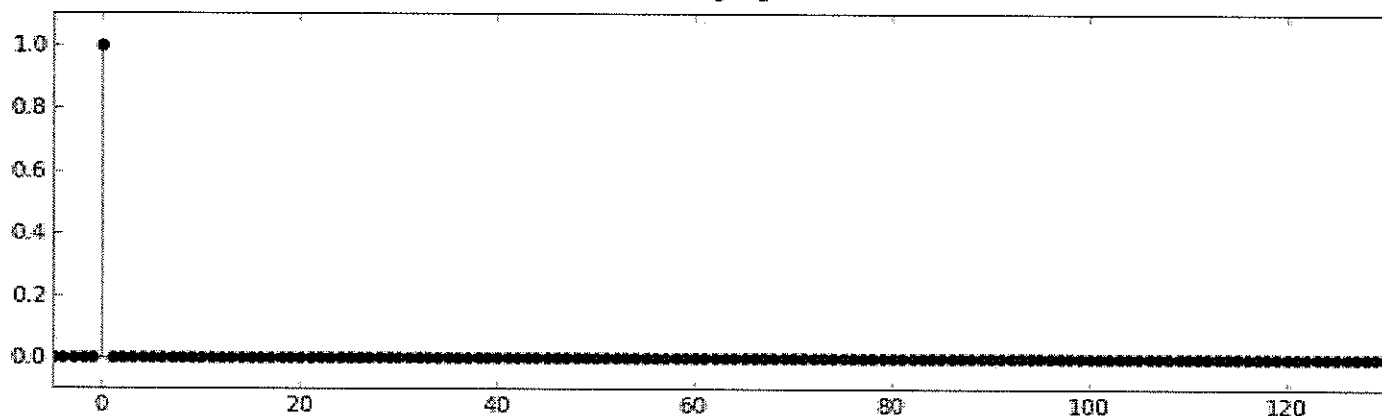
# Unit Sample Reminder



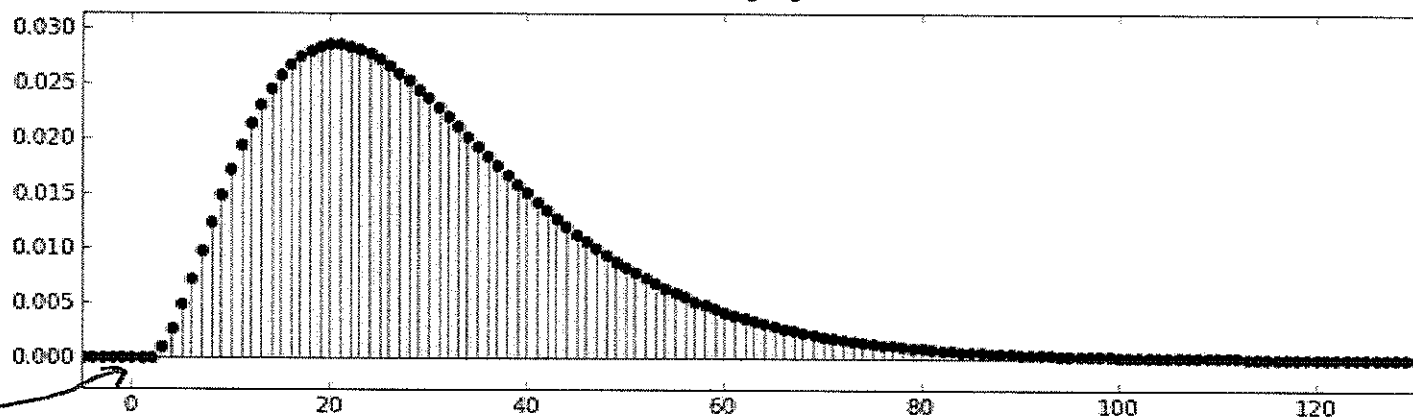
# Unit Sample Response (slow wire)

13

$\delta[n]$

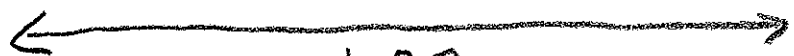


$h[n]$



Notice

$h[0]=0$   
 $h[1]=0$   
 $h[2]=0$



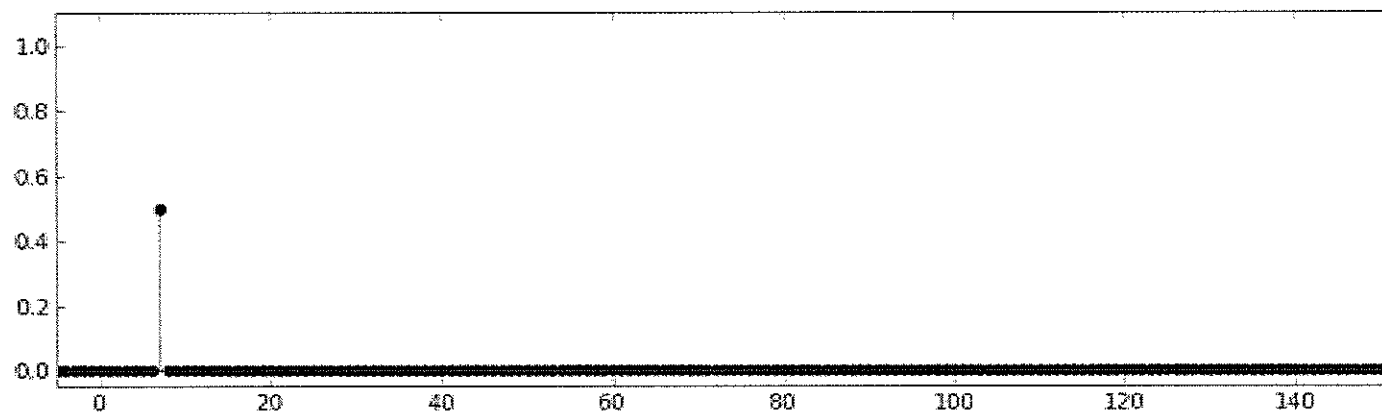
Notice  $h[n]$  is non zero  
for  $\sim 80$  samples (ISI  
for 40 samples/bit was two  
bits)

(That is, when  $h[n]$  is non zero  
for 2 bit periods, then  
two previous bits "interfere"  
with current bit)

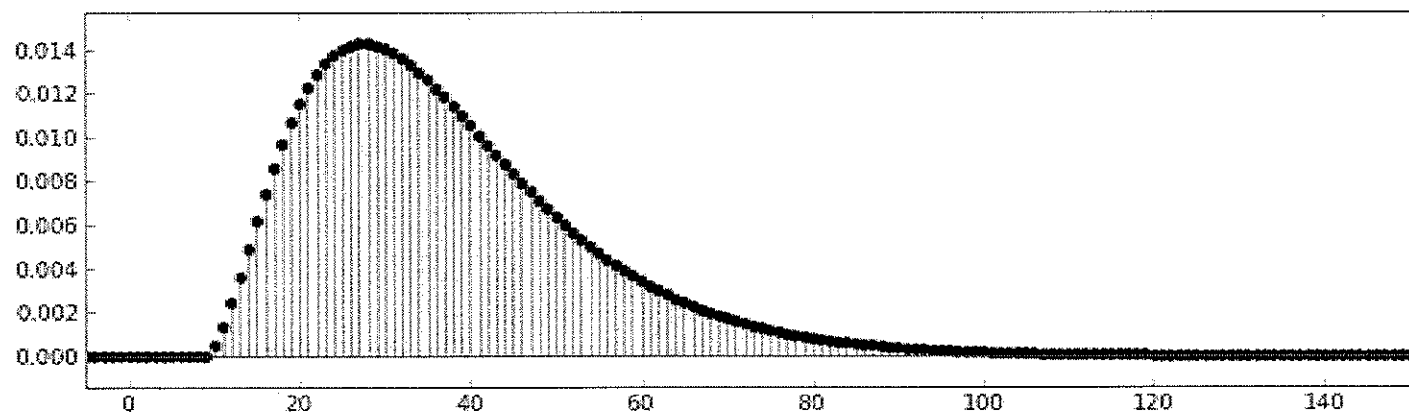
# Shifted and Scaled Unit Sample Response

14

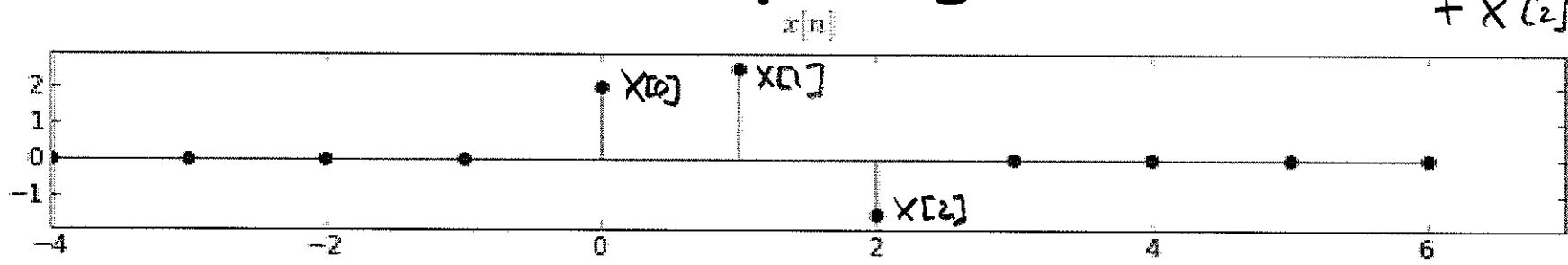
$$0.5\delta[n-7]$$



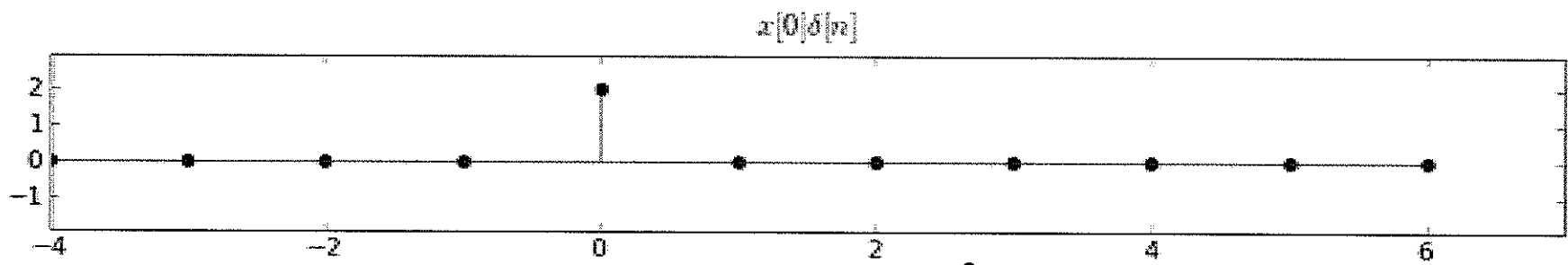
$$0.5h[n-7]$$



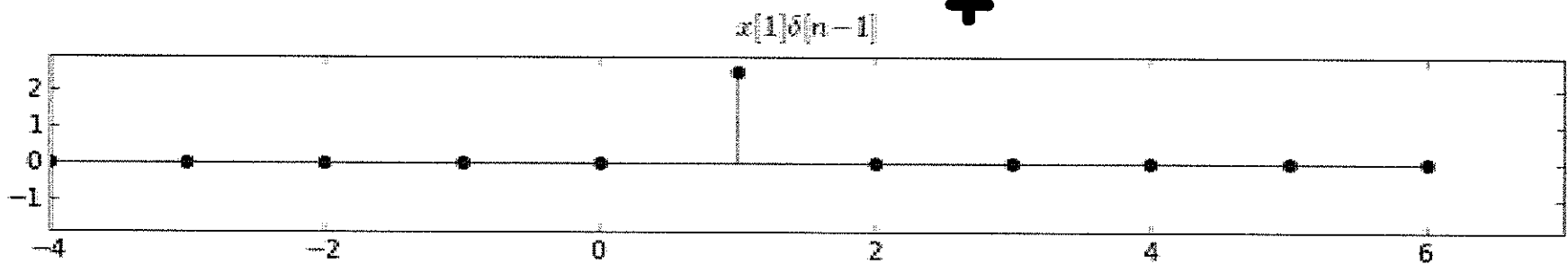
# Decomposing $X[n] = X[0]\delta[n] + X[1]\delta[n-1] + X[2]\delta[n-2]$



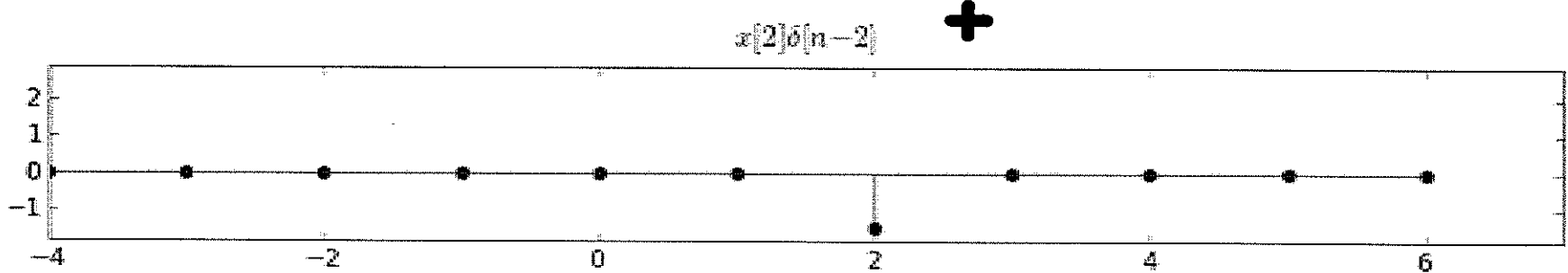
(Note:  
 $x[n]=0$   
 $n > 2$   
 $n < 0$ )



+



+

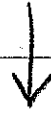
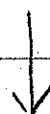
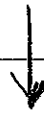




16

In General for LTI

$$X[n] = X[0]\delta[n] + X[1]\delta[n-1] + X[2]\delta[n-2] + \dots$$



$$y[n] = X[0]h[n] + X[1]h[n-1] + X[2]h[n-2] + \dots$$

In summation form:

$$y[n] = \sum_{m=0}^n h[n-m]X[m]$$

Because  $h[n-m] = 0$  for  $n-m < 0$

Because  $X[m] = 0$  for  $m < 0$

Notation:

$$\text{sequences} \rightarrow Y = H * X$$

Convolution symbol

Causality

Unit step response can not be nonzero before the unit step occurs.  
 $h[n] = 0$  for  $n < 0$

Side Note Polynomial multiplication

Looks like convolution

$$(a[0] + a[1]z + a[2]z^2 + \dots + a[n]z^n)(b[0] + b[1]z + \dots + b[m]z^m)$$

$$= (a[0]b[0] + (a[0]b[1] + a[1]b[0])z + \dots + a[n]b[m]z^{n+m})$$

Alternative Form

$$y[n] = \sum_{m=0}^n x[n-m] h[m]$$

$$y[n] = x[n] h[0] + x[n-1] h[1] + x[n-2] h[2] + \dots$$

Suppose  $h[n] = 0, n > K$

$$y[n] = h[0] x[n] + h[1] x[n-1] + \dots + h[K] x[n-K]$$

(If  $n < K$ , equation still holds as  $x[n-K] = 0$  because  $x[n] = 0, n < 0$ )

Difference Equation

\*\*\*\*\* Aside for the curious  
Compare to 6.9 Right-shift operator  $\mathcal{R}$

$$v = \mathcal{R} x \Rightarrow v[n] = x[n-1]$$

$$v = \mathcal{R}^2 x \Rightarrow v[n] = x[n-2]$$

$$\Rightarrow y = (h[0] + h[1] \mathcal{R} + \dots + h[K] \mathcal{R}^K) x$$

$$= \left( \sum_{i=0}^K h[i] \mathcal{R}^i \right) x$$

A  
S  
I  
D  
E

## Deriving Alternate Form

17A

We have

$$y[n] = \sum_{m=0}^{m=n} h[n-m] x[m]$$

Introduce an intermediate variable

$$l \equiv n-m \Rightarrow m = n-l$$

Therefore

Rewriting

$$\begin{aligned} y[n] &= \sum_{(n-l)=n}^{(n-l)=0} h[l] x[n-l] \\ &= \sum_{l=0}^{l=n} h[l] x[n-l] \end{aligned}$$

Since  $m$  and  $l$  are just used for indexing, relabel  $l$  with  $m$

$$y[n] = \sum_{m=0}^{m=n} x[n-m] h[m]$$

Alternate Form

ASIDE

$$Y = \left( \sum_{i=0}^K h[i] x^i \right) X$$

$W$  = reconstructed  $X$

How to derive  $W$  from  $Y$ :

$$W = \left( \frac{1}{\left( \sum_{i=0}^K h[i] x^i \right)} \right) Y$$

or

$$\left( \sum_{i=0}^K h[i] x^i \right) W = Y$$

Deconvolution  $X \rightarrow \boxed{\text{chame}} \rightarrow Y \rightarrow \boxed{\text{Decon}} \rightarrow W$

Why?

$$\left\{ \begin{array}{l} h[0] W[n] + h[1] W[n-1] + \dots + h[K] W[n-K] \\ = Y[n] \end{array} \right.$$

Answer Since  $W$  reconstructs  $X$ :  $W \approx X$

It should satisfy the same difference equation as  $X$

or

See More General Derivation in Aside.

(19)

Solving for WPlug & Chug

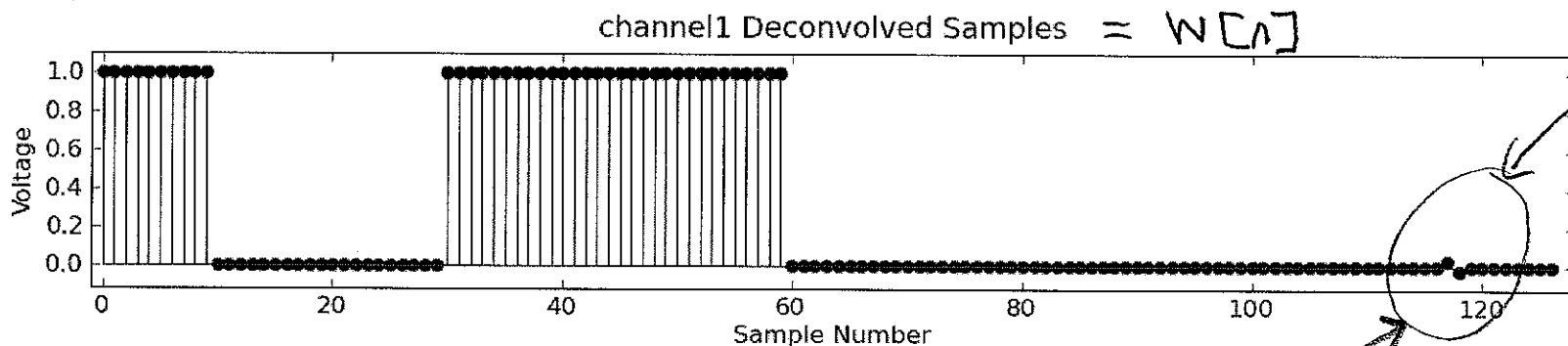
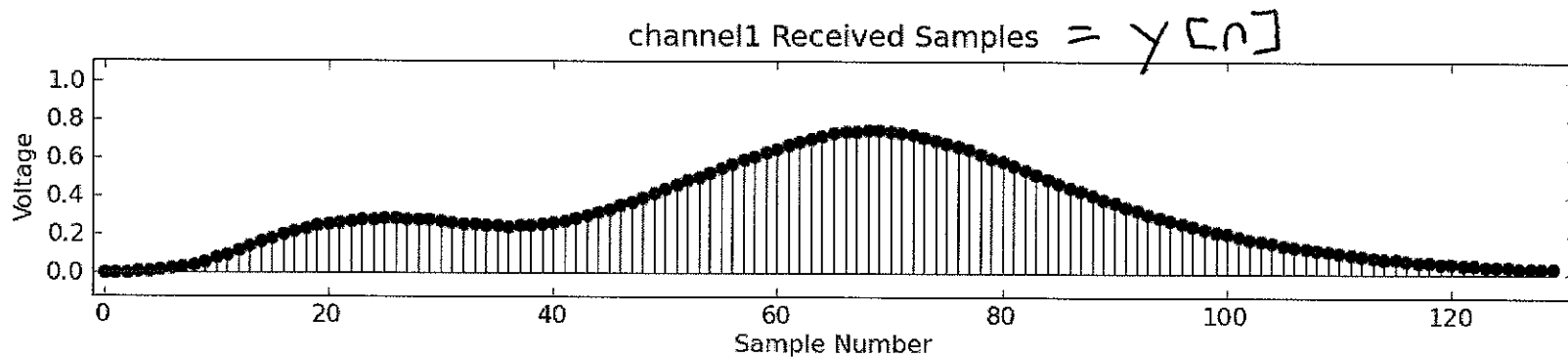
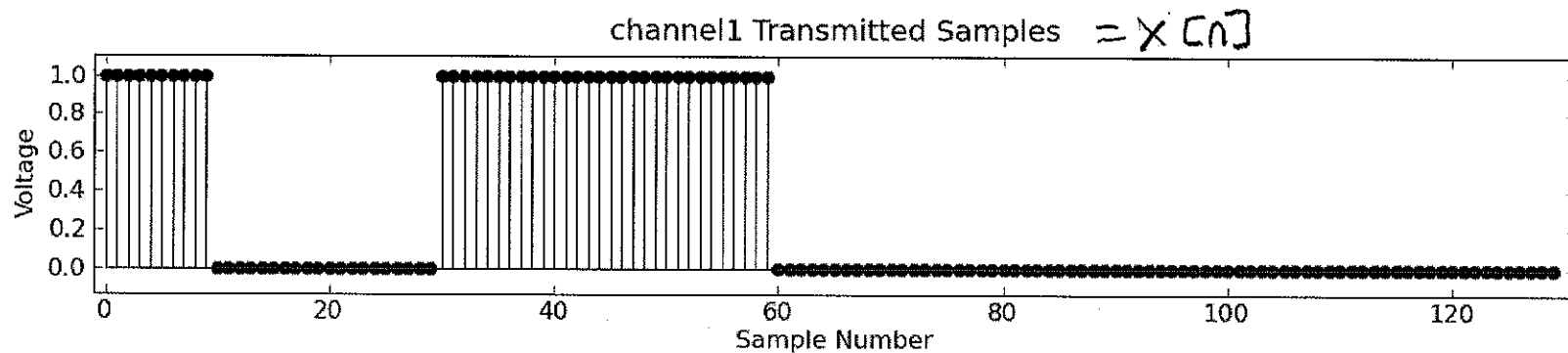
$$W[0] = \frac{1}{h[0]} \left( Y[0] - \underbrace{(h[1]W[-1] + \dots + h[R]W[R])}_{\text{all zero as } W[-1] = 0, \dots, W[R] = 0} \right)$$

$$W[1] = \frac{1}{h[0]} (Y[1] - h[1]W[0])$$

$$W[2] = \frac{1}{h[0]} (Y[2] - (h[1]W[1] + h[2]W[0]))$$

$$\vdots$$

# Noise-Free Deconvolution Result ( $h[n] = 0, n \geq 120$ )



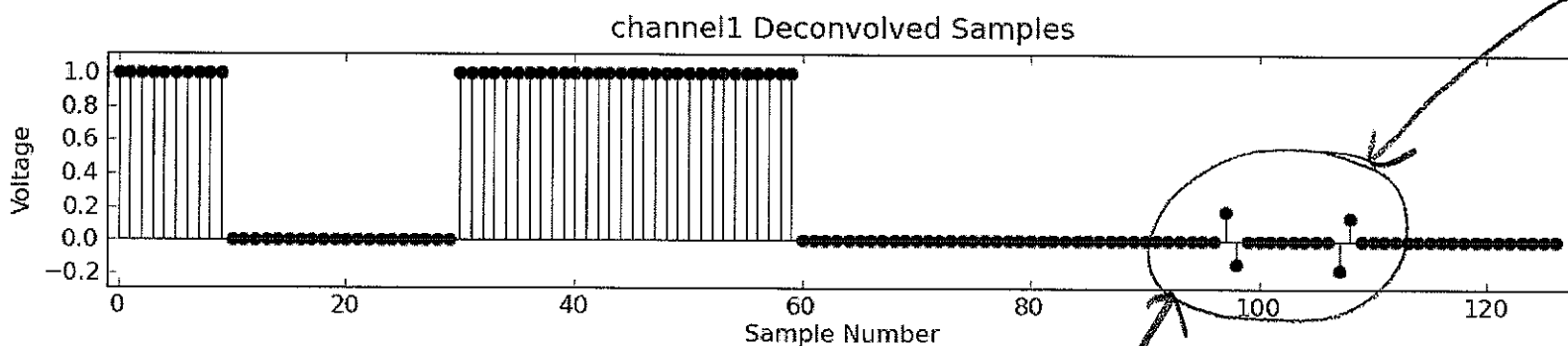
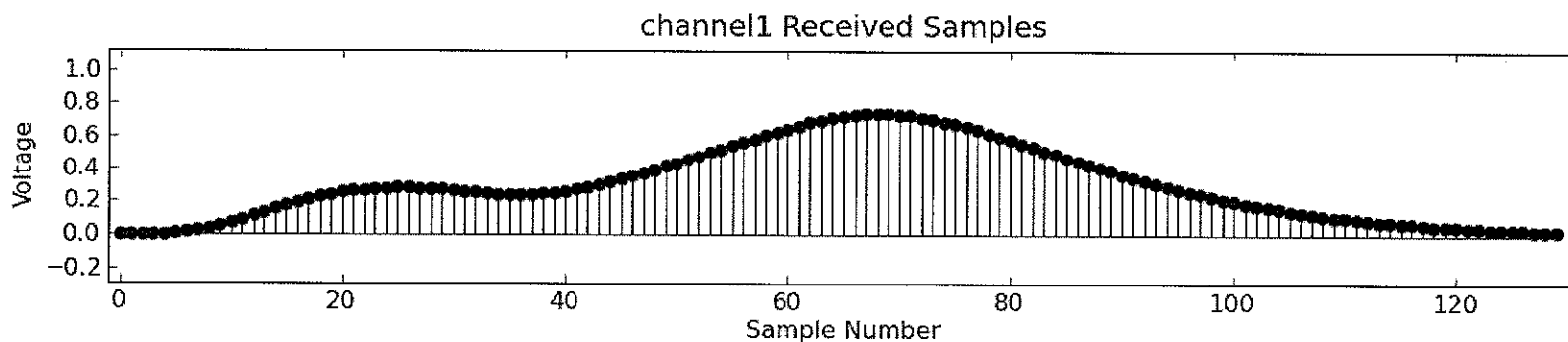
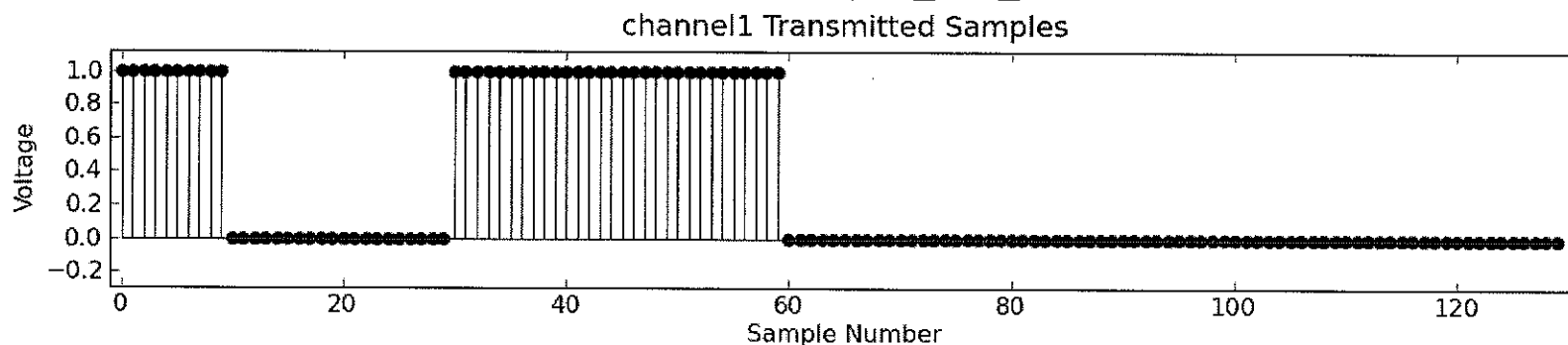
Truncated  
unit  
sample  
response  
 $\Rightarrow W \approx X$   
Imperfect  
Deconvolution  
Because  
channel  
model  
Inexact!

Notice  
small  
error

Error appears near  
sample 120

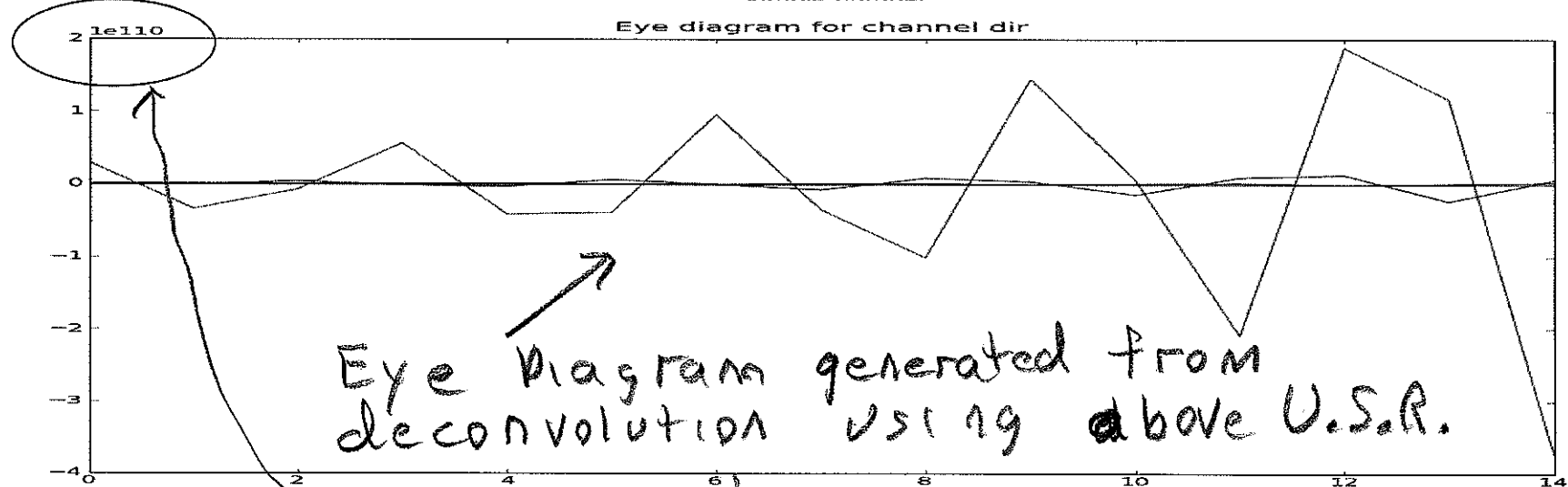
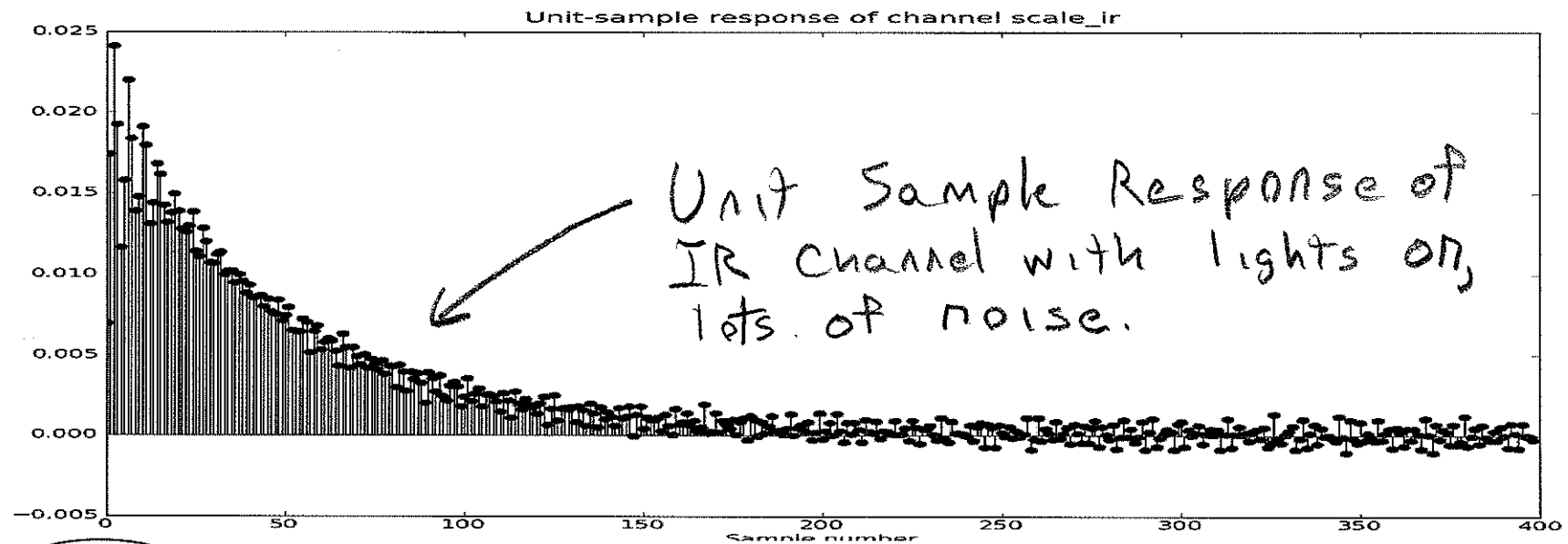
(21)

# Noise-Free Deconvolution Result ( $h[n] = 0, n \geq 100$ )



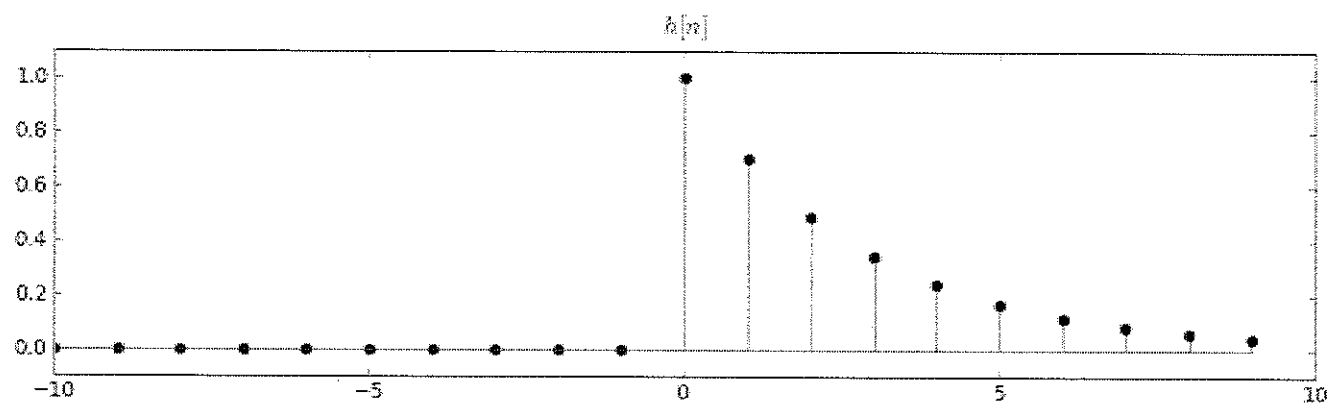
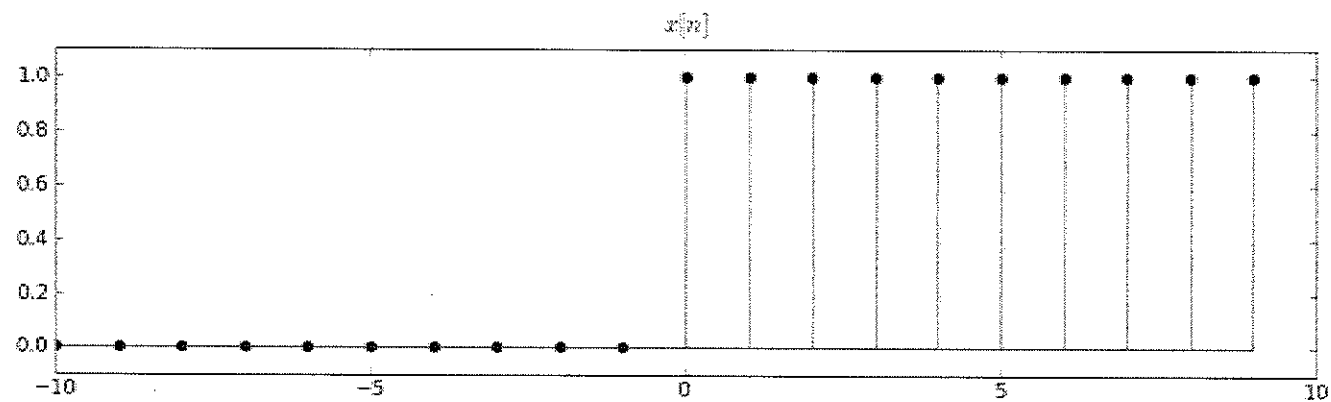
Z1A

# How bad is noise for deconvolution?



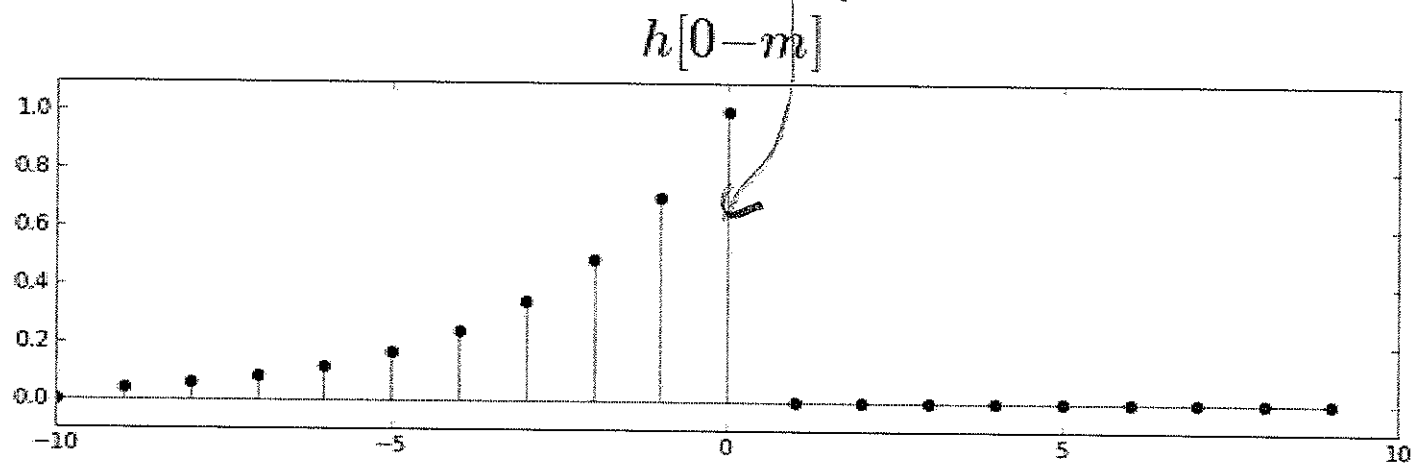
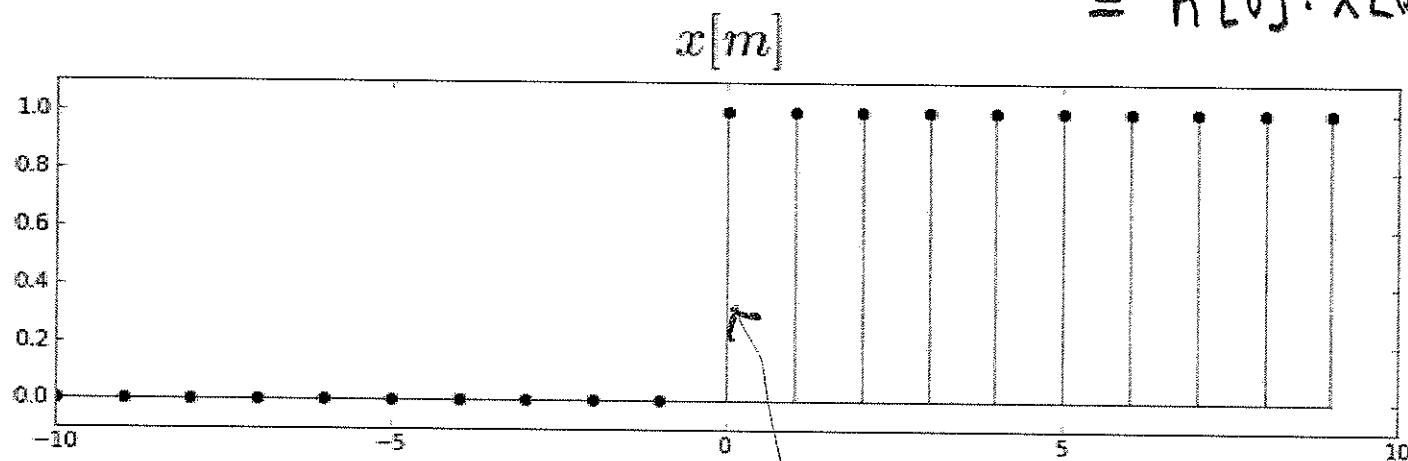


# Example $x[n]$ and $h[n]$



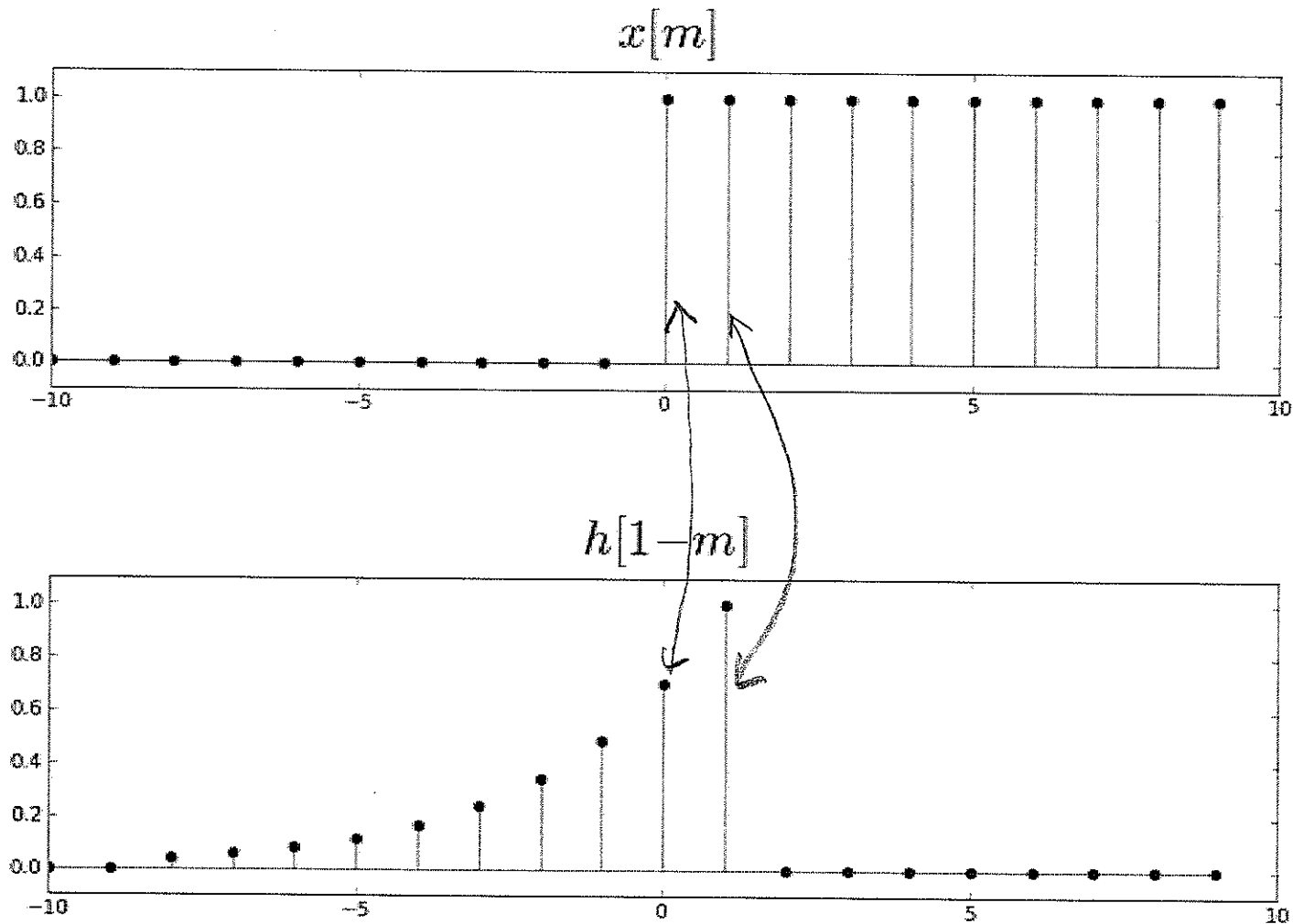
$$\text{Evaluating } y[0] = \sum_{m=0}^p h[0-m] x[m]$$

$$= h[0] \cdot x[0]$$



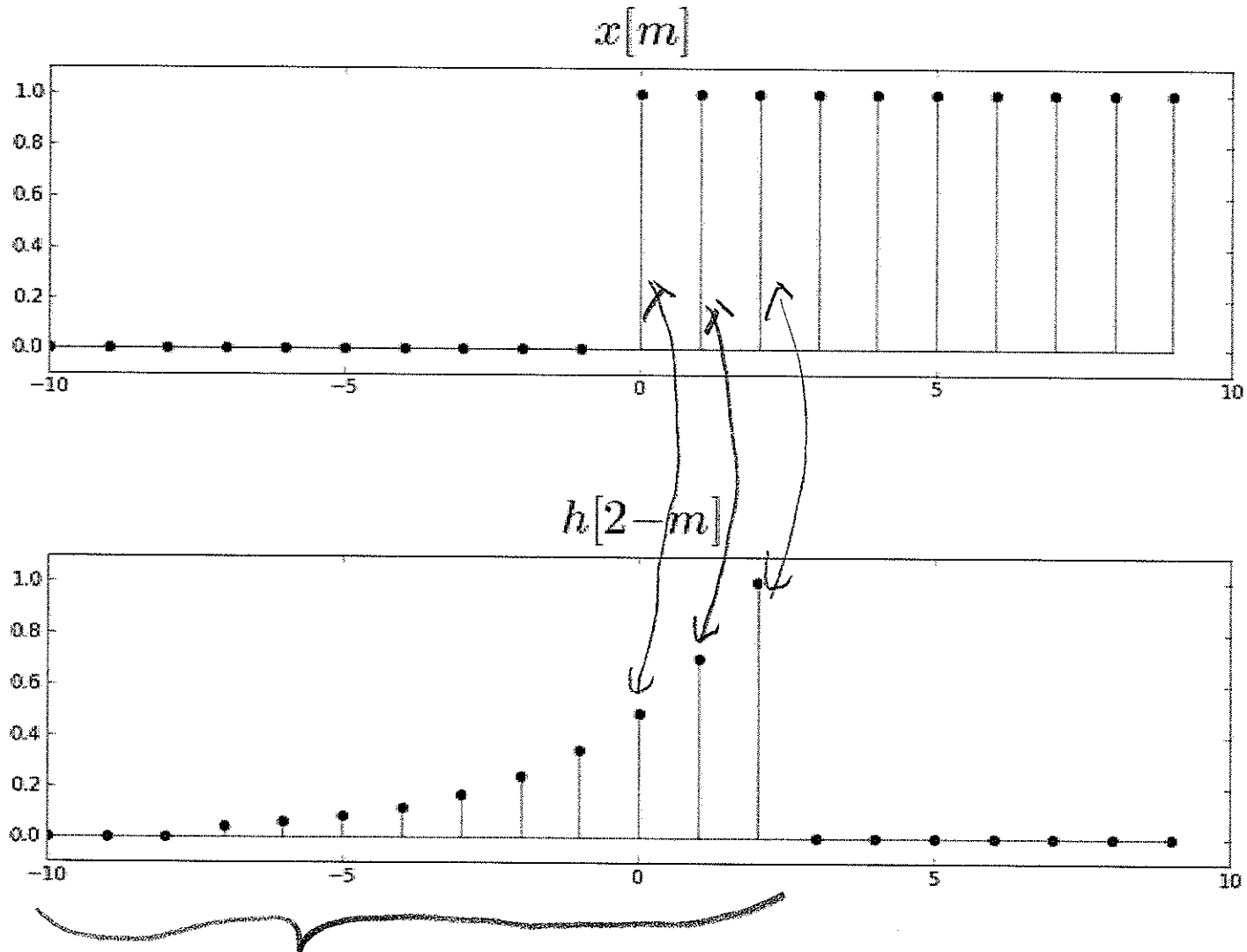
Flipped h

Evaluating  $y[1] = h[0]x[1] + h[1]x[0]$



Flipped  $h$ , slid  
by 1.

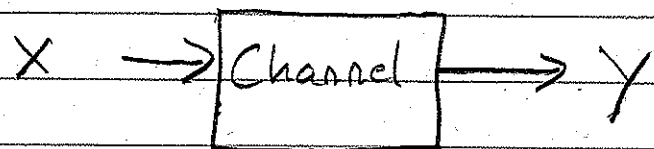
Evaluating  $y[2] = x[0]h[2] + x[1]h[1] + x[2]h[0]$



Flipped  $h$   
Slid by 2

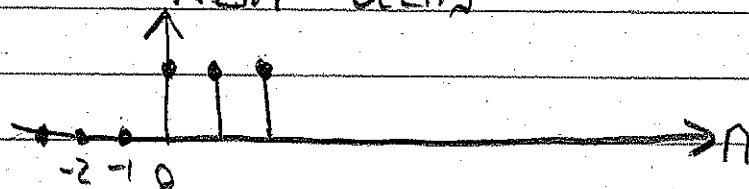
(ZSA)

## Relating Step Response to Unit sample response

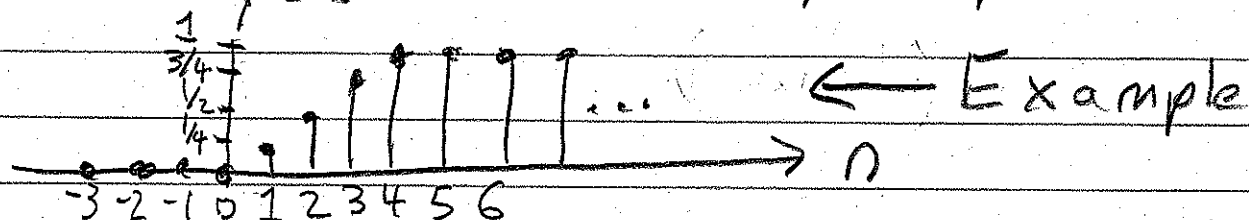


IF  $X[n] = u[n]$

$X[n] = u[n]$

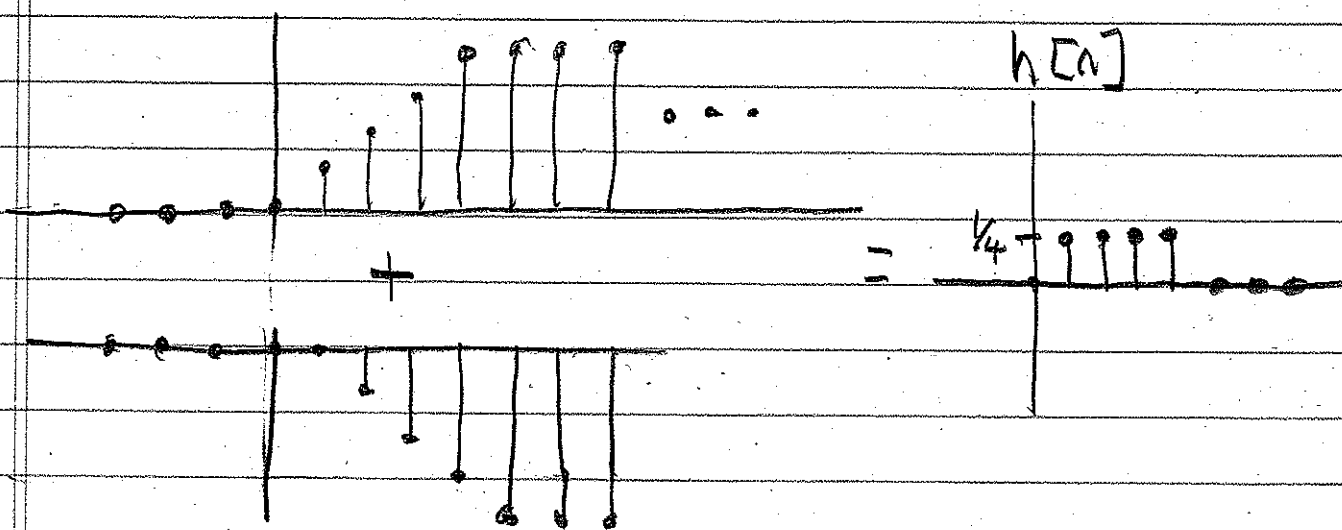


$Y[n] = S[n]$  (step response)



IF  $X[n] = \delta[n]$  then  $X[n] = u[n] - u[n-1]$

$Y[n] = S[n] - S[n-1] = h[n]$



(25B)

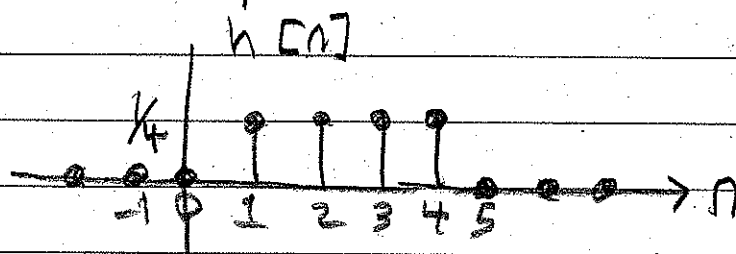
If  $x[n] = u[n]$

$$y[n] = \sum_{m=0}^n h[m] x[n-m]$$

$$= \sum_{m=0}^n h[m] \underbrace{u[n-m]}_{=1 \text{ } n-m \geq 0}$$

$$y[n] = \sum_{m=0}^n h[m]$$

Example



$$y[0] = h[0] = 0$$

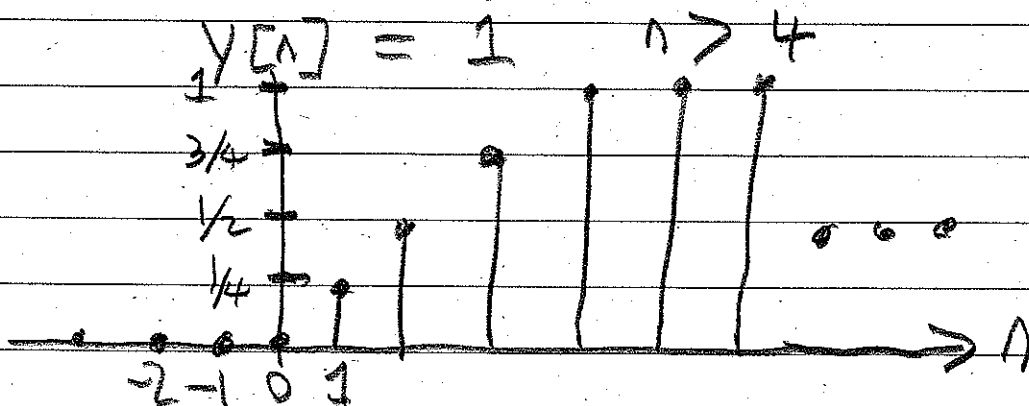
$$y[1] = h[0] + h[1] = 1/4$$

$$y[2] = h[0] + h[1] + h[2] = 1/2$$

$$y[3] = \sum_{m=0}^3 h[m] = 3/4$$

$$y[4] = \sum_{m=0}^4 h[m] = 1$$

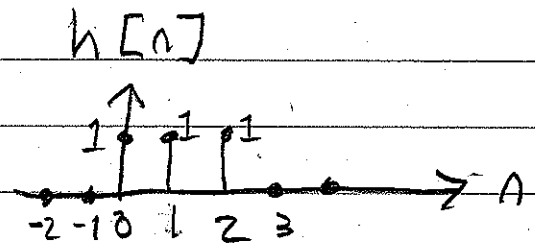
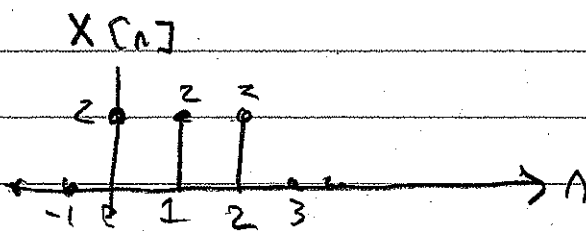
$$y[n] = 1 \quad n > 4$$



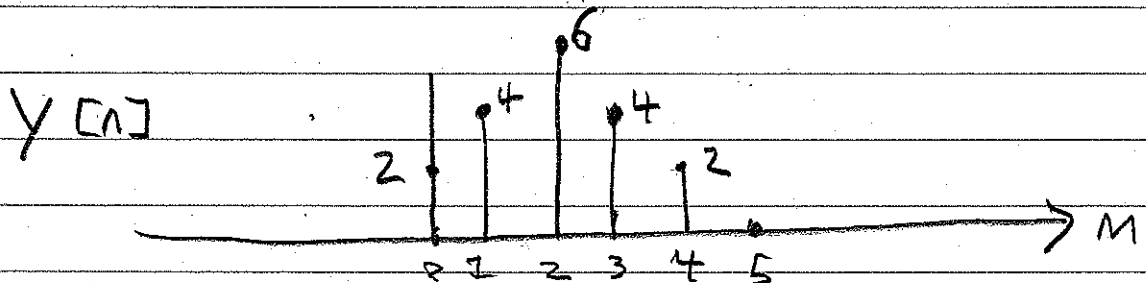
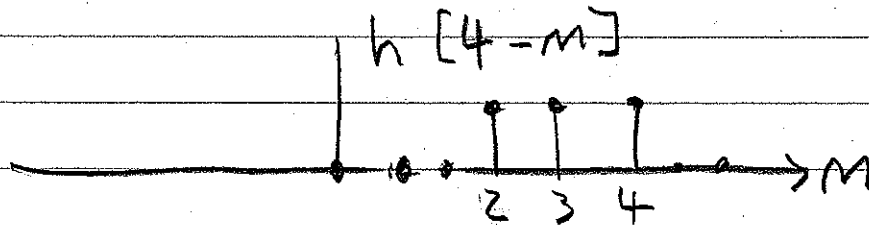
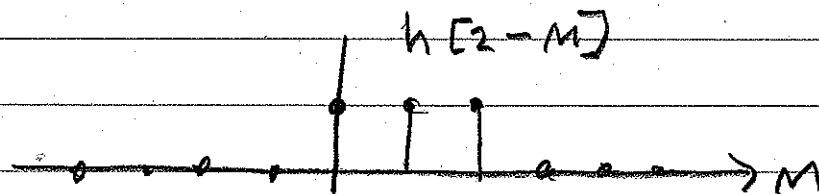
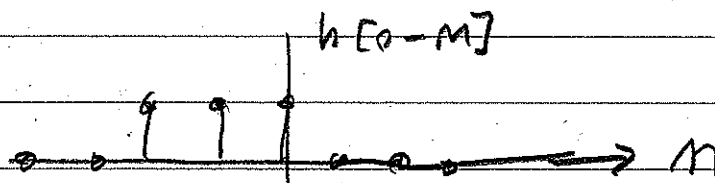
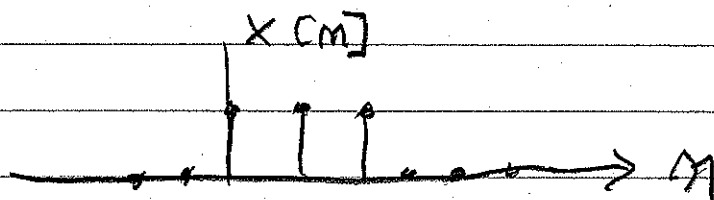
(26)

# Convolution Examples

1)

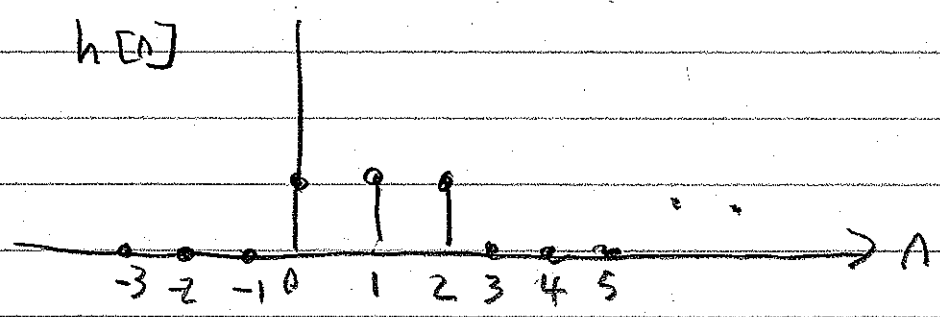
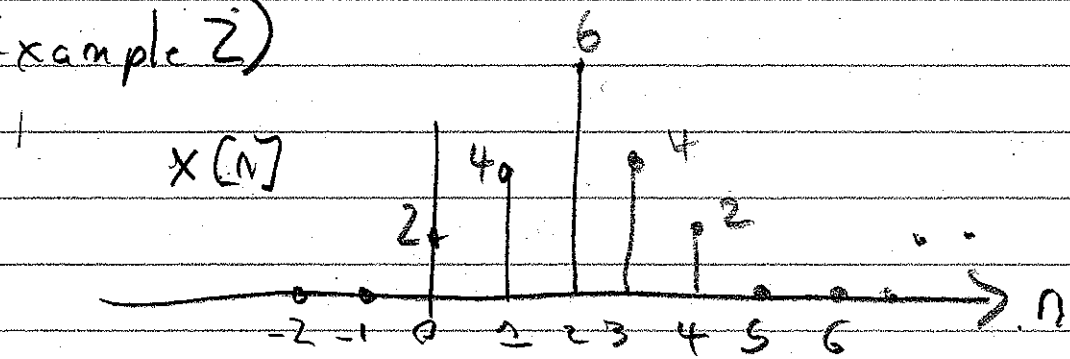


$$Y = H * X$$



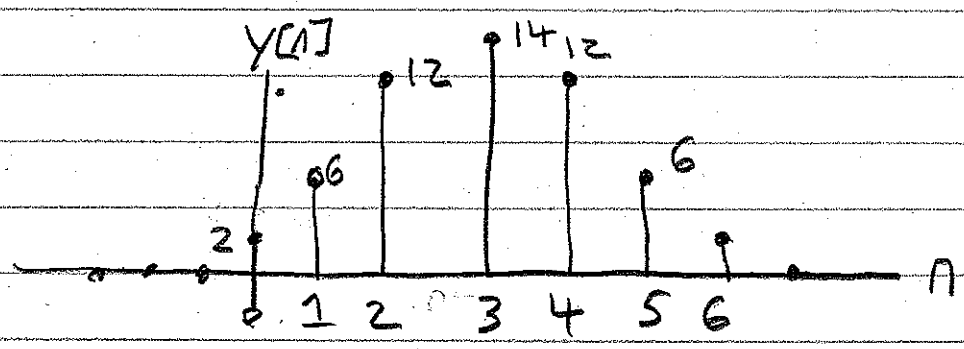
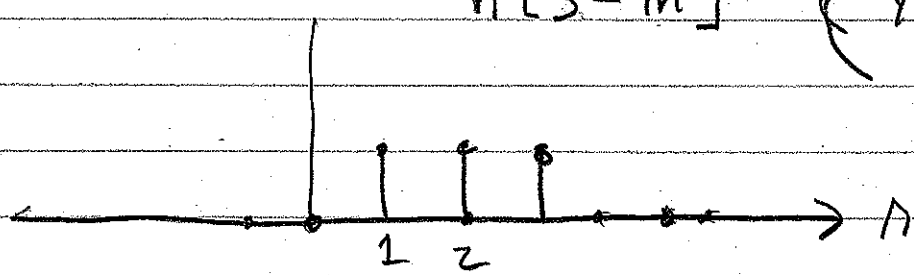
Note  $Y[n]$ 's maximum value is 6 at  $n=2$ .  
Easy to see from flip & slide

# Example 2)



## Flip & Slide

$h[3-m]$  ( $y[n]$  max at  $n=3$ )



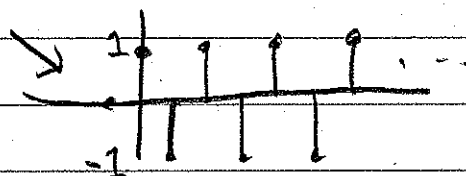
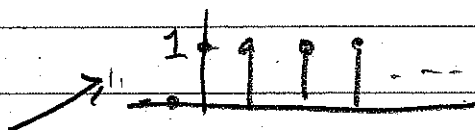
Note Example 1  $\square * \square = \Delta$  } Each Convolution  
 Example 2  $\Delta * \square = \text{trapezoid}$  } 5 smooths



Example 3)

$$x_1[n] = u[n]$$

$$x_2[n] = (-1)^n u[n]$$



$$h[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-2]$$

$$y_1[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-2] + \frac{1}{2} \delta[n-4] + \frac{1}{2} \delta[n-6] + \dots$$

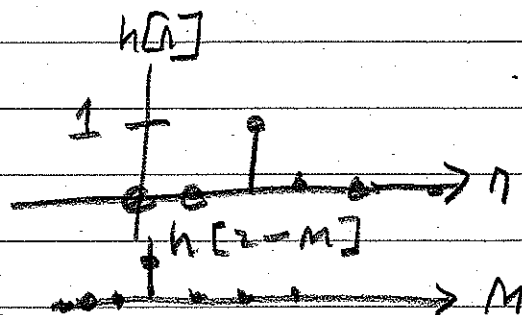
$$y_2[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-2] + \frac{1}{2} \delta[n-4] + \frac{1}{2} \delta[n-6] + \dots$$

Easy  
to  
see  
by  
Flip  
and  
Slide

Example 4)

$$h[n] = \delta[n-2]$$

$$y[n] = x[n-2]$$



# Deconvolution Example

Example 1)

$$x[n] = u[n]$$

$$h[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-1]$$

$$\Rightarrow y[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-1] + \frac{1}{2} \delta[n-2] + \frac{1}{2} \delta[n-3] + \dots$$

$$h[0]w[n] + h[1]w[n-1] = y[n]$$

$$w[0] = \frac{1}{\frac{1}{2}} (y[0] - h[1]w[-1]) = 0$$

$$= 1$$

$$w[1] = \frac{1}{\frac{1}{2}} (y[1] - h[1]w[0])$$

$$= 1$$

$$w[n] = 1 \quad n \geq 0 \Rightarrow \bar{w} = \bar{x}!$$

Example 2)  $x[n] = (-1)^n u[n]$ , same  $h[n]$

$$y[n] = \frac{1}{2} \delta[n] + \frac{1}{2} \delta[n-1]$$

$$\begin{aligned} w[n] &= \left( \frac{1}{h[0]} \right) (y[n] - h[1]w[n-1]) \\ &= 2 (y[n] - \frac{1}{2} w[n-1]) \end{aligned}$$

Example 2 (cont)

$$W[0] = 2 \left( y[0] - \frac{1}{2} W[-1] \right) = 0$$
$$= 2 \left( \frac{1}{2} - 0 \right) = 1$$

$$W[1] = 2 \left( y[1] - \frac{1}{2} W[0] \right) = -1$$

$$W[2] = 2 \left( y[2] - \frac{1}{2} W[1] \right) = 1$$

$$W[n] = (-1)^n u[n]$$

Example 3

$$X[n], \quad h[n] = \delta[n-2]$$

$$\Rightarrow y[n] = x[n-2]$$

Deconvolution Equations:

$$\overset{=0}{h[0]} W[n] + \overset{=0}{h[1]} W[n-1] + \overset{=1}{h[2]} W[n-2] = y[n]$$

$$\Rightarrow W[n-2] = y[n] = x[n-2]$$

$$\underline{\text{or}} \quad W[n] = y[n+2] = x[n]$$

$$\Rightarrow W[n-2] = x[n-2]$$

$$W[n] = x[n]$$

Note

To compute  $W[n]$ , you need  $y[n+2]$ ! The deconvolution equation does NOT describe a causal system!